

Solutions Manual

Second Edition

Field and Wave Electromagnetics

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Chapter 4

Vector Analysis

- Ex 1.1
- $\vec{a} = \frac{\vec{r}}{r} = \frac{x\vec{e}_1 + y\vec{e}_2 + z\vec{e}_3}{\sqrt{x^2 + y^2 + z^2}} = \frac{1}{\sqrt{x^2 + y^2 + z^2}}(x\vec{e}_1 + y\vec{e}_2 + z\vec{e}_3)$
 - $|\vec{a} - \vec{b}| = |\vec{a}_x + \vec{a}_y + \vec{a}_z| = \sqrt{1^2 + 1^2 + 1^2} = \sqrt{3}$
 - $\vec{a} \cdot \vec{b} = 0 = 2(0 \cdot 0) + (-1) = -1$
 - $\vec{a}_y = \cos^2(\vec{a} \cdot \vec{b} \vec{e}_2) = \cos^2(\vec{e}_2 \cdot \vec{e}_2 \sqrt{3} \vec{e}_2) = \cos^2 \pi = 1$
 - $\vec{a} \cdot \vec{a}_y = \vec{a} \cdot \frac{\vec{y}}{y} = \vec{a} \cdot \frac{1}{\sqrt{x^2 + y^2 + z^2}}(x\vec{e}_1 + y\vec{e}_2 + z\vec{e}_3) = \frac{y}{\sqrt{x^2 + y^2 + z^2}}$
 - $\vec{a} \cdot \vec{z} = -\vec{a}_z \cdot \vec{z} = -\vec{e}_3 \cdot \vec{z} = -z$
 - $\vec{a} \cdot (\vec{a} \times \vec{z}) = (\vec{a} \cdot \vec{z}) \cdot \vec{a} = -z\vec{a}$
 - $\vec{z}(\vec{a} \cdot \vec{z}) \cdot \vec{z} = \vec{z}(\vec{z} \cdot \vec{z}) = \vec{z}z = z\vec{z}$
 $\vec{z} \cdot (\vec{a} \times \vec{z}) = \vec{z}(\vec{a} \cdot \vec{z}) = \vec{z}(-z) = -z\vec{z}$

Ex 1.2 Let $\vec{c} = c_1\vec{e}_1 + c_2\vec{e}_2 + c_3\vec{e}_3$,

$$\text{where } c_1^2 + c_2^2 + c_3^2 = 1. \quad \textcircled{*}$$

$$\text{For } \vec{c} \perp \vec{a}: \vec{c} \cdot \vec{a} = 0 \implies c_1 - 2c_2 + 3c_3 = 0. \quad \textcircled{**}$$

$$\text{For } \vec{c} \perp \vec{b}: \vec{c} \cdot \vec{b} = 0 \implies c_1 + c_2 - 2c_3 = 0. \quad \textcircled{***}$$

Solving $\textcircled{*}$, $\textcircled{**}$, and $\textcircled{***}$ simultaneously, we obtain

$$c_1 = \frac{1}{\sqrt{11}}, \quad c_2 = \frac{2}{\sqrt{11}}, \quad c_3 = \frac{1}{\sqrt{11}},$$

$$\text{and } \vec{c} = \frac{1}{\sqrt{11}}(\vec{e}_1 + 2\vec{e}_2 + \vec{e}_3).$$

Ex 1.3 For $\vec{A} \perp \vec{B}$ everywhere, $\vec{A} \cdot \vec{B} = \begin{vmatrix} x_1 & x_2 & x_3 \\ x_1 & x_2 & x_3 \\ x_1 & x_2 & x_3 \end{vmatrix} = 0$,
which requires that $\frac{x_1}{x_2} = \frac{x_2}{x_3} = \frac{x_3}{x_1}$.

Ex 1 From $A \cdot B = A \cdot C$ we have $A \cdot (B-C) = 0$. $\textcircled{1}$
 From $A \cdot B = A \cdot C$ we have $A \cdot (B-C) = 0$. $\textcircled{2}$
 $\textcircled{1}$ implies $A \perp (B-C)$, and $\textcircled{2}$ implies $A \perp (B-C)$.

Since A is not a null vector, $\textcircled{1}$ and $\textcircled{2}$ cannot hold at the same time unless $(B-C)$ is a null vector. Thus, $B=C$, or $B=A$.

Ex 2 Expand $A \times (B+C) = A \times B + A \times C$,
 or $A \times B = \rho A + A \times C$.
 $\therefore C = \frac{1}{\rho} (\rho A + A \times B)$.

Ex 3 Position vector of the three corners

$$\vec{r}_1 = x_1 \hat{i} + y_1 \hat{j}, \quad \vec{r}_2 = x_2 \hat{i} + y_2 \hat{j}, \quad \vec{r}_3 = x_3 \hat{i} + y_3 \hat{j}$$

Vectors representing the three sides of the triangle

$$\vec{r}_{12} = \vec{r}_2 - \vec{r}_1 = x_2 \hat{i} - x_1 \hat{i} + y_2 \hat{j} - y_1 \hat{j}, \quad \vec{r}_{23} = x_3 \hat{i} - x_2 \hat{i} + y_3 \hat{j} - y_2 \hat{j}, \quad \vec{r}_{31} = x_1 \hat{i} - x_3 \hat{i} + y_1 \hat{j} - y_3 \hat{j}$$

$$\therefore \vec{r}_{12} \cdot \vec{r}_{13} = 0. \quad \therefore \triangle r_1 r_2 r_3 \text{ is a right triangle.}$$

$$\therefore \text{Area of triangle} = \frac{1}{2} |\vec{r}_{12} \times \vec{r}_{13}| = 12.$$

Ex 4



$$\vec{r}_1 = B - A, \quad \vec{r}_2 = D - A.$$

$$\vec{r}_1 \cdot \vec{r}_2 = (B - A) \cdot (D - A)$$

$$= B \cdot D - A \cdot A = 0$$

for a rhombus.

$$\therefore \vec{r}_1 \perp \vec{r}_2.$$

Ex. 11



Let $A, B,$ and C denote the vertices of a triangle, and D' and E' be the midpoints of sides AB and AC , respectively. The following vector relations hold:

$$\vec{AD}' = \frac{1}{2} \vec{AB}, \quad \vec{AE}' = \frac{1}{2} \vec{AC}.$$

$$\begin{aligned} \vec{D'E}' &= \vec{AE}' - \vec{AD}' = \frac{1}{2} (\vec{AC} - \vec{AB}) \\ &= \frac{1}{2} \vec{BC}. \end{aligned} \quad \text{Q.E.D.}$$

Ex. 12 $\vec{a}_2 = \vec{a}_1 \cos \alpha + \vec{a}_2' \sin \alpha,$
 $\vec{a}_2' = \vec{a}_2 \cos \beta + \vec{a}_1' \sin \beta.$

a) $\vec{a}_2 - \vec{a}_2' = \cos(\alpha - \beta) = \cos \alpha \cos \beta + \sin \alpha \sin \beta,$

b) $\vec{a}_2 - \vec{a}_2' = \begin{vmatrix} \vec{a}_1 & \vec{a}_2 & \vec{a}_2' \\ \cos \beta & \sin \beta & 0 \\ \sin \alpha & \cos \alpha & 0 \end{vmatrix} = \vec{a}_1 (\sin \alpha \cos \beta - \cos \alpha \sin \beta)$
 $= \vec{a}_1 \sin(\alpha - \beta).$

$\therefore \sin(\alpha - \beta) = \sin \alpha \cos \beta - \cos \alpha \sin \beta.$

Ex. 13



$$\vec{a} + \vec{b} + \vec{c} = \vec{0}.$$

$$\vec{a} \times \vec{a} + \vec{a} \times \vec{b} + \vec{c} \times \vec{a}.$$

$$\vec{c} \times \vec{a} + \vec{c} \times \vec{b} + \vec{b} \times \vec{c}.$$

$$\vec{b} \times \vec{c} + \vec{b} \times \vec{a} + \vec{a} \times \vec{b}.$$

Algebraic relations:

$$a^2 \sin \alpha_2 = b^2 \sin \alpha_1 = c^2 \sin \alpha_3.$$

Hence,

$$\frac{a^2}{\sin \alpha_2} = \frac{b^2}{\sin \alpha_1} = \frac{c^2}{\sin \alpha_3} \quad \left(\frac{\text{Law of Sines}}{\text{Law of Cosines}} \right)$$

Ex 1.1



$$P = (r \cos \theta, r \sin \theta)$$

$$(r \cos \theta - r) \cdot (r \cos \theta - r) + (r \sin \theta - r \sin \theta) = 0,$$

$$\therefore (r \cos \theta) = (r \cos \theta)$$

Ex 1.2 Consider line $L_1: a_1x + b_1y = c_1$, which has a slope equal to $-b_1/a_1$. Draw the normal line passing through the origin and parallel to L_1 as $L_2: a_2x + b_2y = 0$. The position vector of a point (x, y) on L_2 is

$$\vec{r} = a_2x + b_2y.$$

If we introduce the vector $\vec{N} = a_2\hat{i} + b_2\hat{j}$, we can write the equation of L_2 as

$$\vec{N} \cdot \vec{r} = 0.$$

Thus the vector \vec{N} is \perp to L_2 , and is normal to both L_1 and L_2 . It follows that the two lines L_1 and L_2 are perpendicular to each other if and only if their normal vectors \vec{N} and $\vec{N}' = a_1\hat{i} + b_1\hat{j}$ are orthogonal. This which implies

$$a_1a_2 + b_1b_2 = 0, \text{ or } \frac{a_2}{b_2} = -\frac{a_1}{b_1}.$$

That is, the slopes of lines L_1 and L_2 are the negative reciprocals of each other.

Ex 1.3 A) Letting the position vector of a point in the plane be

$$\vec{r} = a_1x + b_1y + c_1\hat{k}$$

and introducing the vector $\vec{N} = a_1\hat{i} + b_1\hat{j} + c_1\hat{k}$, we can write the plane equation as

$$\vec{N} \cdot \vec{r} = c_1 \text{ (constant)}$$

This shows that the projection of the position vector to any point in the plane at P is a constant, and that \vec{n} is a normal vector.

$$b) \quad d_n = \frac{|\vec{r}|}{|\vec{n}|} = \frac{\vec{r}_x \vec{n}_x + \vec{r}_y \vec{n}_y + \vec{r}_z \vec{n}_z}{\sqrt{\vec{n}_x^2 + \vec{n}_y^2 + \vec{n}_z^2}}$$

c) The perpendicular distance from the origin to the plane is

$$d_n = \vec{r} = \frac{a}{|\vec{n}|}$$

For our case, $a = 8$, $|\vec{n}| = \sqrt{3^2 + 4^2 + 5^2} = 7$,
and $d_n = \vec{r} = 8/7$.

Ex 10

$$\vec{r}_1 = -\vec{r}_y \hat{j} - \vec{r}_z \hat{k}, \quad \vec{\sigma}_1 = -\vec{r}_y \hat{j} + \vec{r}_z \hat{k}$$

$$\vec{r}_2 = \vec{r}_x (\cos \phi) + \vec{r}_y (\sin \phi) = \vec{r}_x \cos \phi + \vec{r}_y \sin \phi = \vec{r}_x \frac{1}{2} + \vec{r}_y \frac{\sqrt{3}}{2}$$

$$\vec{\sigma}_2 = \vec{r}_2 - \vec{r}_1 = \vec{r}_x \frac{1}{2} - \vec{r}_y \frac{\sqrt{3}}{2} - \vec{r}_z \hat{k}, \quad |\vec{\sigma}_2| = \sqrt{2}$$

$$\vec{r}_1 \cdot \vec{r}_2 = \vec{r}_1 \cdot \frac{\vec{r}_2}{|\vec{r}_2|} = \frac{\vec{r}_1 \cdot \vec{r}_2}{2} = 1/2$$

Ex 11

$$a) \quad x = r \cos \phi = \frac{r}{2} \cos(2\theta) = -\vec{r}_x$$

$$y = r \sin \phi = \frac{r}{2} \sin(2\theta) = 2\vec{r}_y$$

$$z = z$$

$$b) \quad R = (r^2 + z^2)^{1/2} = (r^2 + z^2)^{1/2} = R$$

$$\theta = \tan^{-1}(r/z) = \tan^{-1}(2\vec{r}_y/z) = 2\theta$$

$$\phi = 2\theta = 2 \tan^{-1}(2\vec{r}_y/z)$$

Ex 12

$$a) \quad \vec{r}_1 = \vec{r}_x \frac{1}{\sqrt{2}} + \vec{r}_y \frac{1}{\sqrt{2}}$$

$$\vec{\sigma}_1 = \frac{1}{2} \left(\frac{1}{\sqrt{2}} \vec{r}_x - \frac{1}{\sqrt{2}} \vec{r}_y \right) = -\vec{r}_x \hat{x}$$

$$b) \quad \vec{r}_2 = \frac{1}{\sqrt{2}} (\vec{r}_x + \vec{r}_y) = \vec{r}_x \frac{1}{\sqrt{2}} + \vec{r}_y \frac{1}{\sqrt{2}} = \frac{1}{\sqrt{2}} (\vec{r}_x + \vec{r}_y)$$

$$z = \cos^{-1}(\vec{r}_1 \cdot \vec{r}_2) = \cos^{-1} \left(\frac{1}{2} \right) = 120^\circ$$

Ex. 11 $\vec{F}_1 = \vec{F}_1 \cos \theta \cos \phi + \vec{F}_2 \sin \theta \cos \phi + \vec{F}_3 \sin \theta \sin \phi = \frac{F_1 \cos \theta \cos \phi}{\sqrt{\cos^2 \theta \cos^2 \phi + \sin^2 \theta \cos^2 \phi + \sin^2 \theta \sin^2 \phi}}$
 $\vec{F}_2 = \vec{F}_2 \cos \theta \cos \phi + \vec{F}_2 \sin \theta \cos \phi + \vec{F}_3 \sin \theta \sin \phi = \frac{F_2 \sin \theta \cos \phi}{\sqrt{\cos^2 \theta \cos^2 \phi + \sin^2 \theta \cos^2 \phi + \sin^2 \theta \sin^2 \phi}}$
 $\vec{F}_3 = -\vec{F}_3 \sin \theta \phi + \vec{F}_3 \cos \theta \phi = \frac{F_3 \sin \theta \sin \phi}{\sqrt{\cos^2 \theta \cos^2 \phi + \sin^2 \theta \cos^2 \phi + \sin^2 \theta \sin^2 \phi}}$

Ex. 12

- (i) $\vec{F}_1 \cdot \vec{F}_2 = \cos \theta$, (ii) $\vec{F}_1 \cdot \vec{F}_3 = \sin \theta \cos \phi$, (iii) $\vec{F}_2 \cdot \vec{F}_3 = \sin \theta \sin \phi$
 (iv) $\vec{F}_1 \cdot \vec{F}_3 = \sin \theta$, (v) $\vec{F}_2 \cdot \vec{F}_3 = \sin \theta \cos \phi$, (vi) $\vec{F}_1 \cdot \vec{F}_2 = \cos \theta$
 (vii) $\vec{F}_1 \cdot \vec{F}_3 = \sin \theta \sin \phi$, (viii) $\vec{F}_2 \cdot \vec{F}_3 = \sin \theta$, (ix) $\vec{F}_1 \cdot \vec{F}_2 = \cos \theta$

Ex. 13 $\vec{F} \cdot d\vec{r} = [F_x xy + F_y (y^2 - x^2)] \cdot (dx dy + dy dx + dz dz)$
 $= 2xy dx + (y^2 - x^2) dy$

(i) Along straight path (1). The equation of (1) is $r = r_0 = \text{const.}$

$$\int_{\text{path (1)}} \vec{F} \cdot d\vec{r} = \int_{\text{path (1)}} [2xy dx + (y^2 - x^2) dy]$$

$$= \int_0^1 \int_0^1 2xy dx + \int_0^1 (1 - x^2) dy$$

$$= 2 \cdot 1 \cdot 1 + 1 = 3$$

(ii) Along path (2). This path has two straight line segments. From (1) to (2) $x=0, dy=dz=0, \vec{F} \cdot d\vec{r} = -x^2 dy$. From (2) to (1) $y=0, dx=dz=0, \vec{F} \cdot d\vec{r} = x^2 dx$. Hence,

$$\int_{\text{path (2)}} \vec{F} \cdot d\vec{r} = \int_0^1 (-x^2) dy + \int_0^1 x^2 dx = -1 + 1 = 0$$

Path (2) $\int_{\text{path (2)}} \vec{F} \cdot d\vec{r} = 0$ Vector field \vec{F} is not conservative.

Ex. 14 $\int_C \vec{F} \cdot d\vec{r} = \int_C (y dx + x dy)$

Let $u = x^2, dv = 2y dy$; $\int_C \vec{F} \cdot d\vec{r} = \int_C (x^2 y^2 + 2y^2) dy = 10$

$$b) \text{ as } \gamma_1 = t, \text{ so } \omega_{\gamma_1} = \int_0^{\pi/2} \frac{1}{2} dt = \frac{1}{2} \int_0^{\pi/2} dt = \frac{1}{2} \left(\frac{\pi}{2} - 0 \right) = \frac{\pi}{4}$$

Equal line integrals along two specific paths do not necessarily imply a conservative field. \oint_C is a conservative field in this case because $\oint_C = 0$ ($\gamma_2 = 1$).

Ex 11

$$\begin{bmatrix} \mathcal{L}_1 \\ \mathcal{L}_2 \end{bmatrix} = \begin{bmatrix} \cos t & \sin t \\ \sin t & \cos t \end{bmatrix} \begin{bmatrix} \mathcal{L}_1 \\ \mathcal{L}_2 \end{bmatrix} = \begin{bmatrix} \cos t & \sin t \\ \sin t & \cos t \end{bmatrix} \begin{bmatrix} -\cos t \\ \sin t \end{bmatrix}$$

$$\mathcal{L}_1 = \mathcal{L}_1 \cos t + \mathcal{L}_2 \sin t$$

$$\mathcal{L}_2 = \mathcal{L}_1 \sin t + \mathcal{L}_2 \cos t$$

$$\text{If } \mathcal{L}_1 = \mathcal{L}_2 = 0 \Rightarrow \mathcal{L}_1(1 - \cos t) = \mathcal{L}_2(\sin t - \cos t) \Rightarrow \mathcal{L}_1(1 - \cos t) = \mathcal{L}_2(\sin t - \cos t)$$

There is no change in t ($t=0$) from \mathcal{L}_1 to \mathcal{L}_2 .

$$\therefore \int_0^{\pi/2} \mathcal{L}_1 dt = \int_0^{\pi/2} \mathcal{L}_2 dt = \int_0^{\pi/2} \cos t dt = \sin t \Big|_0^{\pi/2} = 1 - 0 = 1$$

$$\text{Ex 12} \quad a) \text{ If } r = \left(\mathcal{L}_1 \frac{1}{\sqrt{2}} \cos \frac{t}{\sqrt{2}} + \mathcal{L}_2 \frac{1}{\sqrt{2}} \sin \frac{t}{\sqrt{2}} \right) \mathbf{i} + \left(\mathcal{L}_1 \frac{1}{\sqrt{2}} \sin \frac{t}{\sqrt{2}} - \mathcal{L}_2 \frac{1}{\sqrt{2}} \cos \frac{t}{\sqrt{2}} \right) \mathbf{j} + t^2 \mathbf{k}$$

$$\Rightarrow \mathcal{L}_1 \mathbf{i} = - \left(\mathcal{L}_1 \frac{1}{\sqrt{2}} + \mathcal{L}_2 \frac{1}{\sqrt{2}} \right) t^2 \mathbf{j} + \left(\mathcal{L}_1 \cos t + \mathcal{L}_2 \sin t \right) \mathbf{k}$$

$$b) \mathcal{L}_1 = -\mathcal{L}_2 = \mathcal{L}_3 = \mathcal{L}_4 \quad \mathcal{L}_5 = \frac{1}{\sqrt{2}} (\mathcal{L}_3 + \mathcal{L}_4) = \mathcal{L}_3$$

$$\therefore \mathcal{L}_1 \mathbf{i} = \mathcal{L}_3 = \frac{1}{\sqrt{2}} \left(\frac{1}{\sqrt{2}} + \frac{1}{\sqrt{2}} \right) t^2 = t^2 \mathbf{j}$$

Ex 13 On the surface of the sphere, $\mathcal{L} = 5$.

$$\begin{aligned} \int_C (\mathcal{L}_1 \cos t) dt &= \int_0^{\pi/2} \int_0^{2\pi} (\mathcal{L}_1 \cos t) \cdot (\mathcal{L}_2 \sin t) dt d\theta \\ &= \int_0^{\pi/2} \int_0^{2\pi} 5 \cos t \sin t dt d\theta = 5\pi^2 \end{aligned}$$

■

Ex 14 The first step is to find the equation for the unit normal $\mathbf{n}_n = \mathcal{L}_1 \mathbf{i} + \mathcal{L}_2 \mathbf{j} + \mathcal{L}_3 \mathbf{k}$ to the plane surface. This gives four corner points of the surface and the following four equations



Curve (A,B) $\int_a^b y \, dx = \int_a^b y \, dx = 0$ \odot

Curve (B,C) $\int_b^c y \, dx = \int_b^c y \, dx = 0$ \odot

Curve (C,D) $\int_c^d y \, dx = \int_c^d y \, dx = 0$ \odot

Curve (D,A) $\int_d^a y \, dx = \int_d^a y \, dx = 0$ \odot

The direction cosines satisfy the condition: $\int_a^b dx^2 + dy^2 = 1$ \odot

From \odot - \odot we obtain $A = 0$, and $\cos \theta = \frac{dx}{ds}$

Thus, $\int_a^b dx = \int_a^b \cos \theta \, ds$, $\int_a^b dx = \int_a^b \cos \theta \, ds$ in context
and $\int_a^b dx = \int_a^b \cos \theta \, ds = \int_a^b (1) \, ds = 2a$.

Ex. 106 In spherical coordinates, $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x_1, x_2, x_3)$, $x_1 = x$, $x_2 = y$.

as $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x, y, z)$, $x_1 = x$, $x_2 = y$, $x_3 = z$
 $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x, y, z) = (x, y, z)$

as $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x, y, z)$, $x_1 = x$, $x_2 = y$, $x_3 = z$
 $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x, y, z) = (x, y, z)$

Ex. 107 For radial vector $\mathbf{r} = x_1 \mathbf{i}$, $\mathbf{r} = \frac{1}{\sqrt{2}} \sqrt{2} (x, y, z) = (x, y, z)$

Using all previous theorem, we have

$$\frac{1}{\sqrt{2}} \int_a^b \mathbf{r} \cdot d\mathbf{r} = \frac{1}{\sqrt{2}} \int_a^b \mathbf{r} \cdot \mathbf{r} \, ds = \frac{1}{\sqrt{2}} (1) = \frac{1}{\sqrt{2}}$$

Ex. 108 $\int_a^b \mathbf{r} \cdot d\mathbf{r} = \left(\int_a^b x_1^2 + \int_a^b x_2^2 + \int_a^b x_3^2 \right) \mathbf{r} \cdot d\mathbf{r}$

On the outside face:

$$\int_{\text{outside face}} \vec{T} \cdot d\vec{A} = A_1 \left(\hat{x} + \frac{y}{R} \hat{y} - \hat{z} \right) \cdot \left(\hat{x} + \frac{y}{R} \hat{y} + \hat{z} \right) dA \\ = \left[A_1 \left(x_1^2 + y_1^2 + z_1^2 + \frac{y_1^2}{R} \right) \right]_{R,0,0}^{R,R,0} = 4R^2 A_1 \quad \textcircled{2}$$

Adding ① and ②, we have

$$\left[\int_{\text{inside}} + \int_{\text{outside}} \right] \vec{T} \cdot d\vec{A} = \left(A_1 + \frac{y_1^2}{R} \right) \left[\text{area} + 4R^2 \right] \\ = \frac{1}{R} \left(R A_1 \right) \left[\text{area} + 4R^2 \right] \quad \textcircled{3}$$

where R.A.T. contains internal and higher powers of dy .
The sum of the contributions of the front and back faces (both of area $= 4R^2$) is

$$\left[\int_{\text{front}} + \int_{\text{back}} \right] \vec{T} \cdot d\vec{A} = \frac{4A_1}{R} \left[\text{area} + 4R^2 \right] = 4R A_1 T_1 \quad \textcircled{4}$$

where R.A.T. contains internal and higher powers of dy .
Similarly, the sum of the contributions of the top and bottom faces (both of area $= 4R^2$) is

$$\left[\int_{\text{top}} + \int_{\text{bottom}} \right] \vec{T} \cdot d\vec{A} = \left(r + \frac{y_1^2}{R} \right) \left[\text{area} + 4R^2 \right] \quad \textcircled{5}$$

where R.A.T. contains internal and higher powers of dy .

Combining ③, ④ and ⑤ in ①, dividing by the negative and letting $dV = 4R^2 dA = 0$, we get

$$\vec{T} \cdot \vec{A} = \frac{1}{R} \frac{d}{dy} (R A_1) = \frac{1}{R} \frac{dA_1}{dy} = \frac{dA_1}{dy}$$

where the subscript 1 has been dropped for simplicity.

Ex. 2.11 a) $\vec{r} = x_1 \hat{e}_1 + x_2 \hat{e}_2$, $d\vec{r} = dx_1 \hat{e}_1 + dx_2 \hat{e}_2$.

$$\oint_C \vec{r} \cdot d\vec{r} = \int_0^1 \int_0^1 (x_1 + x_2) dx_1 dx_2 = \frac{1}{2}.$$

b) $\vec{r} = x_1 \hat{e}_1 - x_2 \hat{e}_2$, $d\vec{r} = dx_1 \hat{e}_1 + dx_2 \hat{e}_2$.

$$\oint_C \vec{r} \cdot d\vec{r} = \int_0^1 \int_0^1 (-x_2) dx_1 dx_2 = 0.$$

Ex. 2.12 $\vec{r} = x_1 x_2 \hat{e}_1 + x_2^2 \hat{e}_2$, $d\vec{r} = dx_1 \hat{e}_1 + dx_2 \hat{e}_2$.



$$a) \oint_C \vec{r} \cdot d\vec{r} = \int_0^1 \int_0^1 (x_1 x_2 + 2x_2) dx_1 dx_2$$

$$\text{Part (1): } \int_0^1 \int_0^1 x_1 x_2 dx_1 dx_2 = \frac{1}{6}$$

$$\text{Part (2): } \int_0^1 \int_0^1 2x_2 dx_1 dx_2 = \frac{1}{2}$$

$$\text{Part (3): } \int_0^1 \int_0^1 x_1 x_2 dx_1 dx_2 = \frac{1}{6}$$

$$\therefore \oint_C \vec{r} \cdot d\vec{r} = \frac{1}{6} + \frac{1}{2} + \frac{1}{6} = \frac{2}{3}$$

b) $\vec{r} = -x_1 x_2 \hat{e}_1 + x_2^2 \hat{e}_2$, $d\vec{r} = dx_1 \hat{e}_1 + dx_2 \hat{e}_2$.

$$\oint_C \vec{r} \cdot d\vec{r} = \int_0^1 \int_0^1 (-x_1 x_2 + 2x_2) dx_1 dx_2 = \frac{1}{2}$$

Ex. 2.13 $\oint_C \vec{r} \cdot d\vec{r} = \frac{1}{2} \oint_C (\vec{r} \cdot d\vec{r}) = \frac{1}{2} \oint_C (\vec{r} \cdot d\vec{r})$ (1)



where a, b, c, d are the vertices

and the number opposite of the four sides (1, 2, 3, 4).

Side 1: $d\vec{r} = dx_1 \hat{e}_1$

$$\vec{r} \cdot d\vec{r} = x_1 dx_1 = \frac{1}{2} x_1^2 \Big|_0^a$$

where $a, b, c, d = \frac{1}{2} (x_1^2 + x_2^2 + x_3^2)$

$$= \frac{1}{2} \left(\frac{1}{2} a^2 + \frac{1}{2} b^2 + \frac{1}{2} c^2 \right)$$

$$\oint_C \vec{r} \cdot d\vec{r} = \left[\frac{1}{2} (x_1^2 + x_2^2 + x_3^2) \right]_{a,b,c,d} = \frac{1}{2} (a^2 + b^2 + c^2 + d^2)$$

Solve 1: $\dot{x} = -x_2$ ($\dot{x}_1 = 0$), $\dot{x} = x_1$ ($\dot{x}_2 = 0$) $\Rightarrow \frac{dx_1}{dt} = 0$, $\frac{dx_2}{dt} = -x_1$

$$\int_{x_1(0)}^{x_1(t)} dx_1 = - \left[x_2 + \frac{t^2}{2} \right]_{x_2(0)} + x_2(0) \quad \text{①}$$

Combining ① and ②:

$$\int_{x_1(0)}^{x_1(t)} dx_1 = \left(- \frac{t^2}{2} + x_2(0) \right) \Big|_{x_2(0)} - x_2(0) \quad \text{③}$$

Solve 2: $\dot{x} = x_2$ ($\dot{x}_1 = 0$), $\dot{x} = x_1$ ($\dot{x}_2 = 0$) $\Rightarrow \frac{dx_1}{dt} = 0$, $\frac{dx_2}{dt} = x_1$

$$\int_{x_1(0)}^{x_1(t)} dx_1 = \left[x_2 + \frac{t^2}{2} \right]_{x_2(0)} + x_2(0) \Big|_{x_1(0)} - x_2(0) \quad \text{④}$$

Solve 3:

$$\int_{x_1(0)}^{x_1(t)} dx_1 = \left[x_2 + \frac{t^2}{2} \right]_{x_2(0)} + x_2(0) \Big|_{x_1(0)} - x_2(0) \quad \text{⑤}$$

Combining ③ and ⑤:

$$\int_{x_1(0)}^{x_1(t)} dx_1 = \frac{t^2}{2} \Big|_{x_2(0)} - x_2(0) + x_2(0) + x_2(0) - x_2(0) + x_2(0)$$

$$= \frac{t^2}{2} (x_2(0) + x_2(0)) - x_2(0) + x_2(0) \quad \text{⑥}$$

Substituting ③, ④ and ⑤ in ⑥ we obtain:

$$(P \cdot K)_2 = \frac{1}{2} \left[\frac{t^2}{2} (x_2(0) + x_2(0)) - \frac{t^2}{2} \right] =$$

where the subscript 2 has been dropped for simplicity.

Solve 4: $\dot{P} = \frac{1}{2} \frac{d}{dt} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2})$

$$\int_0^t (P \cdot K)_2 dt = \int_0^t \frac{1}{2} \frac{d}{dt} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2}) dt = \frac{1}{2} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2}) \Big|_0^t = \frac{1}{2} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2}) - 0$$

$$\int_0^t \dot{K} \cdot dt = \int_0^t \frac{d}{dt} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2}) dt = \int_0^t \frac{d}{dt} (x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2}) dt = x_2 \sin t \cos \frac{t}{2} - x_1 \sin t \cos \frac{t}{2} - 0$$

Solve 5: $\dot{P} = x_2(x_1 + x_2) + x_1(x_2 - x_1) + x_2^2(x_1 + x_2) + x_1^2(x_2 - x_1)$

a) \dot{P} (rotational) $\rightarrow P \cdot \dot{P} = 0$

$$\Rightarrow x_2 \left(\frac{2x_2}{2} - \frac{2x_1}{2} \right) + x_1 \left(\frac{2x_2}{2} - \frac{2x_1}{2} \right) + x_2^2 \left(\frac{2x_2}{2} - \frac{2x_1}{2} \right) + x_1^2 \left(\frac{2x_2}{2} - \frac{2x_1}{2} \right) = 0$$

which gives three equations:

$$\frac{\partial}{\partial x}(a + c_1 y + c_2 z) = \frac{\partial}{\partial x}(c_1 x - 2xz) = 0 \implies c_1 + 2xz = 0 \implies c_1 = -2$$

$$\frac{\partial}{\partial y}(a + c_1 y + c_2 z) = \frac{\partial}{\partial y}(a + c_1 y + c_2 z) = 0 \implies c_1 - c_2 = 0 \implies c_1 = c_2$$

$$\frac{\partial}{\partial z}(a + c_1 y + c_2 z) = \frac{\partial}{\partial z}(c_2 x - 2xz) = 0 \implies c_2 = 0$$

∴ \vec{F} is also irrotational $\implies \vec{F} = \nabla \phi = 0$.

$$\text{or } \frac{\partial \phi}{\partial x} + \frac{\partial \phi}{\partial y} + \frac{\partial \phi}{\partial z} = 0.$$

$$\text{or } \frac{\partial}{\partial x}(a + c_1 x) + \frac{\partial}{\partial y}(c_1 y - 2xz) + \frac{\partial}{\partial z}(c_1 y + c_2 z) = 0$$

$$\text{or } 1 + c_1 = 0 \implies c_1 = -1.$$

∴ $\vec{F} = -\nabla \phi \implies 2x(x+2) - 2y, 2z + c_1(y-z)$

$$= -c_1 \frac{\partial \phi}{\partial x} - c_2 \frac{\partial \phi}{\partial y} - c_3 \frac{\partial \phi}{\partial z}$$

$$\frac{\partial \phi}{\partial x} = -2x(x+2) \implies \phi = -\frac{x^2}{2} - 2x + f(y, z)$$

$$\frac{\partial \phi}{\partial y} = 2z \implies \phi = 2yz + f_1(y, z)$$

$$\frac{\partial \phi}{\partial z} = 2x + 2y + 2 \implies \phi = 2xz + 2yz + \frac{z^2}{2} + f_2(y, z)$$

$$\therefore \phi = -\frac{x^2}{2} + 2xz + 2yz + \frac{z^2}{2}$$

Chapter 4

Solutions of Electrostatic Problems

Ex 1. Use subscripts 1 and 2 to denote dielectrics and air regions respectively. $\nabla^2 V = 0$ in both regions.

$$V_1 = a_1 x + a_2, \quad E_1 = -a_1 \hat{x}, \quad D_1 = -\epsilon_0 a_1 \hat{x}$$

$$V_2 = a_3 x + a_4, \quad E_2 = -a_3 \hat{x}, \quad D_2 = -\epsilon_0 a_3 \hat{x}$$

BC: at $y=0$, $V_1 = V_2$, at $y=d$, $V_1 = V_2$
 at $y=0$: $V_1 = V_2$, $E_1 = E_2$

Solving: $a_1 = \frac{V_0}{d}$, with $a_2 = \frac{V_0 d}{2d+d}$, $a_3 = \frac{V_0 d}{2d+d}$

(1) $E_1 = \frac{V_0}{d} \hat{x}$, $E_2 = -\frac{V_0}{d} \hat{x}$

(2) $D_1 = \frac{V_0}{d} \hat{x}$, $D_2 = -\frac{V_0}{d} \hat{x}$

(3) $\sigma_{\text{free}} = -\epsilon_0 \frac{V_0}{d}$

$\sigma_{\text{ind}} = \epsilon_0 \frac{V_0}{d}$

Ex 2. At a point where V is a maximum (minimum), the partial derivatives of V with respect to x , y , and z would all be negative (positive); their sum could not vanish, as required by Laplace's equation.

Ex 3. Solve Laplace eq. $\nabla^2 V = 0 \rightarrow \frac{1}{r} \frac{d}{dr} \left(r \frac{dV}{dr} \right) = 0$

Solution: $V = \frac{A}{r} + B$, let $r = a$.

BC: at $r = a$, $V = \frac{A}{a} + B = V_0$, $a = \frac{V_0 - B}{A/a}$

at $r = b$, $V = \frac{A}{b} + B = 0$, $a = \frac{b(V_0 - B)}{-B}$

Ex 4



$$\oint \epsilon_0 \epsilon_r \frac{dV}{dr} 2\pi r dr = \frac{Q}{\epsilon_0 \epsilon_r} \Rightarrow \frac{dV}{dr} = \frac{Q}{2\pi r^2 \epsilon_0 \epsilon_r}$$

(1) $E = -\frac{Q}{2\pi r^2 \epsilon_0 \epsilon_r} \hat{r}$

(2) $\int E dr = -\frac{Q}{2\pi \epsilon_0 \epsilon_r} \left(\frac{1}{a} - \frac{1}{b} \right)$

Ex. 11



Consider the coordinates in the xy-plane (z=0).

a) $V_0 = \frac{q}{4\pi\epsilon_0} \left(\frac{1}{x} - \frac{1}{x'} + \frac{1}{y} - \frac{1}{y'} \right)$, where

$$x' = [(x-a)^2 + y^2 + d^2]^{1/2}, \quad y' = [(x-a)^2 + (y-b)^2 + d^2]^{1/2}$$

$$x_0 = [(x+a)^2 + y^2 + d^2]^{1/2}, \quad y_0 = [(x+a)^2 + (y-b)^2 + d^2]^{1/2}$$

$$\begin{aligned} E_x = -\nabla V_0 &= -E_x \frac{\partial}{\partial x} - E_y \frac{\partial}{\partial y} \\ &= \frac{q}{4\pi\epsilon_0} \left[-\frac{x}{x^2} - \frac{x}{x_0^2} - \frac{y}{y^2} + \frac{y}{y_0^2} \right] \\ &\quad + E_y \frac{\partial}{\partial y} \left[-\frac{y}{y^2} + \frac{y}{y_0^2} - \frac{x}{x^2} + \frac{x}{x_0^2} \right]. \end{aligned}$$

E_x will have a z-component if the point P does not lie in the xy-plane.

b) On the conducting back-plate, $E_x = E_z = 0 = E_y$.

Along the wires, $E_x = [(x-a)^2 + y^2 + d^2]^{-3/2} = E_y$.

and $E_z = [(x+a)^2 + (y-b)^2 + d^2]^{-3/2} = E_y$.

$$E_x = 0, \quad E_y = \frac{q}{4\pi\epsilon_0} \left(\frac{1}{y} - \frac{1}{y_0} \right)$$

$$\begin{aligned} \therefore E(\text{wire}) &= \frac{q}{4\pi\epsilon_0} \left\{ \frac{1}{[(x-a)^2 + y^2 + d^2]^{3/2}} - \frac{1}{[(x+a)^2 + (y-b)^2 + d^2]^{3/2}} \right\} \\ &= \begin{cases} 0, & \text{at } x=0, \\ \text{wire, at } x=d. \end{cases} \end{aligned}$$

Analogously for $E_z(\text{wire})$ on the vertical conducting conducting plane by changing x to y and d to a .

Ex. 12 Refer to Example 4-4

$$E^{\text{in}} = \frac{2\sigma_0}{\epsilon_0} \frac{1}{\sqrt{(x-a)^2 + (y-b)^2 + d^2}} = \frac{2\sigma_0}{\sqrt{a^2 + b^2 + d^2}} \quad (0,0,0)$$

Ex. 11 From our C_{12} in problem 10-10

Ex. 12 (a) From Eqs. (10-18) and (10-19)

$$V_2 = \frac{Q_2}{4\pi\epsilon_0} \left[\frac{1}{r_2} + \frac{2Q_1}{r_2(r_2 - a)} \right]$$

$$E_r = -\hat{r}_2 \frac{dV_2}{dr_2} = -\hat{r}_2 \frac{d}{dr_2} \left[\frac{Q_2}{4\pi\epsilon_0} \left(\frac{1}{r_2} + \frac{2Q_1}{r_2(r_2 - a)} \right) \right]$$

A potential for our everywhere tangent to the electric field lines is obtained by requiring

$$\frac{dV}{dr} = \frac{dV_2}{dr_2} = -\frac{dV_2}{dr_2}$$

which reduces to $\frac{dV}{dr} = \frac{dV_2}{dr_2} = -\frac{dV_2}{dr_2}$

Integrating, we obtain $\int dV = -\int \frac{dV_2}{dr_2} dr_2$
 or $\int dV = -\int \frac{dV_2}{dr_2} dr_2$

where K is a constant — choice of null point having center of C_1, C_2



$$V = -\frac{Q_2}{4\pi\epsilon_0} \ln \frac{r_2}{a} + V_2 = \frac{Q_2}{4\pi\epsilon_0} \ln \frac{a}{r_2} + V_2$$

Capacitance per unit length
 $C = \frac{Q_2}{V} = \frac{Q_2}{\frac{Q_2}{4\pi\epsilon_0} \ln \frac{a}{r_2} + V_2}$

For equation:

$$\begin{aligned} \epsilon_1 &= \epsilon_2 \epsilon_r & \epsilon_1 &= \epsilon_2 \epsilon_r \\ \epsilon_1 &= \epsilon_2 \epsilon_r & \epsilon_1 &= \epsilon_2 \epsilon_r \end{aligned}$$

Let us obtain

$$\frac{dV}{dr} = \frac{dV_2}{dr_2} \quad \text{and} \quad \epsilon_1^2 = \epsilon_2^2 \epsilon_r^2 = \epsilon_2 \epsilon_r = \epsilon_1$$

$$\epsilon_1 \epsilon_2 = \epsilon_1^2 \epsilon_r = \epsilon_1 = \epsilon_2 \epsilon_r = \epsilon_1$$

$$\begin{aligned} C &= \frac{Q_2}{V} = \frac{Q_2}{\frac{Q_2}{4\pi\epsilon_0} \ln \frac{a}{r_2} + V_2} \\ &= \frac{4\pi\epsilon_0 Q_2}{\ln \frac{a}{r_2} + \frac{V_2}{\frac{Q_2}{4\pi\epsilon_0}}} \quad (10-12) \end{aligned}$$

Ex. 10.11 d_p (small) $v_p = \frac{d_p}{2}(\omega_p^2 - \omega^2 - \omega_0^2)$, d_p (large) $v_p = \frac{d_p}{2}(\omega_p^2 + \omega^2)$



d_p (small): $\omega^2 = \omega_p^2 - \omega_0^2$

d_p (large): $\omega^2 = \omega_p^2 + \omega_0^2$

$$\Rightarrow v = \frac{d_p}{2} \ln \left[\frac{\frac{d_p}{2}(\omega_p^2 - \omega_0^2) + \frac{d_p}{2}(\omega_p^2 + \omega_0^2)}{\frac{d_p}{2}(\omega_p^2 - \omega_0^2) - \frac{d_p}{2}(\omega_p^2 + \omega_0^2)} \right]$$

At b : $v = 0 = d_p \omega_0^2$, $\omega = 0 = \omega_p \omega_0$

At a : $v = 0 = d_p \omega_0^2$, $\omega = 0 = \omega_p \omega_0$

$$v - v_0 = \frac{d_p}{2} \ln \left[\frac{\frac{d_p}{2}(\omega_p^2 - \omega_0^2) + \frac{d_p}{2}(\omega_p^2 + \omega_0^2)}{\frac{d_p}{2}(\omega_p^2 - \omega_0^2) - \frac{d_p}{2}(\omega_p^2 + \omega_0^2)} \right]$$

Expressing ω and ω_0 in terms of b and a

$$\text{and simplifying: } v - v_0 = \frac{d_p}{2} \ln \left[\left(\frac{a^2 + b^2 + r_p^2}{a^2 + b^2 - r_p^2} \right) \left(\frac{a^2 + b^2 + r^2}{a^2 + b^2 - r^2} \right) \right]$$

$$e^{2(v-v_0)/d_p} = \frac{(a^2 + b^2 + r_p^2)(a^2 + b^2 + r^2)}{(a^2 + b^2 - r_p^2)(a^2 + b^2 - r^2)} \quad (10.1)$$

$$\Rightarrow \text{For } v = v_0 \text{ and } b = 0: \frac{a^2 + r_p^2}{a^2 - r_p^2} = \frac{a^2 + r^2}{a^2 - r^2} \quad (10.2)$$

Ex. 10.12



$$v_p = -\frac{d_p}{2} \omega_0^2, \quad v_0 = \frac{d_p}{2} \omega_0^2$$

$$\Rightarrow v = \frac{d_p}{2} \omega_0^2 \left(\frac{a^2}{b^2} - \frac{r_p^2}{r^2} \right)$$

$$v_0 = \frac{d_p}{2} \omega_0^2 \left(\frac{a^2}{b^2} - \frac{r_p^2}{r^2} \right)$$

$$v_0 = \frac{d_p}{2} \omega_0^2 \left(\frac{a^2}{b^2} - \frac{r_p^2}{r^2} \right)$$

$$\Rightarrow v = -\frac{d_p}{2} \frac{\omega_0^2}{\omega_p^2} \left[\frac{a^2 + b^2 + r_p^2}{a^2 + b^2 - r_p^2} - \frac{a^2 + b^2 + r^2}{a^2 + b^2 - r^2} \right]$$

Ex. 10.13 (See next page.)

Ex. 10.14 The potential boundary conditions at end 1: $v = 0$, and $\frac{dv}{dx} = \frac{d^2 v}{dx^2}$

From Eq. 10.11 and the d_p pathlengths in parts (a) and (b)

$$v = \frac{d_p}{2} \frac{\omega_0^2}{\omega_p^2} \left[\frac{a^2 + b^2 + r_p^2}{a^2 + b^2 - r_p^2} - \frac{a^2 + b^2 + r^2}{a^2 + b^2 - r^2} \right]$$

$$v = \frac{d_p}{2} \frac{\omega_0^2}{\omega_p^2} \left[\frac{a^2 + b^2 + r_p^2}{a^2 + b^2 - r_p^2} - \frac{a^2 + b^2 + r^2}{a^2 + b^2 - r^2} \right]$$

In order to satisfy the $v = 0$ at end 1, we require

$$\frac{d_p}{2} \frac{\omega_0^2}{\omega_p^2} \left[\frac{a^2 + b^2 + r_p^2}{a^2 + b^2 - r_p^2} - \frac{a^2 + b^2 + r^2}{a^2 + b^2 - r^2} \right] = 0$$

Ex-16



1) Q_1 and system of image charges?

In left sphere

Charge system

$$Q_1 \text{ at } d_1 = a$$

$$-Q_1 \frac{a}{2a} \text{ at } d_1'$$

$$Q_2 = \frac{Q_1 a}{2a - a} \text{ at } d_2$$

$$-Q_2 \frac{a}{2a} = -\frac{Q_1 a}{2a - a} \frac{a}{2a} \text{ at } d_2'$$

$$Q_3 = \frac{Q_2 a}{2a - a} = Q_1$$

⋮

$$Q_n = Q_1 \left(\frac{a}{2a - a} \right)^{n-1} \text{ at } d_n \quad Q_n = Q_1 \left(\frac{a}{2a - a} \right)^{n-1} \text{ at } d_n'$$

$$\left(\frac{a}{2a - a} \right)^{\infty} = 1 \quad \left(\frac{a}{2a - a} \right)^{\infty} = 1 \quad d_{\infty} = \frac{a}{2a - a} = 2a$$

$$1) C = \frac{Q_1}{V} = \frac{Q_1}{4\pi a^2} \left[1 + \sum_{n=1}^{\infty} \left(\frac{a}{2a - a} \right)^{2n} \right]$$

Ex-17 1) $V(x, y, z)$ - Required boundary conditions:



① $V(x, 0) = 0$ - Grounded conducting plane at $y=0$.

② $V(x, b) = 0$ - Grounded conducting plane at $y=b$.

③ $V(0, y) = \frac{V_0}{b} y$ - Insulated plane at $x=0$.

④ Degree of freedom boundary defined by: $V(x, y) = \frac{V_0}{b} y$ at $x=0$ and $V(x, y) = 0$ at $x=a$.



2) $V(x, y, z)$ - Required boundary conditions:

① $V(x, y, 0) = 0$ - Grounded plane at $z=0$.

② $V(x, y, b) = 0$ - Grounded plane at $z=b$.

③ $V(0, y, z) = \frac{V_0}{b} z$ - Insulated plane at $x=0$.

④ Degree of freedom boundary defined by: $V(x, y, z) = \frac{V_0}{b} z$ at $x=0$ and $V(x, y, z) = 0$ at $x=a$.

$$C = \frac{Q_1}{V} = \frac{Q_1}{4\pi a^2} \left[1 + \sum_{n=1}^{\infty} \left(\frac{a}{2a - a} \right)^{2n} \right]$$

Ex. 10.11 $E_1 = 100 \text{ eV}$, $v_1 = \frac{1}{2}c$, $E_2 = 100 \text{ eV}$, $v_2 = \frac{1}{2}c$, $\theta = 90^\circ$



$$E_1 = 100 \text{ eV} = \gamma m_0 c^2 = \frac{m_0 c^2}{\sqrt{1 - v_1^2/c^2}} = \frac{m_0 c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$E_2 = 100 \text{ eV} = \gamma m_0 c^2 = \frac{m_0 c^2}{\sqrt{1 - v_2^2/c^2}} = \frac{m_0 c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$E = \gamma M c^2 = \frac{M c^2}{\sqrt{1 - v^2/c^2}} = \frac{M c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$M c^2 = \frac{E}{\gamma} = \frac{133 \text{ eV}}{1.25} = 106.4 \text{ eV}$$

$$M = \frac{106.4 \text{ eV}}{c^2} = \frac{106.4 \times 1.6 \times 10^{-19} \text{ J}}{(3 \times 10^8 \text{ m/s})^2} = 1.18 \times 10^{-36} \text{ kg}$$

$$v = \frac{1}{2}c = 1.5 \times 10^8 \text{ m/s}$$

$$\lambda = \frac{h}{M v} = \frac{6.626 \times 10^{-34} \text{ J s}}{1.18 \times 10^{-36} \text{ kg} \times 1.5 \times 10^8 \text{ m/s}} = 3.75 \times 10^{-10} \text{ m}$$

and simplifying: $v_1' = \frac{1}{2}c \left[\frac{1 + \frac{1}{2}}{1 + \frac{1}{2}} \right] = \frac{1}{2}c$

$$v_2' = \frac{1}{2}c = \frac{1}{2}c$$

$$d) \text{ Force per unit length } F = \frac{1}{2} \frac{1}{\lambda} \frac{1}{\lambda} = \frac{1}{2} \frac{1}{(3.75 \times 10^{-10} \text{ m})^2} = 3.56 \times 10^{18} \text{ N/m}$$

Ex. 10.12



$$v_1 = \frac{1}{2}c, \quad v_2 = \frac{1}{2}c$$

$$E = \gamma m_0 c^2 = \frac{m_0 c^2}{\sqrt{1 - v^2/c^2}} = \frac{m_0 c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$E_1 = \gamma_1 m_1 c^2 = \frac{m_1 c^2}{\sqrt{1 - v_1^2/c^2}} = \frac{m_1 c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$E_2 = \gamma_2 m_2 c^2 = \frac{m_2 c^2}{\sqrt{1 - v_2^2/c^2}} = \frac{m_2 c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

$$E = \gamma M c^2 = \frac{M c^2}{\sqrt{1 - v^2/c^2}} = \frac{M c^2}{\sqrt{1 - 1/4}} = 133 \text{ eV}$$

Ex. 10.13 (See next page.)

Ex. 10.14 Required boundary conditions at $x=0$: $\psi = 0$, and $\frac{d\psi}{dx} = \frac{2m_1 V_0}{\hbar^2} \psi$

From Ex. 10.13 and the boundary conditions, we have

$$\psi = \frac{1}{2} \left[\frac{1}{\cos(k_1 x)} - \frac{1}{\cos(k_2 x)} \right]$$

$$\psi = \frac{1}{2} \left[\frac{1}{\cos(k_1 x)} - \frac{1}{\cos(k_2 x)} \right]$$

In order to satisfy the $\psi = 0$ at $x=0$, we require

$$\frac{1}{\cos(k_1 x)} = \frac{1}{\cos(k_2 x)} \text{ and } k_1 + k_2 = 2k_2 \text{ or } k_1 = k_2 = \frac{2m_1 V_0}{\hbar^2}$$

Ex. 10



10) Q_1 and system of image charges:

<u>In left sphere</u>	<u>In right sphere</u>
Q_1 at $a/2$	$-Q_1 - Q_2$ at a
$Q_2 = \frac{2a}{2a-a} Q_1 = 2Q_1$	$-Q_2 = \frac{2a}{2a-a} (-Q_1 - Q_2) = 2Q_1 + 2Q_2$
$Q_3 = \frac{2a}{2a-a} 2Q_1 = 4Q_1$	\vdots
$Q_4 = 4Q_1 \frac{2a}{2a-a} = 8Q_1$ at a	$-Q_4 = -4Q_1 \frac{2a}{2a-a} = -8Q_1$ at a
\vdots	\vdots
$Q_n = \frac{2a}{2a-a} Q_{n-1}$	$-Q_n = -\frac{2a}{2a-a} Q_{n-1}$
$Q_n = \frac{2a}{2a-a} Q_{n-1}$	$Q_n = \frac{2a}{2a-a} Q_{n-1}$

$$a) \phi = \frac{Q_1}{4\pi\epsilon_0} \sum_{n=0}^{\infty} \left[r + \sum_{n=1}^{\infty} \left(\frac{r}{a} \right)^{2n} \right]$$

Ex. 11 a) $\nabla^2 \phi = 0$, Required boundary conditions:



1) $\phi(a) = 0$ — grounded conducting plane at $r=a$

2) $\phi(0) = 0$ — grounded conducting plane at $r=0$

3) $\phi(a) = \phi_0$ — grounded plane at $r=a$ with potential ϕ_0

4) $\phi(0) = \phi_0$ — grounded plane at $r=0$ with potential ϕ_0

5) $\phi(0) = \phi_0$ — grounded plane at $r=0$ with potential ϕ_0

6) $\phi(a) = \phi_0$ — grounded plane at $r=a$ with potential ϕ_0

7) $\phi(0) = \phi_0$ — grounded plane at $r=0$ with potential ϕ_0

8) $\phi(a) = \phi_0$ — grounded plane at $r=a$ with potential ϕ_0

9) $\phi(0) = \phi_0$ — grounded plane at $r=0$ with potential ϕ_0

10) $\phi(a) = \phi_0$ — grounded plane at $r=a$ with potential ϕ_0

11) $\phi(0) = \phi_0$ — grounded plane at $r=0$ with potential ϕ_0

12) $\phi(a) = \phi_0$ — grounded plane at $r=a$ with potential ϕ_0

Ex-12 $V(x,y) = C_1 \sinh \frac{x^2}{2} \cos \frac{y^2}{2}$

Ex-13 $V(x,y) = C_1 \sinh \frac{x^2}{2} \sin \frac{y^2}{2}$

Ex-14 $V = \sum_{n=1}^{\infty} C_n \cos nx = \sum_{n=1}^{\infty} C_n \sinh \frac{x^2}{2} \sin \frac{y^2}{2}$

$V(x,y) = \sum_{n=1}^{\infty} \frac{\sinh(\frac{x^2}{2}) \sin(\frac{y^2}{2})}{\sinh(\frac{x^2}{2}) \sin(\frac{y^2}{2})} \sin \frac{y^2}{2}$

Ex-15 $V(x,y) = \sum \sin \frac{y^2}{2} = [C_1 \sinh \frac{x^2}{2} + C_2 \cosh \frac{x^2}{2}]$

At $y=0$, $V(x,0) = 0 = \sum C_1 \sinh \frac{x^2}{2} \rightarrow C_1 = \begin{cases} 0, & \text{if } x=0 \\ \text{arbitrary}, & \text{if } x \neq 0 \end{cases}$

At $y=\pi$, $V(x,\pi) = 0 = \sum \sin \frac{y^2}{2} [C_1 \sinh \frac{x^2}{2} + C_2 \cosh \frac{x^2}{2}]$
 $\rightarrow C_1 \sinh \frac{x^2}{2} + C_2 \cosh \frac{x^2}{2} = \begin{cases} 0, & \text{if } x=0 \\ \text{arbitrary}, & \text{if } x \neq 0 \end{cases}$
 $\therefore C_2 = \begin{cases} \frac{-\sinh(\frac{x^2}{2})}{\cosh(\frac{x^2}{2})} (C_1 + \sin \frac{y^2}{2}), & \text{if } x \neq 0 \\ 0, & \text{if } x=0 \end{cases}$

Ex-16 $V(x,y) = \sum \sum C_{nm} \sin \frac{nx}{2} \sin \frac{ny}{2} \sinh k_{nm}$

where $k_{nm} = \sqrt{\frac{n^2 x^2}{4} + \frac{n^2 y^2}{4}}$

At $x=0$, $V(0,y) = 0 = \sum \sum C_{nm} \sin \frac{nx}{2} \sin \frac{ny}{2} \sinh k_{nm}$
 $\rightarrow C_{nm} = \begin{cases} \frac{\sin \frac{ny}{2}}{\sinh k_{nm}} & \text{if } x=0 \\ 0, & \text{if } x \neq 0 \end{cases}$

Ex-17 Solution: $V(x,y) = A_1 + A_2$

a) $A_1 = 0 \rightarrow A_1 = 0$
 $A_2 = 0 \rightarrow V(x,y) = A_2 \rightarrow A_2 = \frac{1}{2} \left. \begin{array}{l} \therefore V(x,y) = \frac{1}{2} A_2 \\ \text{if } x=0 \end{array} \right\}$

b) $A_1 = 0 \rightarrow V(x,y) = A_1 + A_2$
 $A_2 = 0 \rightarrow V(x,y) = 0 = A_1 + A_2 \rightarrow A_1 = -\frac{1}{2} A_2, A_2 = \frac{1}{2} A_2$

$\therefore V(x,y) = \frac{1}{2} A_2 (2x - 1)$ $\therefore V(x,y) = \dots$

Chapter 1

Steady Electric Currents

Ex. 1 a) Integrating $\epsilon_r(r) = \epsilon_0 \left(\frac{R}{r} \right)^2$ $\Rightarrow \epsilon_r = \frac{Q}{4\pi r^2}$

$$\epsilon_r = -\frac{1}{r} \frac{d\phi}{dr} = -\frac{1}{r} \frac{d\phi}{dr} \Rightarrow \frac{d\phi}{dr} = -\frac{Q}{4\pi r^2}$$

b) $\int \epsilon_r \frac{d\phi}{dr} dr = -\frac{Q}{4\pi} \int \frac{1}{r^2} dr$

$$\phi = \int_0^r \epsilon_r \frac{d\phi}{dr} dr = -\frac{Q}{4\pi \epsilon_0} \int_0^r r^{-2} dr = \frac{Q}{4\pi \epsilon_0} \frac{1}{r}$$

c) On the inside, $r < a$, $E = -\nabla \phi = \frac{Q}{4\pi \epsilon_0 r^2}$

Total surface charge on inside $= Q$ $\Rightarrow \frac{Q}{4\pi a^2} \cdot 4\pi a^2 = Q$

Total charge on outside $= 0$.

d) Substituting ϕ in Eq. (1) (ii)

$$u = \frac{\epsilon_r}{2} E^2 = \left(\frac{\epsilon_r}{2} \right) \left(\frac{Q}{4\pi r^2} \right)^2 = \frac{Q^2}{32\pi^2 \epsilon_0 r^4}$$

Integrating $u = \left(\frac{Q^2}{32\pi^2 \epsilon_0} \right) \frac{1}{r^4}$

\therefore Energy flow $T_r = -\frac{1}{4\pi} \left(\frac{Q^2}{16\pi \epsilon_0} \right) \frac{1}{r^4}$

For $\frac{1}{2} \text{ cm} < r < 1 \text{ cm}$, $u = \frac{Q^2}{32\pi^2 \epsilon_0 r^4}$

and $u = 0$ for $r > 1 \text{ cm}$, $T_r = 0$ for $r > 1 \text{ cm}$.

Ex. 2 $R_1 =$ Resistance per unit length of wire $= \frac{\rho}{A_1} = \frac{\rho}{\pi r_1^2}$

$R_2 =$ Resistance per unit length of coating $= \frac{\rho}{A_2} = \frac{\rho}{\pi(r_2^2 - r_1^2)}$

Let $t =$ thickness of coating, $\Rightarrow r_2 = r_1 + t$ $\Rightarrow r_2^2 - r_1^2 = 2r_1 t$

a) $R_1 + R_2 = \frac{\rho}{\pi} \left(\frac{1}{r_1^2} + \frac{1}{2r_1 t} \right)$

b) $R_1 = R_2 = \frac{\rho}{\pi} \left(\frac{1}{r_1^2} + \frac{1}{2r_1 t} \right) = \frac{\rho}{\pi} \left(\frac{1}{r_1^2} + \frac{1}{2r_1 t} \right)$

$$R_1 = \frac{\rho}{\pi} = \frac{\rho}{\pi r_1^2} \Rightarrow R_1 = \frac{\rho}{\pi r_1^2}$$

Thus, $r_1 = 10 \text{ cm}$ and $R = R_1$.

Ex-10) At E_1 (interface) $E_{inc} + E_{ref} \rightarrow E_{trn} = E_1 \sin \alpha_1 = E_2 \sin \alpha_2$
 $E_1 \cos \alpha_1 = E_2 \cos \alpha_2 \rightarrow E_1 \cos \alpha_1 = E_2 \cos \alpha_2 \rightarrow E_1 \cos \alpha_1 = E_2 \cos \alpha_2$
 $\therefore E_2 = E_1 \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2}} = \left(\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \right)^{1/2} E_1$ (1)

$E_{trn} = \frac{\mu_2 \epsilon_2}{\mu_1 \epsilon_1} E_{inc} \rightarrow \alpha_2 = \sin^{-1} \left(\frac{\mu_2 \epsilon_2}{\mu_1 \epsilon_1} \sin \alpha_1 \right)$ (2)

b) At E_2 (interface) $E_{trn} - E_{ref} = E_2 \rightarrow E_2 \cos \alpha_2 - E_2 \cos \alpha_2 = E_2$

$E_2 = \left(\frac{\mu_2 \epsilon_2}{\mu_1 \epsilon_1} \right) E_{inc} = \left(\frac{\mu_2 \epsilon_2}{\mu_1 \epsilon_1} \right)^{1/2} E_1 \cos \alpha_1$

- c) If both media are perfect dielectrics, $\mu_1 = \mu_2 = \mu_0$
 (1) and (2) reduce to $E_2 = E_1 \cos \alpha_2$ and $\alpha_2 = \alpha_1$ respectively
 and $E_2 = E_1$

Ex-11



$C_{total} = C_1 + C_2 = \epsilon_0 \epsilon_1 \frac{A_1}{d} + \epsilon_0 \frac{A_2}{d}$

At the leading edge of the slab, the electric field is constant vertically.

$E = -\epsilon_1 E_1 = E = -\frac{V}{d} = -\epsilon_1 \frac{V}{d}$

$Q_1 = \int E \cdot dA_1 = \int \frac{V}{d} \epsilon_1 dA_1 = \frac{V}{d} \epsilon_1 A_1 = \frac{V}{d} \epsilon_1 A_1 \frac{l}{L}$

$Q_2 = \frac{V}{d} = \frac{V}{d} = \frac{V}{d} \frac{A_2}{L} = \frac{V}{d} \frac{A_2}{L}$

b) $C_1 = \frac{Q_1}{V} = \frac{\epsilon_1 \epsilon_0 A_1 l}{dL} = \frac{\epsilon_1 \epsilon_0 A_1 l}{dL}$ as upper plate.

$C_2 = \frac{Q_2}{V} = \frac{\epsilon_0 A_2}{dL} = \frac{\epsilon_0 A_2}{dL}$ as lower plate.

c) $C = \frac{Q}{V} = \frac{Q_1 + Q_2}{V} = \frac{\epsilon_1 \epsilon_0 A_1 l}{dL} + \frac{\epsilon_0 A_2}{dL} = \frac{\epsilon_0 A_1}{dL} \left[\epsilon_1 \frac{l}{L} + \frac{L-l}{L} \right] = \frac{\epsilon_0 A_1}{dL} \left[\frac{\epsilon_1 l + L - l}{L} \right]$

Ex-12



a) $V_1 = \frac{V R_2}{R_1 + R_2}$, $V_2 = \frac{V R_1}{R_1 + R_2}$

$V_3 = \frac{V R_4}{R_3 + R_4}$, $V_4 = \frac{V R_3}{R_3 + R_4}$

b) $P = V_1 I_1 = V_2 I_2 = V_3 I_3 = V_4 I_4$

$= V^2 \frac{R_1 R_2}{R_1^2 + R_2^2}$

Ex-13 Refer to Fig. 1-6. In the transient state, the equation of continuity may be written at the interface.

$\frac{\partial \rho_{int}}{\partial t} + J_1 = J_2 = J_3 = J_4$ (1)

Now

$J_1 = J_2 = J_3 = J_4$ (2)

$V_1 = V_2 = V_3 = V_4$ (3)

Solving ① and ② for x_1 and x_2 in terms of α and β :

$$x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \quad \text{③} \quad x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \quad \text{④}$$

(ii) Substituting ③ and ④ in ②:

$$-\frac{\beta}{\alpha} = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} x_1 + \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} x_2 \quad \text{⑤}$$

Solution of ⑤:

$$x_2 = \left(\frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \right) \alpha [1 - \alpha^{-2\beta}] \quad \text{⑥}$$

$$\text{where } T = \text{Substitution term} = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \quad \text{⑦}$$

(iii) Using ③ and ④:

$$x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} (1 - \alpha^{2\beta}) = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} (1 - \alpha^{2\beta}) = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$\text{Thus (i) } x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \quad x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

$$x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta} = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta} = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$(ii) x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$x_3 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

$$\text{Thus (ii) } x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha} \alpha^{2\beta}$$

Solution: (iii) $x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha}$

Substituting equations: (iii) $x_1 = \frac{\beta(1-\alpha) + \alpha}{\beta(1-\alpha) + \alpha}$

$$x_2 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

$$x_3 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

$$x_4 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

$$x_5 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

$$x_6 = \frac{\beta(1-\alpha) - \alpha}{\beta(1-\alpha) + \alpha}$$

Ex-15 Assume a potential difference V_0 between the inner and outer spheres.

$$\oint \vec{E} \cdot d\vec{s} = Q = \int_0^R \frac{\rho(r)}{\epsilon_0} 4\pi r^2 dr \quad \rightarrow \quad V = \frac{Q}{C} = \frac{Q}{4\pi\epsilon_0} \left(\frac{1}{R_1} - \frac{1}{R_2} \right)$$

$$V_0 = -\int_{R_1}^{R_2} E_r dr = -\int_{R_1}^{R_2} \frac{Q}{4\pi\epsilon_0 r^2} dr = -\frac{Q}{4\pi\epsilon_0} \left(\frac{1}{R_2} - \frac{1}{R_1} \right)$$

$$\rightarrow C = \frac{Q}{V_0} = \frac{4\pi\epsilon_0 R_1 R_2}{R_1 - R_2} \quad \quad U = \frac{1}{2} Q V_0 = \frac{Q^2}{2C} = \frac{Q^2}{8\pi\epsilon_0} \left(\frac{1}{R_1} - \frac{1}{R_2} \right)$$

$$E = \int_0^R \frac{\rho(r)}{\epsilon_0} r' dr = \frac{\rho(r) R^2}{2\epsilon_0}$$

$$E = \frac{\rho(r)}{\epsilon_0} = \frac{1}{2\epsilon_0 R^2} \left(\frac{1}{R_1} - \frac{1}{R_2} \right) \quad \text{which matches against boundary Eqs. (1) and (2) above.}$$

Ex-16 Assume a current I between the spherical conductors.

$$\vec{J} = \epsilon_0 \frac{dE}{dt} = r E.$$

$$I = -\int_0^R E_r dr = -\int_0^R \frac{d}{dt} \left(\frac{Q}{4\pi\epsilon_0 r^2} \right) dr = \frac{dQ}{dt} \int_0^R \frac{dr}{2\pi\epsilon_0 r^3}$$

$$= \frac{dQ}{dt} \int_0^R \frac{1}{2} \left(\frac{1}{R_1} - \frac{1}{R_2} \right) dr = \frac{dQ}{dt} \ln \left(\frac{R_2 R_1}{R_1 R_2} \right)$$

$$I = \frac{dQ}{dt} = \frac{1}{2\pi\epsilon_0} \ln \left(\frac{R_2 R_1}{R_1 R_2} \right)$$

Ex-17 Assume $E_r = \frac{A}{r^2} = \epsilon_0 \frac{dV}{dr}$.

$$E(r) = \int_0^R \frac{\rho(r')}{\epsilon_0} r' dr = \frac{1}{2\pi\epsilon_0} \int_0^R \rho(r') r' dr = \frac{1}{2\pi\epsilon_0} \int_0^R \rho(r') r' dr.$$

$$E(r) = \frac{1}{2\pi\epsilon_0} \int_0^R \rho(r') r' dr = \frac{1}{2\pi\epsilon_0} \int_0^R \rho(r') r' dr$$

$$V = -\int_0^R E_r dr = -\frac{1}{2\pi\epsilon_0} \int_0^R \frac{A}{r^2} dr = \frac{A}{2\pi\epsilon_0} \left(\frac{1}{R} - \frac{1}{R_1} \right)$$

$$C = \frac{Q}{V} = \frac{4\pi\epsilon_0 R_1 R_2}{R_1 - R_2} \left(\frac{1}{R_1} - \frac{1}{R_2} \right)$$

Ex-18 $\vec{E} \cdot \vec{J} = 0 = \vec{E} \cdot \rho \vec{r} = \rho \vec{E} \cdot \vec{r} = (\vec{E} \cdot \vec{r}) \rho = 0$.

$$E = \frac{\rho(r)}{\epsilon_0} \quad \vec{E} \cdot \vec{r} = \frac{\rho(r)}{\epsilon_0} (r^2) \quad \text{or } \rho = \epsilon_0 \frac{dE}{dr} = \epsilon_0 \frac{d}{dr} \left(\frac{A}{r^2} \right)$$

Substituting back: $R = \frac{d^2}{4} - a^2 \rightarrow L = \frac{d}{2} \sqrt{\frac{d^2}{4} - a^2}$
 $V = \int_0^L \rho \cdot dV = \rho \cdot \frac{d}{2} \int_0^L \sqrt{\frac{d^2}{4} - a^2} dx \rightarrow C = \frac{\rho d^2}{2\sqrt{\frac{d^2}{4} - a^2}} L = \frac{\rho d^2 \sqrt{\frac{d^2}{4} - a^2}}{2}$
 $E = \int_0^L \rho \cdot dV = \int_0^L \rho \cdot dV = \int_0^L \left(\frac{d^2}{4} - a^2\right) \left[\frac{\rho d^2}{2\sqrt{\frac{d^2}{4} - a^2}}\right] dx = \frac{\rho d^2 \sqrt{\frac{d^2}{4} - a^2}}{2}$
 $E = \frac{C^2}{2\rho} = \frac{d^2 C^2}{4\rho d^2}$

Ex. 11 A convex lens of ρ and of L to construct at the center of spheres R and r respectively. d and d_1



$$R = \frac{d}{2} \left(\frac{1}{d} - \frac{1}{d_1} \right)$$

$$r = \frac{d}{2} \left(\frac{1}{d_1} - \frac{1}{d} \right)$$

$$C = \frac{d}{2} \rho = \frac{d \rho}{2} \frac{d_1}{d_1} = \frac{d \rho d_1}{2(d_1 - d)}$$

$$= \frac{C^2}{2\rho} = \frac{d^2 C^2}{4\rho d^2}$$

$$E = \frac{1}{2\rho} \left(\frac{1}{d} + \frac{1}{d} - \frac{1}{d_1} - \frac{1}{d_1} \right) = \frac{1}{2\rho} \left(\frac{1}{d} + \frac{1}{d} - \frac{2}{d_1} \right)$$

Ex. 12



The curved line passes of the lens in (a) and (b) has the same area and is longer than that with the same radius is exactly the same as that of R_1 in (a). All boundary conditions are satisfied.

$$E = \frac{C^2}{2\rho} = \frac{C^2}{2\rho}$$

We can write $E = \frac{C^2}{2\rho}$

where ρ and C are respectively potential and energy related. The structures are similar to the E -lines of a cavity with one of its lengths, both carrying a charge q in the electrical case.

Soln According to problem 12-19, the current flow pattern would be the same as that of a whole sphere in an unbounded space condition. Hence the current flow would be radial. Assume a current I .

$$J = \epsilon_0 \frac{\partial E}{\partial t} \quad E = \epsilon_0 \frac{\partial V}{\partial r}$$

$$\nabla \cdot J = -\int \text{div} = -\text{div} \int \frac{\partial V}{\partial r} = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 J)$$

$$0 = \frac{1}{r^2} = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 J) = \frac{1}{r^2} \frac{\partial}{\partial r} (r^2 I) \quad (1)$$

Soln Specified boundary conditions can be satisfied by both cases of Laplace's equation with zero separation constants: $\epsilon_2 = 0$ or $\epsilon_3 = 0$. (Take $\epsilon_2 = 0$, $\nabla^2 V = 0$, $V(r) = A_2 r^2 + B_2$)

$$\Rightarrow \text{At } r = a, \quad V(a) = \frac{1}{2} = A_2 a^2 + B_2 \Rightarrow A_2 a^2 = \frac{1}{2} - B_2$$

$$\therefore V = \frac{1}{2} a^2$$

$$\text{At } E = -\nabla V = -\epsilon_0 \frac{\partial V}{\partial r} \Rightarrow J = \text{div} \int \epsilon_0 \frac{\partial V}{\partial r}$$

Soln $V(r) = \frac{1}{2} (a_2 r^2 + a_3 r^{-2}) + a_4 \cos \theta + a_5 \sin \theta$

S.C.1: $V(a, \theta) = V(b, \theta) = 0 \Rightarrow a_2 = 0$

$r \rightarrow \infty, \quad V = \frac{1}{2} \cos \theta \Rightarrow a_3 = a_4 = 0, \quad a_5 = \frac{1}{2}$

When $V(r, \theta) = (a_2 r^2 + \frac{1}{2}) \cos \theta, \quad E = -\epsilon_0 \frac{\partial V}{\partial r} = -\epsilon_0 a_2 r \cos \theta$

S.C.2: $\frac{\partial V}{\partial r} \Big|_{r=a} = 0 \Rightarrow a_2 = \frac{1}{2a^2} = 0, \quad a_3 = a^2 a_4 = -\frac{1}{2} a^2$

$\therefore V(r, \theta) = -\frac{1}{2} (a - \frac{1}{2}) \cos \theta$

$$J = -\text{div} \int \epsilon_0 \frac{\partial V}{\partial r} = \epsilon_0 \frac{\partial V}{\partial r}$$

$$= \epsilon_0 \frac{\partial}{\partial r} \left(-\frac{1}{2} \right) \cos \theta = \epsilon_0 \frac{\partial}{\partial r} \left(a - \frac{1}{2} \right) \cos \theta$$

$$= \epsilon_0 \left(\frac{\partial}{\partial r} \cos \theta + \frac{\partial}{\partial r} \sin \theta \right) \frac{\partial}{\partial r} \left(a - \frac{1}{2} \right) \cos \theta + \epsilon_0 \cos \theta$$

$$= \epsilon_0 \frac{\partial}{\partial r} = \frac{\partial}{\partial r} \left(\epsilon_0 \cos \theta + \epsilon_0 \sin \theta \right), \quad r = a;$$

$$J = 0, \quad \text{radial}$$

Chapter 6

Static Magnetic Fields

Ex 1



$$\frac{\partial B_x}{\partial x} = \frac{\partial B_z}{\partial z} \quad \text{or} \quad -\mu_0 j_y = 0 \quad \text{--- (1)}$$

$$\frac{\partial B_y}{\partial y} = -\frac{\partial B_z}{\partial z} \quad \text{or} \quad \mu_0 j_x = 0 \quad \text{--- (2)}$$

Combining (1) and (2)

$$\frac{\partial B_z}{\partial z} = \mu_0 j_x = 0$$

--- B_z not a function of z

At $z=0$, $B_z = 0$ due to symmetry

Substituting B_z in (1): $B_x = -\mu_0 j_y z$ At $z=0$, $B_x = 0$ due to symmetry

∴ $B_x = -\mu_0 j_y z$ --- $y = \frac{B_x}{-\mu_0 j_y z}$ --- $y = 0$ at $z=0$

$B_y = -\mu_0 j_x z$ --- $z = \frac{B_y}{-\mu_0 j_x}$ --- $z = 0$ at $B_y = 0$

From (1) and (2): $B^2 = B_x^2 + B_y^2 = \left(\frac{B_x}{-\mu_0 j_y}\right)^2 + \left(\frac{B_y}{-\mu_0 j_x}\right)^2$

Ex 2 $\frac{\partial B_x}{\partial x} = -\frac{\partial}{\partial z}(B_y + \mu_0 j_z)$

(1) $B_x = \mu_0 j_z z$, $B_y = \mu_0 j_x z$

$\frac{\partial B_x}{\partial x} = 0$

$\frac{\partial B_x}{\partial x} = -\mu_0 j_z z$ --- $\begin{cases} \mu_0 j_z = 0 \\ \mu_0 j_z = \left(\frac{\partial B_x}{\partial x} - \mu_0 j_z z\right) = \frac{\partial B_x}{\partial x} \end{cases}$

$\frac{\partial B_x}{\partial x} = -\mu_0 (j_z - j_x z)$ --- $\mu_0 j_z = \left(\frac{\partial B_x}{\partial x} - \mu_0 j_x z\right)$; $\mu_0 j_z = \frac{\partial B_x}{\partial x}$

If the electric field is neglected, the only source of current is $j = 0$, $j = \frac{\partial B_x}{\partial x} z + \frac{\partial B_y}{\partial y} z$, $j = -\mu_0 (j_z - j_x z) z$ --- $\frac{\partial B_x}{\partial x} = -\mu_0 j_z$

∴ $B_x = \mu_0 j_z z$, $B_y = \mu_0 j_x z$, $B_z = 0$

∴ $B^2 = B_x^2 + B_y^2 = (\mu_0 j_z z)^2 + (\mu_0 j_x z)^2$

∴ $\frac{\partial B^2}{\partial z} = 2\mu_0 j_z z + 2\mu_0 j_x z$

∴ $B^2 = \mu_0^2 (j_x^2 + j_z^2) z^2$

$$c) \vec{E} = -\nabla \phi, \quad \vec{E} = -\frac{\partial \phi}{\partial x} \hat{x} - \frac{\partial \phi}{\partial y} \hat{y} - \frac{\partial \phi}{\partial z} \hat{z}$$

$$\left. \begin{aligned} \frac{\partial \phi}{\partial x} &= -\frac{\partial}{\partial x} \left(\frac{1}{4\pi\epsilon_0} \frac{q}{r} \right) = \frac{1}{4\pi\epsilon_0} \frac{qx}{r^3} \\ \frac{\partial \phi}{\partial y} &= -\frac{\partial}{\partial y} \left(\frac{1}{4\pi\epsilon_0} \frac{q}{r} \right) = \frac{1}{4\pi\epsilon_0} \frac{qy}{r^3} \\ \frac{\partial \phi}{\partial z} &= -\frac{\partial}{\partial z} \left(\frac{1}{4\pi\epsilon_0} \frac{q}{r} \right) = \frac{1}{4\pi\epsilon_0} \frac{qz}{r^3} \end{aligned} \right\} \begin{array}{l} \text{Component} \\ \text{method} \\ \text{method} \end{array} \left. \begin{array}{l} \text{Electric field vector} \\ \text{method} \\ \text{method} \end{array} \right\} \text{Electric field vector} \text{ method}$$

Ex-4 Application of Ampere's circuital law

$$\text{wire, } \vec{I} = I_0 \frac{\vec{r}}{r}$$

$$\text{wire, } \vec{I} = I_0 \frac{\vec{r}}{r}$$

$$\text{wire, } \vec{I} = I_0 \frac{\vec{r}}{r}$$

Ex-4



Using Ampere's law

$$\oint \vec{B} \cdot d\vec{l} = \mu_0 I_{enc}$$

$$B \cdot l = \mu_0 I_0$$

$$B = \frac{\mu_0 I_0}{l}$$

$$\vec{B} = \frac{\mu_0 I_0}{l} \hat{z}$$

$$\text{where } B = \frac{\mu_0 I_0}{2\pi r} \int_0^{2\pi} \cos^2 \theta d\theta = \frac{\mu_0 I_0}{2\pi r} \int_0^{2\pi} \frac{1 + \cos 2\theta}{2} d\theta$$

$$B = \frac{\mu_0 I_0}{2\pi r} \left[\theta + \frac{\sin 2\theta}{2} \right]_0^{2\pi} = \frac{\mu_0 I_0}{2\pi r} (2\pi) = \frac{\mu_0 I_0}{r}$$

To find B at $P(x, y, z)$, we add vertically the contributions of the current along the right and the left of point P using the result in part (a)

$$\vec{B}_P = \vec{B}_1 + \vec{B}_2$$

$$\vec{B}_1 = \frac{\mu_0 I_0}{2\pi r} \left[\cos^2 \theta + \cos^2 \theta \right]$$

$$\vec{B}_2 = \frac{\mu_0 I_0}{2\pi r} \left[\cos^2 \theta + \cos^2 \theta \right]$$

$$\therefore \vec{B}_P = \frac{\mu_0 I_0}{2\pi r} \left[\cos^2 \theta + \cos^2 \theta \right]$$

Soln:



We find that \vec{B}_1 at P , due to wire, will be directed as wire carrying a current I and making an angle α with the other end as shown.

$$d\vec{B}_1 = \frac{\mu_0 I}{4\pi r^2} dl \times \vec{r} \quad \text{Dir's rule, dipole-like}$$

$$= \frac{\mu_0 I}{4\pi r^2} (dl \sin \alpha) \quad \vec{B} \perp \vec{r} \text{ and } \vec{I}$$

$$B_1 \sin \alpha = B_2 \sin \alpha$$

$$\vec{B}_1 = -B_2 \frac{dl \sin \alpha}{r^2} \hat{r} \text{ (into page)}$$

$$= -B_2 \frac{dl \sin^2 \alpha}{r^2} \text{ (into)}$$

Applying the above result to the lower slanted leg of wire, we have

$$\vec{B}_2 = B_2 \frac{\mu_0 I}{4\pi} \left(\int_0^l \frac{\sin^2 \alpha}{r^2} dl + \int_0^h \frac{\sin^2 \alpha}{r^2} dl + \int_0^l \frac{\sin^2 \alpha}{r^2} dl + \int_0^h \frac{\sin^2 \alpha}{r^2} dl \right)$$

For this problem, we'll do the upper

$$\vec{B}_2 = \frac{\mu_0 I}{4\pi} \left(\frac{2}{l} \sin^2 \alpha + \frac{2}{h} \sin^2 \alpha \right)$$

$$B_2 = B_1 \frac{2}{\sin^2 \alpha} \left(\frac{l}{h} + 1 \right)$$

Soln: The problem can be decomposed into two sub-problems for carrying a current of interest I .

1. A cylindrical tube carrying a uniformly distributed longitudinal surface current K (A/m).

$$\vec{K} = \begin{cases} K \hat{\phi} & \text{inside} \\ 0 & \text{outside} \end{cases}$$

2. A solenoid with a turns per unit length carrying a current I (A).

$$\vec{K} = \begin{cases} K_1 \hat{\phi} & \text{inside} \\ 0 & \text{else} \end{cases}$$

$$\text{Total } \vec{K} = \vec{K}_1 + \vec{K}_2$$

Ex. 1 From Example 4-6, Eq. (10):



Direction of \vec{H} is determined by the right-hand rule.

$$d\vec{H} = \frac{\mu_0 I dy}{2\pi r^2} \left(\frac{y}{r}\right) \hat{z}$$

$$\begin{aligned} \vec{H} &= \frac{\mu_0 I}{2\pi} \int_{-h/2}^{h/2} \frac{y}{(y^2 + z^2)^{3/2}} dy \\ &= \frac{\mu_0 I}{2\pi} \left[\frac{y}{z\sqrt{y^2 + z^2}} + \frac{1}{z^2} \ln \left| \frac{y + \sqrt{y^2 + z^2}}{z} \right| \right] \end{aligned}$$

$$\rightarrow \mu_0 \left(\frac{I}{2}\right) \hat{z} \text{ at } z=0.$$

$$\vec{H} = H_x \hat{x}$$

At $z=0$, the magnetic flux density due to an infinitely long strip of width dy is

$$d\vec{H} = \frac{\mu_0 I dy}{2\pi r^2} \left(-\frac{z}{r} \hat{x} + \frac{y}{r} \hat{y}\right) \\ r = \sqrt{y^2 + z^2}$$

$$\therefore \vec{H} = \int d\vec{H} = H_x \hat{x} + H_y \hat{y},$$

where

$$\begin{aligned} H_x &= -\frac{\mu_0 I}{2\pi} \int_{-h/2}^{h/2} \frac{z}{(y^2 + z^2)^{3/2}} dy \\ &= -\frac{\mu_0 I}{2\pi} \left[\frac{z}{y\sqrt{y^2 + z^2}} - \frac{z}{y^2} \ln \left| \frac{y + \sqrt{y^2 + z^2}}{z} \right| \right] \end{aligned}$$

$$\begin{aligned} H_y &= \frac{\mu_0 I}{2\pi} \int_{-h/2}^{h/2} \frac{y}{(y^2 + z^2)^{3/2}} dy \\ &= \frac{\mu_0 I}{2\pi} \ln \left| \frac{y + \sqrt{y^2 + z^2}}{z} \right| \end{aligned}$$



Top view

Ex. 2 This problem is a superposition of two problems:

$$\vec{H} = \vec{H}_1 + \vec{H}_2,$$

where

\vec{H}_1 is the magnetic flux density of \vec{I} due to the

Simultaneously wires carrying equal and opposite currents
 Attractive \vec{B}_2 points out of paper:

$$\vec{B}_1 = \mu_0 \frac{I_1}{2\pi r_1}$$

2. \vec{B}_2 is the magnetic flux density at P due to a finite
 wire. Taking one half of the wire as length $2a$ the
 field:

$$\vec{B}_2 = \mu_0 \frac{I_2}{4a^2}$$

$$\therefore \vec{B} = \mu_0 \frac{I_2}{4a^2} \left(\frac{1}{r_1} + \frac{1}{r_2} \right)$$

Ex 21 Use Eq (19-21) $\vec{B} = \frac{\mu_0 I}{4\pi} \left[\frac{\vec{r}_2 - \vec{r}_1}{(r_1^2 - a^2)^{3/2}} - \frac{\vec{r}_1 - \vec{r}_2}{(r_2^2 - a^2)^{3/2}} \right]$



$$\text{At } P, \vec{B}_1 = \frac{\mu_0 I}{4\pi} \frac{2a}{(r^2 - a^2)^{3/2}}$$

$$\text{As } \frac{d\vec{B}}{dt} = \frac{d\vec{E} \times \vec{r}}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt}$$

At the midpoint, $a = \frac{1}{2} \sqrt{2} r$

$$\text{As } \frac{d\vec{B}}{dt} = \frac{d\vec{E} \times \vec{r}}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt}$$

$$\text{At } P, \frac{d\vec{B}}{dt} = \frac{d\vec{E} \times \vec{r}}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt} = \frac{d(\vec{E} \times \vec{r})}{dt}$$

Ex 22 Use Eq (19-21) for a wire of length $2a$.

$$\vec{B} = \mu_0 \frac{I}{4\pi} \frac{2a}{r^2}$$



In this problem, $a = \frac{1}{2} \sqrt{2} r$ and $\theta = \frac{\pi}{4}$

$$\vec{B} = \mu_0 \frac{I}{4\pi} \left(\frac{2a}{r^2} \right) = \mu_0 \frac{I}{4\pi} \frac{2a}{r^2} \sin \frac{\pi}{4}$$

When θ is very large, $\cos \frac{\theta}{2} \approx \frac{\theta}{2}$, $\vec{B} \rightarrow \mu_0 \frac{I}{4\pi} \frac{2a}{r^2}$,
 which is the same as Eq (19-21) with $\theta = \pi$.

Ex-12 $B_0 = \frac{\mu_0 I}{2a}$; $B = \int B_0 dr = \frac{\mu_0 I}{2a} \int_a^b \frac{r}{r} dr = \frac{\mu_0 I}{2a} \ln \frac{b}{a}$

If B_0 at $r = \frac{a+b}{2}$ is used, $B = \frac{\mu_0 I}{2a} \ln \left(\frac{b}{a} \right)$

% error = $\frac{B - B_0}{B_0} \times 100 = \left[\frac{\ln \frac{b}{a} - \ln \left(\frac{b+a}{2} \right)}{\ln \frac{b}{a}} \right] \times 100$

Ex-13 $B = B_0$; $\oint B \cdot dl = \mu_0 I$



If there is no hole,

$$\oint B_0 \cdot dl = \mu_0 I$$

$$\rightarrow B_0 = \frac{\mu_0 I}{2\pi r} \text{ from } \begin{cases} B_{\theta} = \frac{\mu_0 I}{2\pi r} \\ B_r = 0 \end{cases}$$

For $r > b$ in the hole portion,

$$B_0 = \frac{\mu_0 I}{2\pi r} \text{ from } \begin{cases} B_{\theta} = \frac{\mu_0 I}{2\pi r} \\ B_r = 0 \end{cases}$$

Superposing B_0 and B_{in} and noting that I_{in} and I_{out} are in

opposite directions, we have $B_0 = B_{in} + B_{out} = 0$, and $B_{in} = B_{out} = \frac{\mu_0 I}{2\pi r}$

Ex-14 $B = \mu_0 I \rightarrow B = B_0 r = \mu_0 \left(\frac{I}{2a} - \frac{I}{2b} \right) = \mu_0 \frac{I}{2a}$

For $r < a$, $B_0 = \mu_0 I$ gives $B = \mu_0 \frac{I}{2a} r$

For $a < r < b$, $B_0 = \mu_0 I$ gives $B = \mu_0 \frac{I}{2a} r$

Integrating, $B_z = \mu_0 \left[\frac{I}{2a} \left(\frac{r^2}{2} + c_1 \right) \right]$, $0 < r < a$,

$$B_z = \mu_0 \left[\frac{I}{2a} (ar + c_2) \right]$$
, $a < r < b$

At $r = b$, $B_z = 0 \rightarrow c_2 = -\frac{I}{2a} (ab + b^2)$

$$\therefore B_z = \mu_0 \left[-\frac{I}{2a} (ab + b^2) + c_1 \right]$$
, $r > b$

Ex-15 $B_0 = \mu_0 I$ for one wire; $B = \mu_0 \frac{I}{2a}$ is $\frac{\mu_0 I}{2a} \ln \frac{b}{a}$

For two wires carrying equal and opposite currents,

a) $B = \mu_0 \frac{I}{2a} \ln \left[\frac{\mu_0 I}{2a} \ln \frac{b}{a} + \frac{\mu_0 I}{2a} \ln \frac{b}{a} \right] = \mu_0 \frac{I}{a} \ln \left[\frac{1}{2} \left(\frac{\mu_0 I}{2a} \ln \frac{b}{a} \right) \right]$

b) For a very long transmission line, assume

$$R = R_0 \frac{dl}{L} \quad \Delta V = R_0 \frac{dl}{L} \left(\frac{V}{L} \right) = \frac{R_0 V}{L} \frac{dl}{L}$$

c) $R = R_0 \frac{dl}{L} = R_0 \frac{dl}{L}$

$$= R_0 \frac{dl}{L} \left[\frac{1}{1 - \beta^2} - \frac{1}{1 - \beta^2} \right] = R_0 \frac{dl}{L} \left[\frac{1}{1 - \beta^2} - \frac{1}{1 - \beta^2} \right]$$

$$= R_0 \frac{dl}{L} \left[\frac{1}{1 - \beta^2} - \frac{1}{1 - \beta^2} \right]$$

or by using the equation for magnetic flux (line)

$$\frac{d\Phi}{dt} = \frac{d}{dt} \left[\frac{\mu_0 I l}{2\pi r} \right] = \frac{\mu_0 I}{2\pi r} \frac{dl}{dt} = 0$$

(since $dl = 0$ and $I = \text{constant}$)

Thus, $\frac{dV}{dt} = \frac{d}{dt} \left[\frac{R_0 I l}{L} \right] = 0$

Ex 11 Apply divergence theorem to $(\nabla \cdot \mathbf{E})$, where \mathbf{E} is a constant vector.

$$\int_V \nabla \cdot (\mathbf{E}) d\tau = \int_V (\nabla \cdot \mathbf{E}) d\tau \quad \text{--- (1)}$$

Now, from problem 11.11: $\nabla \cdot (\mathbf{E}) = \frac{\rho}{\epsilon_0}$ (11.11)

$$= \frac{\rho}{\epsilon_0} \quad \text{--- (2)}$$

From Eq. 11.11: $(\nabla \cdot \mathbf{E}) = \frac{\rho}{\epsilon_0}$

Substituting (2) and (3) in (1)

$$\int_V (\nabla \cdot \mathbf{E}) d\tau = \int_V \left(\frac{\rho}{\epsilon_0} \right) d\tau = \int_V \left(\frac{\rho}{\epsilon_0} \right) d\tau$$

Ex 12

$$\int_V \nabla \cdot \mathbf{E} d\tau$$

a) Given $\mathbf{E} = E_0 \hat{x}$

$$\nabla \cdot \mathbf{E} = \frac{\partial E_x}{\partial x} = \frac{\partial E_0}{\partial x} = 0$$

b) Given $\mathbf{E} = E_0 \hat{y}$

$$\nabla \cdot \mathbf{E} = \frac{\partial E_y}{\partial y} = \frac{\partial E_0}{\partial y} = 0$$

Ex 13 a) Given

$$\mathbf{E} = E_0 \hat{x}$$

or $\mathbf{E} = E_0 \hat{x}$

$$\nabla \cdot \mathbf{E} = \frac{\partial E_x}{\partial x} = \frac{\partial E_0}{\partial x} = 0$$

or $\nabla \cdot \mathbf{E} = 0$

b) $\int_V \nabla \cdot \mathbf{E} d\tau = \int_V \left(\frac{\rho}{\epsilon_0} \right) d\tau = \int_V \left(\frac{\rho}{\epsilon_0} \right) d\tau$

Ex 11.11



$$\text{a) } \vec{E} = \gamma \rho_0 \vec{E}_z = -\vec{E}_z \rho_0 \frac{\partial \phi}{\partial z} = \vec{E}_z \frac{\rho_0 h}{2} \frac{\partial \phi}{\partial z}, \text{ which is the same as } \vec{E}_z \text{ (11.11).}$$

Ex 11.12 A cylindrical bar magnet having a uniform magnetization M along the z -axis is equivalent to a \vec{E}_z field (11.11) and a \vec{E}_θ field $\vec{E}_\theta = \vec{M} \times \vec{e}_\theta = M \sin \theta \vec{e}_\theta = M \sin \theta \vec{e}_\theta$ on the cylinder with \vec{e}_θ at its local point. It also has the \vec{E}_z field along an a cylindrical shell of length h and radius b in the same as direction as a circular loop of radius b carrying a current I (11.11). It is given by \vec{E}_θ (11.11), which is the same as \vec{E}_θ (11.11) obtained in Example 1.7 where the field of a current element of the cylindrical magnet is \vec{E}_θ (11.11) (11.11).

Ex 11.13



$$\text{a) } \vec{E}_z = \vec{E}_z \vec{e}_z = \vec{E}_z \vec{e}_z$$

$$\vec{E}_\theta = \vec{E}_\theta \vec{e}_\theta = \vec{E}_\theta \vec{e}_\theta = \vec{E}_\theta \vec{e}_\theta$$

$$\text{b) Apply } \vec{E}_\theta \text{ (11.11) to a loop of radius } b \text{ and } \vec{E}_\theta \text{ carrying a current } I. \\ \vec{E}_\theta \text{ (11.11):} \\ d\vec{E} = \frac{\mu_0 I}{4\pi R^2} \frac{d\vec{l} \times \vec{r}}{r^3} \\ = \frac{\mu_0 I}{4\pi R^2} \frac{d\vec{l} \times \vec{r}}{r^3}$$

$$\vec{E} = \int d\vec{E} = \frac{\mu_0 I}{4\pi R^2} \int \frac{d\vec{l} \times \vec{r}}{r^3} = \frac{\mu_0 I}{4\pi R^2} \int \frac{d\vec{l} \times \vec{r}}{r^3}$$

Ex. 12 a) $V_1 = \frac{1}{R_1} = \frac{1}{20 + j10} = \frac{20 - j10}{20^2 + 10^2} = 0.2 - j0.1 \text{ (A)}$
 $V_2 = \frac{20 \times j10}{20^2 + 10^2} = 0.2 + j0.1 \text{ (A)}$

b) $V_3 = V_1 + V_2 = \frac{20 - j10}{20^2 + 10^2} + \frac{20 + j10}{20^2 + 10^2} = \frac{40}{20^2 + 10^2} \text{ (A)}$
 $V_4 = \frac{1}{R_2} = \frac{1}{20} = 0.05 \text{ (A)}$
 $V_5 = \frac{1}{R_3} = \frac{1}{20} = 0.05 \text{ (A)}$

c) $V_1 + V_2 = 2(V_4 + V_5)$, $V_3 = \frac{1}{R_1} + \frac{1}{R_2} = 0.25 \text{ (A)}$

Ex. 13 Magnetic circuit:



$\frac{1}{R_1} = \frac{1}{20 + j10} = 0.2 - j0.1$

Multiplying by conjugate

Then add remaining identical lines along with it

$V_1 = \frac{20}{20^2 + 10^2} = \frac{20}{500} = 0.04 \text{ (A)}$

$V_2 = \frac{10}{20^2 + 10^2} = 0.02 \text{ (A)}$

a) $V_3 = \frac{20}{20 + 20} = 0.5 \text{ (A)}$, $V_4 = \frac{1}{20} = 0.05 \text{ (A)}$

b) $V_5 = \frac{1}{20 + 20} = 0.025 \text{ (A)}$

$V_6 = \frac{1}{20} = 0.05 \text{ (A)}$

$V_7 = \frac{1}{20} = 0.05 \text{ (A)}$

Ex. 14 a) $V_1 = 100 \text{ V}$ per unit length in direction of

$V_2 = 100 \text{ V}$

which per unit volume in direction of $V_3 = 100 \text{ V}$

Then $V_4 = 100 \text{ V}$

Ex. 15 $V_1 = 100 \text{ V}$, $V_2 = 100 \text{ V}$

$V_3 = 100 \text{ V}$, $V_4 = 100 \text{ V}$

$V_5 = 100 \text{ V}$

$V_6 = 100 \text{ V}$

Ex 11



$$a) \vec{r}_1 = \vec{r}_2 \cos \alpha = \vec{r}_2 \sin \beta$$

$$\vec{r}_1 = \vec{r}_2 \sin \beta = \vec{r}_2 \cos \alpha$$

$$\vec{r}_2 = \frac{\vec{r}_1}{\sin \beta} = \vec{r}_2 = \frac{\vec{r}_1}{\cos \alpha}$$

$$\rightarrow \vec{r}_2 = \vec{r}_1 \sec \alpha$$

$$\vec{r}_2 = \vec{r}_1 \sec \alpha$$

$$\therefore \vec{r}_1 = \vec{r}_2 \cos \alpha = \vec{r}_2 \sin \beta$$

$$\cos \alpha = \frac{\vec{r}_1}{\vec{r}_2} \cos \alpha = \cos \left(\frac{\vec{r}_1}{\vec{r}_2} \right) = \cos \alpha \quad \text{--- (1) ---}$$

$$b) \text{ If } \vec{r}_1 = \vec{r}_2 \cos \alpha = \vec{r}_2 \sin \beta, \quad \vec{r}_1 = \vec{r}_2 \cos \alpha + \vec{r}_2 \sin \beta$$

$$\vec{r}_2 = \frac{\vec{r}_1}{\cos \alpha} = \vec{r}_2 = \frac{\vec{r}_1}{\sin \beta} \rightarrow \vec{r}_2 = \frac{\vec{r}_1}{\cos \alpha} = \frac{\vec{r}_1}{\sin \beta}$$

$$\vec{r}_2 = \vec{r}_2 \cos \alpha + \vec{r}_2 \sin \beta$$

$$\vec{r}_2 = \vec{r}_2 \cos \alpha + \vec{r}_2 \sin \beta = \vec{r}_2 \cos \alpha + \vec{r}_2 \sin \beta$$

Ex 12) a) Consider two situations (I) and (II) both in air and (II) in water. Both in magnetic medium with relative permeability μ .



Find \vec{r}_1 and \vec{r}_2 at P (air)

$$\vec{r}_1 = \frac{\vec{r}_2}{\cos \alpha} = \frac{\vec{r}_2}{\sin \beta}$$

$$\vec{r}_2 = \frac{\vec{r}_1}{\cos \alpha} = \frac{\vec{r}_1}{\sin \beta}$$

$$\vec{r}_2 = \frac{\vec{r}_1}{\cos \alpha} = \frac{\vec{r}_1}{\sin \beta}$$

$\therefore \vec{r}_1 = \vec{r}_2 \cos \alpha$ and $\vec{r}_2 = \frac{\vec{r}_1}{\sin \beta}$ (Boundary conditions satisfied)

b) For $\mu_1 = \mu_2$, $\vec{r}_1 = \frac{\vec{r}_2}{\cos \alpha} = \frac{\vec{r}_2}{\sin \beta}$

Refer to the following figure.



$$E_1 = \frac{xy}{r^3} (-x, -y, 0)$$

$$E_2 = \frac{xy}{r^3} (-x, -y, 0)$$

$$\therefore E = E_1 + E_2$$

$$= -E_1 \frac{xy}{r^3} \left[\frac{y^2+z^2}{r^2} + \frac{xy^2}{r^2} \right] + E_2 \frac{xy}{r^3} \left[\frac{y^2+z^2}{r^2} + \frac{xy^2}{r^2} \right]$$

Ex-14



$$\text{a) If } z=0, \quad E_1 = E_2 = 0$$

$$E_1 \text{ continuous at } (0,0,0)$$

$$E_2 = E_1 \implies E = -E_1 E_2$$

$$\text{Average } E = 0 \text{ (Average of 0)}$$

$$\text{b) If } z \neq 0, \quad E_1 = 0, \text{ but } E_2 \text{ is finite.}$$

At surface current $\implies E_1 = E_2 = 0$

E_1 continuous at $(0,0,0)$

Average E_2 (-z) pointing into the page

$$\text{d) } E_2 = E_1 + (E_2)_z, \text{ where } E_1 = \frac{xy}{r^3} (-x, -y, 0) \text{ and } (E_2)_z = \frac{xy}{r^3} (-x, -y, 0)$$

$$\text{At } E_2 = E_1 = (E_2)_z = E_1 = (E_2)_z$$

$$\text{At } E_2 = -E_1 (E_2)_z = -E_1 (E_2)_z$$

$$\text{At } E_2 = 0$$

Ex-15



$$E = K_1 K_2 = K_1 K_2 \frac{r^2}{R^2}, \quad r = R - \text{radius}$$

$$E = \frac{K_1 K_2}{2\pi} \int_0^{2\pi} \int_0^R \frac{r^2}{R^2} \frac{2\pi R dr}{R} = K_1 K_2 (R - \frac{R^2}{2R})$$

$$\therefore E = \frac{K_1 K_2}{2} (R - \frac{R}{2}) = \frac{K_1 K_2 R}{4}$$

$$\text{At } z=0, \quad E_2 \text{ is } \frac{K_1 K_2 R}{4} \text{ (continuous)}$$

$$E = K_1 K_2 = K_1 K_2 \frac{R^2}{R^2} = K_1 K_2$$

Ex. 22. For parallel, $E_1 = \frac{1}{2} E_2 = \frac{1}{2} \frac{\mu_0 I}{2\pi r} \left[1 - \frac{r^2}{(a+r)^2} \right]$
 $= \frac{1}{4} \frac{\mu_0 I}{\pi r} \left[\frac{2a^2 + 4ar + 2r^2}{(a+r)^2} \right]$.

Magnetic energy per unit length stored in the tube is

$$W'_L = \frac{1}{2} \int_{a-r}^{a+r} B_1^2 \pi r dr$$

$$= \frac{\mu_0^2 I^2}{8\pi} \left\{ \frac{2a^2 + 4ar + 2r^2}{(a+r)^2} \pi r \right\}$$

From Eqs. (2-197), (2-198) and (2-194) we have

$$L = \frac{1}{I^2} (W'_L + W'_L + W'_L)$$

$$= \frac{\mu_0}{4} \left[\frac{1}{2} + \frac{1}{2} + \frac{2a^2 + 4ar + 2r^2}{(a+r)^2} \pi r \right] \pi$$

Ex. 23



Let a distance r from an infinitely long line carrying a current $I = I_0 \frac{z}{\sqrt{z^2 + a^2}}$

For a unit length the flux due to I is less than that due to the current flow pair I & $-I$

$$E_1 = \frac{\mu_0 I}{2\pi} \int_{a-r}^a \frac{1}{r^2} = \frac{\mu_0 I}{2\pi} \ln \frac{a}{a-r}$$

That due to $-I$ is E_2 is

$$E_2 = \frac{\mu_0 I}{2\pi} \ln \frac{b}{b-r}$$

Total flux linkage per unit length

$$L'_L = E_1 - E_2 = \frac{\mu_0 I}{2\pi} \ln \frac{a(b-r)}{(a-r)b}$$

$$= \frac{\mu_0 I}{2\pi} \ln \frac{a(b-r)}{a(b-r)} = \frac{\mu_0 I}{2\pi} \ln \frac{a}{a-r}$$

$$\therefore W'_L = \frac{1}{2} L'_L = \frac{\mu_0 I^2}{4\pi} \ln \left(1 + \frac{r}{a-r} \right)$$

Ex. 24 For I in the long straight wire, E is $\frac{\mu_0 I}{2\pi r}$

$$W'_L = \int E \cdot dl = \int \frac{\mu_0 I}{2\pi r} \pi r dr = \frac{\mu_0 I^2}{4} \ln \left(\frac{b}{a} \right) \pi$$

$$= \frac{\mu_0 I^2}{4} \left[\frac{1}{2} \ln \left(\frac{b}{a} \right) + \frac{1}{2} \right] \pi = \frac{\mu_0 I^2}{4} \left[\frac{1}{2} \ln \left(\frac{b}{a} \right) + \frac{1}{2} \right] \pi$$

Ex:42



Assume a current I .

$$B \text{ at } P \text{ due to } I = \frac{\mu_0 I a}{2\pi (a^2 + z^2)^{3/2}}$$

$$A_{\perp} = \frac{\mu_0 I}{2\pi} \int_0^{2\pi} \int_0^R \frac{a \, dr \, d\theta}{(a^2 + z^2)^{3/2}}$$

$$= \frac{\mu_0 I}{2\pi} \int_0^{2\pi} \frac{2\pi R a}{(a^2 + z^2)^{3/2}} = \frac{\mu_0 I R a}{(a^2 + z^2)^{3/2}}$$

$$A_{\parallel} = \mu_0 I a \int_0^{2\pi} \frac{R \cos \theta \, d\theta}{(a^2 + z^2)^{3/2}}$$

Ex:43 Approximate the magnetic flux due to the long loop taking with the small loop by that due to two infinitely long wires carrying equal and opposite current I .

$$A_{\perp} = \frac{\mu_0 I a}{2\pi} \int \left(\frac{1}{a+r} - \frac{1}{a+r+2a} \right) dr = \frac{\mu_0 I a}{2\pi} \ln \left(\frac{a+R}{a-R} \right)$$

$$A_{\parallel} = \frac{\mu_0 I}{2\pi} \int \frac{2a \cos \theta \, d\theta}{(a+r+2a)^2}$$

Ex:44 $\vec{B} = (a - bx) \hat{i} + (c + 2bx) \hat{j} + (d + bx) \hat{k}$

$$\Rightarrow \nabla \cdot \vec{B} = \frac{\partial}{\partial x} (a - bx) + \frac{\partial}{\partial y} (c + 2bx) + \frac{\partial}{\partial z} (d + bx) = -b + 2b + b = 2b$$

$$\frac{\partial \rho}{\partial x} = \frac{\partial}{\partial x} (2bx) = 2b, \quad \frac{\partial \rho}{\partial y} = \frac{\partial \rho}{\partial z} = 0.$$

$$\therefore \rho = \frac{2b}{2} = b \text{ for minimum } \rho.$$

$$\Rightarrow \nabla \times \vec{B} = \frac{\partial}{\partial x} (c + 2bx) \hat{j} - \frac{\partial}{\partial x} (d + bx) \hat{k} \rightarrow b \hat{j} - \frac{1}{2} b \hat{k}.$$

Ex:45



$$B_1 = B_2 = B_3 = \frac{\mu_0 I}{2\pi a}, \quad a = \frac{a}{\sqrt{3}}$$

$$B_1 = \frac{\mu_0 I}{2\pi a} \hat{i}, \quad B_2 = \frac{\mu_0 I}{2\pi a} \hat{j}$$

Force per unit length on wire 1

$$F_1 = -I_1 B_2 = -I_1 \frac{\mu_0 I}{2\pi a} \hat{j}$$

$$\Rightarrow -I_1 \mu_0 I_2 \hat{j} = -I_1 I_2 \frac{\mu_0}{2\pi a} \hat{j} \text{ (out)}$$

Force on all three wires are of equal magnitude and towards the center of the triangle.

Ex 11 Magnetic field intensity at the wire due to the current of $I = \frac{1}{2} \text{ amp}$ in an elemental dip is



$$dH = \frac{\mu_0 I}{4\pi r'^2} = \frac{\mu_0 I dl \cos \theta}{4\pi r'^2}$$

Symmetry \rightarrow H at the wire has only in y-component.

$$H = \mu_0 I \int \cos \theta \left(\frac{1}{r'^2} \right) = \mu_0 I \int \frac{\cos \theta dl}{r'^2}$$

$$= \mu_0 I \int \frac{\cos^3 \theta}{r'^2} dl$$

$$F = F \sin \theta = (-\mu_0 I a) \left[\mu_0 I a \right] = \mu_0 I^2 \frac{a^2}{r^2} \cos^3 \left(\frac{\theta}{2} \right) \quad (1)(2)$$

Ex 12 From Problem 11-11 we have the y-component of the magnetic flux density at an arbitrary point P(x, z) in the right-hand wire due to I_1 in the left-hand wire

$$B_{y1} = -\frac{\mu_0 I_1}{2\pi r} \left[\cos^{-1} \left(\frac{z}{r} \right) + \cos^{-1} \left(\frac{2a+z}{r} \right) \right]$$

The x-component of the force on a strip of width dy due to I_2 in the right-hand conductor is

$dF_x = \left(\frac{\mu_0}{2\pi} I_2 \right) dy$ (in the +x direction, a repulsive force)

$$F_x = \mu_0 I_2 \int dy = \mu_0 I_2 \frac{dy}{2\pi r} \int \left[\cos^{-1} \left(\frac{z}{r} \right) + \cos^{-1} \left(\frac{2a+z}{r} \right) \right] dy$$

$$= \mu_0 I_2 \frac{dy}{2\pi r} \left[2 + \cos^{-1} \left(\frac{z}{r} \right) + \cos^{-1} \left(\frac{2a+z}{r} \right) \right] \quad \text{per unit length}$$

There is equal force in the y-direction.

Ex 13 B due to I_2 in the straight wire in the x-direction at an elemental area dA in the circular loop is

$$dB = \mu_0 \frac{I_2 dA \sin \theta}{4\pi r^2}$$



F has no net y-component

$$F_x = -\mu_0 I_2 \int \left(\frac{1}{r} \right) \cos \theta$$

$$= -\mu_0 I_2 \int \frac{dA \sin \theta \cos \theta}{4\pi r^2} = -\mu_0 I_2 \int \frac{dA \sin \theta \cos \theta}{4\pi (a^2 + x^2)}$$

$$= \mu_0 I_2 \int \frac{dA}{2\pi (a^2 + x^2)} = \mu_0 I_2 \int \frac{dA}{2\pi (a^2 + x^2)} \quad (\text{repulsive force})$$

Ex. 22 $\vec{E} = \epsilon_0 \frac{\partial \vec{E}}{\partial t} \left(\frac{1}{r} + \frac{1}{r^2} \right)$ $d\vec{E} = \epsilon_0 \vec{E} d\tau$



(A rectangular problem.)

$$d^2 \vec{E} = \epsilon_0 \frac{\partial^2 \vec{E}}{\partial t^2} \left(\frac{1}{r} + \frac{1}{r^2} \right) d\tau$$

$$F = -\epsilon_0 \frac{\partial \vec{E}}{\partial t} \left(\frac{1}{r} + \frac{1}{r^2} \right) d\tau$$

$$= -\epsilon_0 \frac{\partial^2 \vec{E}}{\partial t^2} \ln \left(\frac{1}{r} + \frac{1}{r^2} \right)$$

Ex. 23



Separate in $x(z)$, $y(z)$, $z(z)$

$(\epsilon_0 \text{ etc. } \rightarrow \text{constant, } \text{etc. } \rightarrow \text{etc. } \rightarrow \text{etc.})$

Sum \vec{E}_x at \vec{E}_y at \vec{E}_z at \vec{E}

Separating:

$$F = -\vec{E} \cdot \vec{E} = -(\vec{E}_x \cdot \vec{E}_x) + \vec{E}$$

$$= -(\vec{E}_x \cdot \vec{E}_x) + \vec{E} = -\epsilon_0 \frac{\partial^2 \vec{E}}{\partial t^2} \ln \left(\frac{1}{r} + \frac{1}{r^2} \right)$$

Ex. 24 Let $\mu = \epsilon_0 \mu_0$ be the permeability of the straight wire.

The magnetic energy stored in a section of length L is

$$W_M = \frac{1}{2} L I^2$$

$$L = \frac{W_M}{I^2} = \frac{1}{2} \int_0^L \mu_0 \vec{E} \cdot d\vec{E} = \frac{1}{2} \int_0^L \frac{\mu_0 I^2}{2\pi r} dr = \frac{\mu_0 I^2 L}{4\pi}$$

$$\vec{E} = \epsilon_0 \frac{\partial \vec{E}}{\partial t} = \epsilon_0 \left(\frac{1}{r} \right) \frac{\partial \vec{E}}{\partial t} = \epsilon_0 \frac{\partial^2 \vec{E}}{\partial t^2} \ln \left(\frac{1}{r} + \frac{1}{r^2} \right)$$

Ex. 25 Resolve the circular loop into many small loops, each with a magnetic dipole moment $\vec{m} = \epsilon_0 \vec{E} \cdot d\vec{E}$, $d\vec{E} = \epsilon_0 \vec{E} d\tau$.

$$F = \int d\vec{E} \cdot \vec{E} = \epsilon_0 \int d\vec{E} \cdot \vec{E} = \epsilon_0 \int d\vec{E} \cdot \vec{E} = \epsilon_0 \int d\vec{E} \cdot \vec{E} = \epsilon_0 \int d\vec{E} \cdot \vec{E}$$

— This torque is in the direction of aligning the dipoles produced by \vec{E} in the loop with that of \vec{E} due to \vec{E} in the straight wire.

Ex 11 \vec{E} at the end of the large slender bar of wire carrying a current I_0 is \vec{E}_1 (along \vec{e}_1) and \vec{E}_2 (along \vec{e}_2)

$$\vec{E} = I_0 \mu_0 \frac{2\pi R}{l}$$

Force on the small slender wire: $F = I_0 \int \vec{e}_1 \times (I_0 \mu_0 \frac{2\pi R}{l} \vec{e}_1) + I_0 \int \vec{e}_2 \times (I_0 \mu_0 \frac{2\pi R}{l} \vec{e}_2) = I_0^2 \mu_0 \frac{2\pi R}{l} \int \vec{e}_1 \times \vec{e}_1 + I_0^2 \mu_0 \frac{2\pi R}{l} \int \vec{e}_2 \times \vec{e}_2$
 must be perpendicular to \vec{e}_1 and \vec{e}_2 in a direction to align the magnetic flux produced by I_0 & I_1 .

Ex 12



\vec{E} (superposition of current elements)

$$= \frac{\mu_0 I dl}{4\pi r^2} (\cos\theta \vec{e}_z + \sin\theta \vec{e}_r) \frac{R}{r} \frac{2\pi R}{l}$$

$$= \frac{\mu_0 I dl R^2}{2\pi r^3} (\cos\theta \vec{e}_z + \sin\theta \vec{e}_r)$$

$$= \frac{\mu_0 I R^2}{2\pi r^3} (\cos\theta \vec{e}_z + \sin\theta \vec{e}_r) \int dl$$

\vec{E} (current) = $-I_0 \mu_0 \vec{e}_z$ (??)

Min. deflection occurs when $|\frac{dE}{dz}|$ is max. at when

$$\left| \frac{dE}{dz} \right| = \left| \frac{d}{dz} \left(\frac{\mu_0 I R^2}{2\pi r^3} (\cos\theta \vec{e}_z + \sin\theta \vec{e}_r) \right) \right|$$
 is max.

Set $\frac{dE}{dz} = 0$ (at $\theta = 0$) and solve for z .
 At $\theta = 0$, $\vec{e}_r = \vec{e}_z$, and $r = \sqrt{R^2 + z^2}$.

Ex 13

$$F = \frac{dW}{dx} = \frac{d}{dx} \left(\frac{\mu_0 I_1 I_2}{4\pi} \int \frac{dl_1 dl_2}{r} \right) = \frac{d}{dx} \left(\frac{\mu_0 I_1 I_2}{4\pi} \int \frac{dl_1 dl_2}{\sqrt{a^2 + x^2}} \right)$$

$$F = \frac{\mu_0 I_1 I_2}{4\pi} \frac{d}{dx} \left(\frac{2\pi a}{\sqrt{a^2 + x^2}} \right) = \frac{\mu_0 I_1 I_2}{2} \frac{-2ax}{(a^2 + x^2)^{3/2}}$$

$$F = -\frac{\mu_0 I_1 I_2 a^2 x}{(a^2 + x^2)^{3/2}}$$

Setting: $\frac{dF}{dx} = 0$ (at $x = 0$) and solve for x .

Ex 14

$$W_C = \int_C \vec{F} \cdot d\vec{r}$$

Assume a virtual displacement δx of the free end.

$$W_C(\delta x) = W_C(x) + \delta W_C = \int_C \vec{F} \cdot d\vec{r} + \delta W_C$$

$$= W_C(x) + \delta W_C = \int_C \vec{F} \cdot d\vec{r} + \delta W_C$$

$$\delta W_C = \frac{\partial W_C}{\partial x} \delta x = \frac{\partial}{\partial x} \left(\int_C \vec{F} \cdot d\vec{r} \right) \delta x = \delta W_C$$

is the variation of W_C .

Chapter 7

Time-Varying Field and Maxwell Equations

Ex. 1 $\nabla \cdot \mathbf{E} = \int_V \frac{\partial \rho}{\partial t} d\tau = -\int_V \frac{\partial}{\partial t} (\nabla \cdot \mathbf{A}) d\tau = -\int_V \nabla \cdot \frac{\partial \mathbf{A}}{\partial t} d\tau$

Ex. 2 $\mathbf{E} = -\nabla \phi - \text{curl}(\mathbf{A} \times \mathbf{r}) - \text{grad}(\mathbf{A} \cdot \mathbf{r}) - \text{grad}(\mathbf{A} \cdot \mathbf{r})$

$$\int_V \mathbf{E} \cdot d\mathbf{r} = \int_V \left[-\nabla \phi - \text{curl}(\mathbf{A} \times \mathbf{r}) - \text{grad}(\mathbf{A} \cdot \mathbf{r}) - \text{grad}(\mathbf{A} \cdot \mathbf{r}) \right] \cdot d\mathbf{r} = -\int_V \nabla \cdot \left[\phi + (\mathbf{A} \times \mathbf{r}) \cdot \nabla + (\mathbf{A} \cdot \mathbf{r}) \cdot \nabla \right] d\tau$$

$$\mathbf{V} = -\int_V \nabla \cdot \mathbf{E} d\tau = \int_V \left[\nabla \cdot (\mathbf{A} \times \mathbf{r}) + \nabla \cdot (\mathbf{A} \cdot \mathbf{r}) + \nabla \cdot (\mathbf{A} \cdot \mathbf{r}) \right] d\tau$$

$$= \int_V \left[\nabla \cdot (\mathbf{A} \times \mathbf{r}) + \nabla \cdot (\mathbf{A} \cdot \mathbf{r}) + \nabla \cdot (\mathbf{A} \cdot \mathbf{r}) \right] d\tau = \int_V \left[\nabla \cdot (\mathbf{A} \times \mathbf{r}) + \nabla \cdot (\mathbf{A} \cdot \mathbf{r}) \right] d\tau$$

Ex. 3 In the rectangular loop with the assigned direction for \hat{t}_1

$$\int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} = \int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} = \mathcal{E}_1 \quad (1)$$

$$\text{where } \mathcal{E}_1 = \frac{d}{dt} \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} = \frac{d}{dt} \int_{\hat{t}_1} \frac{\mu_0 \mathbf{J} \times \mathbf{r}}{4\pi r^3} \cdot d\mathbf{r} = \frac{d}{dt} \int_{\hat{t}_1} \frac{\mu_0 \mathbf{J} \cdot d\mathbf{r}}{4\pi r^2}$$

(2) At $t=0$, $\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$ is applied and (2) becomes

$$\int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} + \mathcal{E}_1 = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} \quad (3)$$

Solution of (3): $\int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} - \mathcal{E}_1$

At $t=0$, $\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$ when a negative sign function $-\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$ is applied. If $\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$, then $\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$ is the reverse of $\int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$

$$\text{So, } \int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} - \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$$



(4) Energy dissipated in \hat{t}_1

$$W = \int_{\hat{t}_1} \frac{d\mathbf{E}}{dt} \cdot d\mathbf{r} = \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r} - \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$$

$$= \int_{\hat{t}_1} \mathbf{E} \cdot d\mathbf{r}$$

4) For 2 insulated elementary parts, each with an area $S = 2R^2 \sin^2 \alpha = \pi R^2 \sin^2 \alpha$ and $\sigma_1 = \sigma_2 = \sigma$,

$$\text{Power loss in 2 elements } P_{\text{loss}} = \left(\frac{2\sigma R^2 \sin^2 \alpha}{2} \right) \left(\frac{2\sigma R^2 \sin^2 \alpha}{2} \right) \int_0^{2\pi} \int_0^{2\pi} \cos^2 \alpha \, d\alpha \, d\alpha = \frac{\sigma^2 R^4 \sin^4 \alpha}{2}$$

$$P_{\text{loss}} = \frac{\sigma^2 R^4 \sin^4 \alpha}{2}$$

Ex 2.12 $\vec{E}(t) = E_0 \cos(\omega t) \hat{z} = -(\nabla \phi \cos \omega t) = -\nabla \phi \cos \omega t$
 $= -\Delta \phi \cos \omega t \quad (\nabla \cdot \cos \omega t \hat{z} = 0 = \nabla^2 \phi)$

$$\Delta \phi = -\frac{1}{\epsilon_0} \frac{\rho}{4\pi} = -\frac{1}{\epsilon_0} \Delta \phi \cos \omega t \quad \text{in part 1}$$

$$\Rightarrow -\epsilon_0 \Delta \phi \cos \omega t = \Delta \phi \cos \omega t \quad \text{in part 2}$$

Ex 2.13



Assuming the loop to have N turns each with an area each, the torque on the loop is \vec{T} with $\vec{T} = N \vec{\mu} \times \vec{E}$

Mechanical work done by the motor in rotating through an angle 2π rad:

$$W_{\text{mech}} = \int_0^{2\pi} \tau \, d\alpha = N \mu E \sin \alpha \int_0^{2\pi} d\alpha$$

Flux: Linking with the loop, $\vec{B} = \mu_0 N I \cos \alpha \hat{z}$

Magnet moment in the loop, $\vec{\mu} = \frac{q v}{2} = \mu_0 N I \left(\frac{q v}{2} \right) \hat{z}$

Electric energy required to send current I against this field in time Δt , $W_{\text{el}} = \int I \Delta \phi \, dt = N I \Delta \phi \cos \alpha \int_0^{2\pi} d\alpha$

Ex 2.14 $\vec{E} = \frac{\rho}{\epsilon_0} = \frac{1}{\epsilon_0} \frac{dQ}{dr} \hat{r} = \frac{1}{\epsilon_0} \frac{d}{dr} \left(\int_0^r \rho(r') r'^2 dr' \right) \hat{r}$
 $= \frac{\rho}{\epsilon_0} \hat{r}$

$\vec{J}_p = \sigma \hat{r}$ (radial current) (Power distributed in R_1)

On the other hand,

for $R_1 < r < R_2$, $\vec{J}_p = \sigma \hat{r}$, $\vec{E} = \frac{\rho}{\epsilon_0} \hat{r}$, $\vec{J}_p \cdot \vec{E} = \sigma \rho$

for $r > R_2$, $\vec{J}_p = 0$, $\vec{E} = \frac{\rho}{\epsilon_0} \hat{r}$, $\vec{J}_p \cdot \vec{E} = 0$

Mechanical power required to rotate $\omega(t)$:

$$P_{\text{mech}} = \int (\vec{J}_p \cdot \vec{E}) \, dV = \int \sigma \rho \, dV = P_{\text{el}}$$

(Alternatively, $\vec{E} = \nabla \phi$, where $\nabla \cdot \vec{E} = \rho / \epsilon_0$, and $\vec{J}_p \cdot \vec{E} = \sigma \rho$)

Ex-10 Median of two years.

Median 1: $\mu_1 = 10$, σ_1 must be such as that $\frac{10 - \mu_2}{\sigma_1} = 0$

$$\text{Standard deviation, } \mu_2 = 9, \sigma_2 = 6.$$

$$\mu_1 = \mu_2 + \sigma_2 \cdot (Z - 0) = 9.$$

Ex-11 We will use given $F'(x) = \frac{1}{2} \frac{dF}{dx} = \frac{1}{2}$. (1)

$$\text{where } F(x) = \frac{1}{2} \int_{-\infty}^x f(x) dx. \quad (2)$$

$$\text{We need } F\left(\frac{1}{2}\right) = \frac{1}{2} F\left(\frac{1}{2}\right) + \frac{1}{2} F\left(\frac{1}{2}\right) = 2F\left(\frac{1}{2}\right) \quad (3)$$

$$\left[\text{Formula: } F'(x) = F - F(0) + \int_0^x F'(x) dx = 2F(x) \right]$$

$$\text{Let } x = \frac{1}{2}, F\left(\frac{1}{2}\right) = \frac{1}{2} \int_{-\infty}^{\frac{1}{2}} f(x) dx = \frac{1}{2} \int_{-\infty}^{\frac{1}{2}} f(x) dx. \quad (4)$$

$$F\left(\frac{1}{2}\right) = -\ln\left(\frac{1}{2}\right) \quad (5), \quad F\left(\frac{1}{2}\right) = \frac{1}{2} \int_{-\infty}^{\frac{1}{2}} f(x) dx. \quad (6)$$

$$\text{Substituting (5) in (6) } -\ln\left(\frac{1}{2}\right) = \frac{1}{2} \int_{-\infty}^{\frac{1}{2}} f(x) dx = \ln(2). \quad (7)$$

$$\text{From (7) } F(x) = \frac{1}{2} \int_{-\infty}^x f(x) dx = \ln(2) \int_{-\infty}^x [f(x) dx - \ln(2) dx] \quad (8)$$

$$\frac{1}{2} \int_{-\infty}^x f(x) dx = \ln(2) \int_{-\infty}^x f(x) dx$$

$$\therefore F(x) = \frac{1}{2} \int_{-\infty}^x f(x) dx = \ln(2) \int_{-\infty}^x [f(x) dx - \ln(2) dx] dx \\ = \frac{1}{2} \quad \text{Q.E.D.}$$

Ex-12 a)



$$\text{Ex-12 } \int_0^1 (x - x^2) dx = \int_0^1 x dx - \int_0^1 x^2 dx = \left[\frac{x^2}{2} \right]_0^1 - \left[\frac{x^3}{3} \right]_0^1 = \frac{1}{2} - \frac{1}{3} = \frac{1}{6}$$

$$\int_0^1 (x - x^2) dx = \int_0^1 x dx - \int_0^1 x^2 dx = \left[\frac{x^2}{2} \right]_0^1 - \left[\frac{x^3}{3} \right]_0^1 = \frac{1}{2} - \frac{1}{3} = \frac{1}{6}$$

$$\text{Answer: } \int_0^1 (x - x^2) dx = \frac{1}{6} \left[\frac{x^2}{2} - \frac{x^3}{3} \right]_0^1 = \frac{1}{6} \left[\frac{1}{2} - \frac{1}{3} \right] = \frac{1}{6} \left[\frac{3}{6} - \frac{2}{6} \right] = \frac{1}{6} \left[\frac{1}{6} \right] = \frac{1}{36}$$

$$= \frac{1}{36} \left[\frac{1}{6} \right] = \frac{1}{216} \quad \text{Q.E.D.}$$

$$\therefore \int_0^1 (x - x^2) dx = \frac{1}{36} \quad \text{Q.E.D.}$$

$$\begin{aligned} \text{E11.13} \quad \Phi \cdot \mathcal{L} - \gamma \mu \frac{\partial^2 \Phi}{\partial x^2} &= 0 & \Phi \cdot \mathcal{H} = \mathcal{F} + \mu \frac{\partial \Phi}{\partial x} &= 0 \\ \Phi \cdot \mathcal{L} = \mathcal{F} & & \Phi \cdot \mathcal{H} = 0 & \end{aligned}$$

$$\mathcal{F} = 0: \quad \Phi \cdot \partial_x \mathcal{L} = \gamma \mu \frac{\partial^2 \Phi}{\partial x^2} (\mathcal{F} = 0) = \gamma \mu \frac{\partial^2 \Phi}{\partial x^2} (\mathcal{F} + \mu \frac{\partial \Phi}{\partial x}) = \Phi \cdot \mathcal{H} = 0 \cdot \mathcal{L}$$

Wave equation for \mathcal{L} : $\partial^2 \mathcal{L} - \gamma \mu \frac{\partial^2 \Phi}{\partial x^2} = \mathcal{F} = 0$

$$\mathcal{F} = 0: \quad \Phi \cdot \partial_x \mathcal{H} = \partial_x \mathcal{F} + \mu \frac{\partial^2 \Phi}{\partial x^2} (\mathcal{H} = \mathcal{F} + \mu \frac{\partial \Phi}{\partial x}) = \partial_x \mathcal{F} = 0 \cdot \mathcal{H}$$

Wave equation for \mathcal{H} : $\partial^2 \mathcal{H} - \gamma \mu \frac{\partial^2 \Phi}{\partial x^2} = -\partial_x \mathcal{F}$

For sinusoidal time dependence: $\partial_t \rightarrow -i\omega$, $\partial_t^2 \rightarrow -\omega^2$

Helmholtz equations: $\partial^2 \mathcal{L} + \omega^2 \mu \mathcal{L} = \gamma \mu \partial^2 \Phi + \mathcal{F} = 0$

for \mathcal{H} : $\partial^2 \mathcal{H} + \omega^2 \mu \mathcal{H} = -\partial_x \mathcal{F}$

E11.14 $\mathcal{L} = \mathcal{H}$ as in Exercise E11.13 - $\gamma \mu \mathcal{L}$ (cont.)

Use plane wave $\mathcal{H} = \gamma \frac{\partial \mathcal{L}}{\partial x} = \frac{\partial}{\partial x} [\gamma \mu \frac{\partial \mathcal{L}}{\partial x} + \mathcal{F} + \mu \frac{\partial \mathcal{L}}{\partial x}]$

$$\mathcal{H} = \gamma \frac{\partial}{\partial x} \Phi \cdot \mathcal{H} = \gamma \frac{\partial}{\partial x} [\gamma \mu \frac{\partial \mathcal{L}}{\partial x} + \mathcal{F} + \mu \frac{\partial \mathcal{L}}{\partial x}]$$

Plane wave, $\mathcal{L} = \mathcal{H}$ as in Exercise E11.13

Equating (1) and (2): $\partial_x \mathcal{L} + \mathcal{F} = \gamma \mu \partial_x \mathcal{L} + \mathcal{F} + \mu \partial_x \mathcal{L}$
 $\implies \mu = \gamma \mu$ or $\mu = 0$ (trivial)

From (2): $\mathcal{H} = \gamma \mu \partial_x \mathcal{L}$
 $= \gamma \mu \partial_x \mathcal{L}$ (in total as $\mathcal{H} = \mathcal{L}$)
 $= \gamma \mu \partial_x \mathcal{L}$ (in total as $\mathcal{H} = \mathcal{L}$)

E11.15 $\mathcal{H} = \mathcal{L}$ as in Exercise E11.13 - $\gamma \mu \mathcal{L}$ (cont.)

Plane wave $\mathcal{H} = \mathcal{L}$ as in Exercise E11.13

Similar to Problem 11.14: $\partial_x \mathcal{L} + \mathcal{F} = \gamma \mu \partial_x \mathcal{L} + \mathcal{F} + \mu \partial_x \mathcal{L}$
 $\implies \mu = \gamma \mu$ or $\mu = 0$ (trivial)

$$\mathcal{H} = \gamma \mu \partial_x \mathcal{L} = [\gamma \mu \partial_x \mathcal{L} + \mathcal{F} + \mu \partial_x \mathcal{L}]$$

$$\mathcal{H} = \gamma \mu \partial_x \mathcal{L} = \gamma \mu \partial_x \mathcal{L} \text{ (in total as } \mathcal{H} = \mathcal{L} \text{)} \\ = \gamma \mu \partial_x \mathcal{L} \text{ (in total as } \mathcal{H} = \mathcal{L} \text{)} \text{ (trivial)}$$

Ex. 2.12 Use polar coordinates: $\mathbb{R}^2 = \mathbb{R}_+ \times \mathbb{S}^1$ via $(r, \theta) \mapsto (r \cos \theta, r \sin \theta)$
 $\mathbb{P} = \mathbb{R}^2 = \mathbb{R}_+ \times \mathbb{S}^1 \xrightarrow{\mathbb{R}^2} \mathbb{R}^2 \xrightarrow{\mathbb{R}^2} \mathbb{R}^2 \xrightarrow{\mathbb{R}^2} \mathbb{R}^2$
 $\xrightarrow{\mathbb{R}^2} \mathbb{R}^2 \xrightarrow{\mathbb{R}^2} \mathbb{R}^2 \xrightarrow{\mathbb{R}^2} \mathbb{R}^2$

In four space, $\mathbb{R}^4 = \mathbb{R}_+ \times \mathbb{S}^3 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4$
 $\mathbb{R}^4 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4 \xrightarrow{\mathbb{R}^4} \mathbb{R}^4$

Ex. 2.13 Maxwell's curl eqs: $\nabla \times \mathbb{E} = -j \omega \mu \mathbb{H}$, $\nabla \cdot \mathbb{D} = j \omega \rho$ (1)

From $\nabla \cdot \mathbb{E} = 0$, define \mathbb{A} such that $\mathbb{E} = \nabla \times \mathbb{A}$. (2)

From (1), $\mathbb{H} = \frac{1}{j\omega\mu} \nabla \times \mathbb{E} = \frac{1}{j\omega\mu} \nabla \times (\nabla \times \mathbb{A})$
 $= \frac{1}{j\omega\mu} [\nabla(\nabla \cdot \mathbb{A}) - \nabla^2 \mathbb{A}]$ (3)

From (2), $\nabla \cdot (\nabla \times \mathbb{A}) = 0$. Let $\mathbb{H} = j\omega\mu \mathbb{A} = -\nabla^2 \mathbb{A}$. (4)

Substituting (3) from (1): $\nabla \times \mathbb{A} = \frac{1}{j\omega\mu} [\nabla(\nabla \cdot \mathbb{A}) - \nabla^2 \mathbb{A}] = j\omega \mu \mathbb{H}$ (5)

Choose $\nabla \cdot \mathbb{A} = j\omega \mu \mathbb{H}$. (6)

Subst. (6) becomes $\mathbb{H} = j\omega \mu \mathbb{A} = \frac{1}{j\omega\mu} \nabla^2 \mathbb{A}$

Subst. (6) becomes $\nabla^2 \mathbb{A} = \omega^2 \mu \mathbb{A} = 0$, a homogeneous Helmholtz eq.

Ex. 2.14 $\mathbb{H} = j\omega \mu \mathbb{P} \times \mathbb{E}$ (1)

$\nabla \times \mathbb{E} = -j\omega \mu \mathbb{H} = -\omega^2 \mu \mathbb{P} \times \mathbb{E}$ (2)

$\nabla \cdot (\mathbb{E} - \epsilon_0 \mathbb{E}) = 0$. Let $\mathbb{E} - \epsilon_0 \mathbb{E} = \mathbb{P} \times \mathbb{E}$. (3)

$\nabla \times \mathbb{E} = j\omega \mu \mathbb{P} \times \mathbb{E} = j\omega \mu (\mathbb{E} - \epsilon_0 \mathbb{E})$ (4)

Substituting (3) and (4) in (2):

$j\omega \mu \mathbb{P} \times \mathbb{E} = j\omega \mu (\epsilon_0 \mathbb{E} + \mathbb{P} \times \mathbb{E} - \epsilon_0 \mathbb{E})$
 $= j\omega \mu (\mathbb{P} \times \mathbb{E} - \mathbb{P} \times \mathbb{E})$ (5)

Choose $\nabla \cdot \mathbb{E} = 0$. Eq. (3) becomes

$\mathbb{E} - \epsilon_0 \mathbb{E} = \mathbb{P} \times \mathbb{E} = -\frac{1}{\epsilon_0} \mathbb{E}$ (6)

① If $\text{Im} \mathbb{E}$ becomes

$$\mathbb{E} = \zeta \mathbb{E}_0 + \mathbb{P} \mathbb{P} - \mathbb{E}_0$$

$$= \zeta_0^2 \mathbb{E}_0 = (\mathbb{P}^* \mathbb{E}_0 + \mathbb{P} \mathbb{P} \mathbb{E}_0).$$

②

Combination of Eqs. (1) and ② gives

$$\mathbb{E} = \mathbb{P} \cdot \mathbb{P} \cdot \mathbb{E}_0 = \frac{\mathbb{E}_0}{\zeta}.$$

Ex. 7-14

$$\text{①} \quad \left| \frac{\text{Transmission current}}{\text{Excitation current}} \right| = \frac{I_{21}}{I_1} = \frac{I_{21} \cos \theta_2 \sin \theta_1}{I_1 \sin \theta_1 \cos \theta_2}$$

$$= \frac{I_{21}}{I_1} \tan \theta_1 \cot \theta_2.$$

② In a source-free conductor:

$$\nabla \times \mathbb{H} = \sigma \mathbb{E}, \quad \text{③}$$

$$\nabla \cdot \mathbb{E} = \rho = j\omega \epsilon \mathbb{H}, \quad \text{④}$$

$$\nabla \times \text{③} \Rightarrow \nabla \times \nabla \times \mathbb{H} = \sigma (\nabla \times \mathbb{E}) = \sigma \mathbb{P} \mathbb{H} = \sigma \nabla \times \mathbb{E}. \quad \text{⑤}$$

But $\nabla \times \mathbb{H} = 0$. If $\text{Im} \mathbb{E}$ becomes

$$\nabla^2 \mathbb{E} = -\sigma \nabla \times \mathbb{E} = 0. \quad \text{⑥}$$

Combining ⑤ and ⑥

$$\nabla^2 \mathbb{H} - \mu \sigma \nabla \times \mathbb{H} = 0.$$

Chapter 8

Plane Electromagnetic Waves

Ex. 1 In a source-free simple medium,

$$\nabla \cdot (\nabla \times \mathbf{A}) = \nabla \cdot \mathbf{B} = \nabla \cdot \mathbf{J} = \frac{\partial \rho}{\partial t} = \nabla \cdot \mathbf{E} + \epsilon \frac{\partial \rho}{\partial t} \quad (1)$$

$$\nabla \cdot (\nabla \times \mathbf{E}) = \nabla \cdot \mathbf{B} = \nabla \cdot \mathbf{J} = \nabla \cdot \mathbf{E} + \epsilon \frac{\partial \rho}{\partial t} \quad (2)$$

Substituting (1) in (2) and noting that $\nabla \cdot \mathbf{E} = 0$:

$$\epsilon \frac{\partial \rho}{\partial t} = \epsilon \frac{\partial \rho}{\partial t} + \epsilon \frac{\partial \rho}{\partial t} = 0$$

Similarly for \mathbf{H} .

Ex. 2 Assume that the velocity vector with a velocity u is in the z -direction, which is the direction of propagation of the incident wave.

$$(1) \quad \mathbf{E}_i = \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}, \quad \mathbf{E}_r = \mathbf{E}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}$$

$\mathbf{E}_i + \mathbf{E}_r = 0$ must be satisfied on reflecting surface for all t

and \mathbf{r} :

$$(u - \omega' t) = (u + \omega t)$$

$$\implies \omega' - \omega = -(\omega' + \omega) \implies -(\omega' - \omega) = -(\omega' + \omega) \implies$$

$$\implies \frac{\omega'}{\omega} = 1 - \frac{\omega'}{\omega} \left(1 + \frac{\omega'}{\omega}\right)$$

$$\implies \frac{\omega'}{\omega} = \frac{\omega'}{\omega} = \frac{1 - \frac{\omega'}{\omega}}{1 + \frac{\omega'}{\omega}} \implies \omega' = \omega \frac{1 - \frac{\omega'}{\omega}}{1 + \frac{\omega'}{\omega}} \text{ for wave}$$

$$\implies \omega' = \omega \frac{1 - \frac{\omega'}{\omega}}{1 + \frac{\omega'}{\omega}}$$

(2) For $\omega' = \omega \frac{1 - \frac{\omega'}{\omega}}{1 + \frac{\omega'}{\omega}}$ and $\omega' = \omega \frac{1 - \frac{\omega'}{\omega}}{1 + \frac{\omega'}{\omega}}$

Ex. 3 Harmonic time dependence: $e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}$

$$\text{Assume: } \mathbf{E} = \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}, \quad \mathbf{H} = \mathbf{H}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}, \text{ where } \mathbf{k} \text{ and } \mathbf{k}'$$

are constant vectors.

$$\text{Now: } \nabla \cdot (\mathbf{E} \times \mathbf{H}) = \mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)} \cdot \nabla \cdot (\mathbf{H}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{H}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)})$$

$$= \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{H}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{H}_0 e^{i(\mathbf{k}' \cdot \mathbf{r} - \omega' t)})$$

Assume

$$\text{equation: } \nabla \cdot \mathbf{E} = \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)})$$

$$\nabla \cdot \mathbf{E} = \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)})$$

$$\nabla \cdot \mathbf{E} = \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)}) = \mathbf{E}_0 \cdot \nabla \cdot (\mathbf{E}_0 e^{i(\mathbf{k} \cdot \mathbf{r} - \omega t)})$$

Ex. 2.21 Let $\vec{r} = \vec{r}_0 + \omega r^2 \cos(\omega t) \hat{i} + \vec{r}_0 + \frac{\pi}{2} \hat{j}$ (2.16).

(a) $\vec{r}_0 = \omega \sqrt{2} \hat{i} = \omega \frac{\sqrt{2}}{2} \hat{i} = \frac{\omega}{\sqrt{2}} \hat{i} = \omega \cos \omega t \hat{i}$
 $\omega = \frac{10}{\sqrt{2}} \hat{i} = 5\sqrt{2} \hat{i}$

At $t = \pi/2 \omega = \pi$, we require the argument of vector to be
 $\omega^2 \cos \omega t = \frac{\omega}{\sqrt{2}} \cos \pi = \frac{\omega}{\sqrt{2}} = \frac{\omega}{\sqrt{2}} = \omega \cos \omega t$
 $\rightarrow y = \omega \cos \omega t = \omega \cos \pi = -\omega \cos \omega t$ (2.17)

Ex. 2.22 Show: $\vec{r} = \vec{r}_0 e^{i\omega t} + \vec{r}_1 e^{-i\omega t}$ (2.18)

(a) $\omega = \omega e^{i\omega t}$ (radius) $\rightarrow \vec{r} = \omega e^{i\omega t} \hat{i} = \omega \cos \omega t \hat{i}$

$\vec{r} = \omega \cos \omega t \hat{i} \rightarrow \omega = \frac{10}{\sqrt{2}} \hat{i} = 5\sqrt{2} \hat{i}$

(b) $\omega = \frac{10}{\sqrt{2}} \hat{i} \rightarrow \omega = \left(\frac{10}{\sqrt{2}}\right)^2 = 50$

(c) Left-hand elliptically polarized.

(d) $\vec{r} = \frac{1}{\sqrt{2}} \hat{i} - \frac{1}{\sqrt{2}} \hat{j} = \frac{1}{\sqrt{2}} \hat{i}$ (2.19)

$\vec{r} = \frac{1}{\sqrt{2}} \hat{i}, \vec{r} = \frac{1}{\sqrt{2}} \hat{i} + \frac{1}{\sqrt{2}} \hat{j} e^{i\omega t} - \frac{1}{\sqrt{2}} \hat{j} e^{-i\omega t}$

$\vec{r}(t) = \frac{1}{\sqrt{2}} \hat{i} \left[\frac{1}{\sqrt{2}} \cos(\omega t) - \frac{1}{\sqrt{2}} \cos(\omega t) + \frac{1}{\sqrt{2}} \sin(\omega t) + \frac{1}{\sqrt{2}} \sin(\omega t) \right]$ (2.20)

Ex. 2.23 Let $\vec{r} = \vec{r}_0 \cos \omega t + \vec{r}_1 \sin \omega t$ in $\vec{r}_0 = \omega \hat{i}, \vec{r}_1 = \omega \hat{j}$

$\frac{d\vec{r}}{dt} = \omega \hat{j} \cos \omega t - \omega \hat{i} \sin \omega t = \omega \cos \omega t \hat{j} - \omega \sin \omega t \hat{i}$
 $\frac{d\vec{r}}{dt} = \omega \cos \omega t \hat{j} + \omega \sqrt{1 - \cos^2 \omega t} \hat{i}$

$\left(\frac{d\vec{r}}{dt} - \omega \cos \omega t \hat{j} \right) = \left(\omega \sqrt{1 - \cos^2 \omega t} \right) \hat{i}$

$\left(\frac{d\vec{r}}{dt} - \omega \cos \omega t \hat{j} \right)^2 = \left(\omega \sqrt{1 - \cos^2 \omega t} \right)^2 = \omega^2 (1 - \cos^2 \omega t)$ (2.21)

which is the equation of an ellipse. In order to find the parameters of the polarization ellipse, rotate the coordinate axes by any counterclockwise by an angle θ to ωt . Assume the equation of this ellipse in terms of the new coordinates to be

$\left(\frac{x'}{a} \right)^2 + \left(\frac{y'}{b} \right)^2 = 1$ (2.22)



$$\text{where } E_x = E_0 \cos \theta - E_1 \sin \theta, \quad \textcircled{1}$$

$$\text{and } E_y = -E_0 \sin \theta + E_1 \cos \theta, \quad \textcircled{2}$$

Substituting $\textcircled{1}$ and $\textcircled{2}$ in $\textcircled{3}$ and rearranging

$$E_0 \left(\frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} \right) + E_1 \left(\frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} \right) = E_0 \mu_1 \cos \theta \left(\frac{1}{\mu_1} - \frac{1}{\mu_2} \right) + E_1 \quad \textcircled{3}$$

Comparing $\textcircled{1}$ and $\textcircled{3}$, we obtain

$$\begin{cases} \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_0 + \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_1 = \frac{E_0 \mu_1 \cos \theta}{\mu_1} \\ \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_0 + \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_1 = \frac{E_0 \mu_1 \sin \theta}{\mu_2} \end{cases} \quad \textcircled{4}$$

$$\begin{cases} \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_0 + \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_1 = \frac{E_0 \mu_1 \cos \theta}{\mu_1} \\ \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_0 + \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_1 = \frac{E_0 \mu_1 \sin \theta}{\mu_2} \end{cases} \quad \textcircled{5}$$

$$\begin{cases} \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_0 + \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_1 = \frac{E_0 \mu_1 \cos \theta}{\mu_1} \\ \frac{\mu_1 \mu_2}{\mu_2^2 - \mu_1^2} E_0 + \frac{\mu_1 \mu_2}{\mu_1^2 - \mu_2^2} E_1 = \frac{E_0 \mu_1 \sin \theta}{\mu_2} \end{cases} \quad \textcircled{6}$$

Eqs. $\textcircled{4}$, $\textcircled{5}$, and $\textcircled{6}$ can be solved for three unknowns:

$$\theta = \frac{1}{2} \tan^{-1} \left(\frac{2\mu_1 \mu_2 \sin 2\alpha}{\mu_1^2 - \mu_2^2} \right),$$

$$a = \sqrt{\frac{(\mu_1^2 - \mu_2^2) \cos 2\alpha}{(\mu_1^2 - \mu_2^2) \cos 2\alpha + \mu_1^2 \sin^2 2\alpha}} \sin \theta,$$

$$b = \sqrt{\frac{(\mu_1^2 - \mu_2^2) \sin 2\alpha}{(\mu_1^2 - \mu_2^2) \cos 2\alpha + \mu_1^2 \sin^2 2\alpha}} \sin \theta.$$

In particular, if $E_0 = E_1 = E_1$, $\mu_1 = \mu_2 = \mu$, $\mu_1 > \mu_2$, $\mu_1 < \mu_2$, $\mu_1 < \mu_2$.

Ex. 11 Let an elliptically polarized plane wave be represented by the phasor (with propagation factor $e^{i(kz - \omega t)}$ omitted)

$$\mathbf{E} = E_x \hat{x} + E_y \hat{y} e^{i\alpha}$$

where E_x , E_y , and α are arbitrary constants.

Right-hand circularly polarized wave: $E_x = E_0 \cos(\omega t - kz)$

Left-hand circularly polarized wave: $E_x = E_0 \sin(\omega t - kz)$

$$\text{If } E_x = \frac{1}{2}(E_0 + jE_0 e^{i\alpha}) \text{ and } E_y = \frac{1}{2}(E_0 + jE_0 e^{i\alpha}),$$

$$\text{then } \mathbf{E} = E_0 + E_0.$$

$$\text{(a) Right-hand circularly polarized wave: } E_x = E_0 \cos(\omega t - kz)$$

$$= E_0 \frac{1}{2}(e^{j(\omega t - kz)} + e^{-j(\omega t - kz)}) + E_0 \frac{j}{2}(e^{j(\omega t - kz)} - e^{-j(\omega t - kz)})$$

$$= E_0 + E_0, \text{ where } E_0 \text{ and } E_0 \text{ are}$$

right-hand and left-hand circularly polarized waves respectively.

(b) Left-hand circularly polarized wave: $E_x = E_0 \sin(\omega t - kz)$

$$= E_0 \frac{1}{2}(e^{j(\omega t - kz)} - e^{-j(\omega t - kz)}) + E_0 \frac{j}{2}(e^{j(\omega t - kz)} + e^{-j(\omega t - kz)})$$

$$= E_0 + E_0.$$

Ex.11 For calculating matrix: $\lambda_1 = \beta - j\alpha$.

$$\begin{aligned} \lambda_1^2 &= \beta^2 - \alpha^2 - 2j\alpha\beta \\ &= \alpha^2 j\alpha\beta = \alpha^2 j\alpha\beta (1 - j\frac{\beta}{\alpha}) \end{aligned}$$

$$\therefore \beta^2 - \alpha^2 = \alpha^2 (\lambda_1^2) = \alpha^2 j\alpha\beta \quad \text{--- (1)}$$

$$\beta + \alpha^2 = |\lambda_1^2| = \alpha^2 j\alpha\beta \sqrt{1 + (\frac{\beta}{\alpha})^2} \quad \text{--- (2)}$$

From (1) and (2) we obtain

$$\alpha = \alpha^2 j\alpha\beta \left[\sqrt{1 + (\frac{\beta}{\alpha})^2} - 1 \right]^{-1/2}, \quad \beta = \alpha^2 j\alpha\beta \left[\sqrt{1 + (\frac{\beta}{\alpha})^2} + 1 \right]^{-1/2}$$

Ex.12 All three matrices are good candidates, $\frac{d(\lambda)}{dt} = 0$.

$$\lambda_1 = \alpha + j\beta, \quad \lambda_2 = \alpha - j\beta, \quad \lambda_3 = (\alpha + j\beta)^2$$

(i) $f = \sin(\lambda t)$

	λ_1 (1)	λ_2 (1)	λ_3 (1)	f' (1)
Upper	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$
Middle	$\alpha + j\beta \sin(\lambda t)$	$\alpha - j\beta \sin(\lambda t)$	$\alpha + j\beta \sin(\lambda t)$	$\alpha \sin(\lambda t)$
Lower	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$

(ii) $f = t \cos(\lambda t)$

	λ_1 (1)	λ_2 (1)	λ_3 (1)	f' (1)
Upper	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$
Middle	$\alpha + j\beta \sin(\lambda t)$	$\alpha - j\beta \sin(\lambda t)$	$\alpha + j\beta \sin(\lambda t)$	$\alpha \sin(\lambda t)$
Lower	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$

(iii) $f = t^2 \cos(\lambda t)$

	λ_1 (1)	λ_2 (1)	λ_3 (1)	f' (1)
Upper	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$
Middle	$\alpha + j\beta \sin(\lambda t)$	$\alpha - j\beta \sin(\lambda t)$	$\alpha + j\beta \sin(\lambda t)$	$\alpha \sin(\lambda t)$
Lower	$\alpha + j\beta \cos(\lambda t)$	$\alpha - j\beta \cos(\lambda t)$	$\alpha + j\beta \cos(\lambda t)$	$\alpha \cos(\lambda t)$

Beispiel $f = 2 \cos^2(\theta)$, $\theta = \frac{1}{2}$, $\tan \zeta = \frac{f}{\theta} = 4$

4) $\log_2(\cos \theta) = \log_2 \frac{\sqrt{1-f^2}}{2} = \frac{1}{2} \log_2 \frac{1-f^2}{4} = \log_2 \sqrt{1-f^2} - 1$

$e^{\log_2 \cos \theta} = \frac{1}{2} \sqrt{1-f^2} \implies x = \frac{1}{2} \sqrt{1-f^2} = 0.391 \text{ cm}$

4) $\log_2(\cos \theta) = \frac{1}{2} \log_2 \frac{1-f^2}{4} = \log_2 \sqrt{1-f^2} - 1$

$\log_2(\cos \theta) = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$x = \frac{1}{2} \sqrt{1-f^2} = 0.391 \text{ cm}$

$y_1 = \frac{1}{2} \sqrt{1-f^2} = 0.391 \text{ cm}$

$y_2 = \frac{1}{2} \sqrt{1-f^2} = 0.391 \text{ cm}$

4) $\log_2 \cos \theta = \frac{1}{2} \log_2 \frac{1-f^2}{4}$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \frac{1-f^2}{4} = \log_2 \sqrt{1-f^2} - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \frac{1-f^2}{4} = \log_2 \sqrt{1-f^2} - 1$

Beispiel $\log_2 \cos \theta = \frac{1}{2} \log_2 \frac{1-f^2}{4} = \log_2 \sqrt{1-f^2} - 1$

4) $\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

4) $\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

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$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

Beispiel 4) $\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

4) $\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

$\log_2 \cos \theta = \frac{1}{2} \log_2 \left[1 - \left(\frac{f}{2}\right)^2 \right] - 1$

Ex 11



Assume the interface to be stratified into layers having thickness dx

$$n_1 \neq n_2 \neq n_3 \neq \dots \neq n_{10}$$

The corresponding equivalent permittivities of the layers are:

$$\epsilon_1 = \epsilon_2 \left(1 - \frac{dx}{r}\right) \text{ with } r = \frac{1}{\sin^2 \theta_i} \frac{dx}{dn_1^2}$$

$$\text{and } \epsilon_2 = \epsilon_3 > \epsilon_4 > \epsilon_5 > \dots > \epsilon_{10} \left(\frac{dx}{n_{10}^2}\right)$$

From Snell's law of refraction

$$\sin \theta_1 = \sin \theta_2 \sqrt{\epsilon_2 / \epsilon_1} = \sin \theta_3 \sqrt{\epsilon_3 / \epsilon_2} \sqrt{\epsilon_1 / \epsilon_2}$$

$$\sin \theta_2 = \sin \theta_3 \sqrt{\epsilon_3 / \epsilon_2} = \sin \theta_4 \sqrt{\epsilon_4 / \epsilon_3}$$

$$\sin \theta_3 = \sin \theta_4 \sqrt{\epsilon_4 / \epsilon_3} = \sin \theta_5 \sqrt{\epsilon_5 / \epsilon_4}$$

For total reflection at the layer with n_{10} , the angle of refraction $\theta_{10} = 90^\circ$, and $\sin \theta_{10} = 1 = \sin \theta_9 \sqrt{\epsilon_{10} / \epsilon_9}$

$$\therefore \theta_{10} = \theta_9 \left(1 - \frac{dx}{r_{10}}\right)^{1/2} = \theta_9 \sin \theta_9$$

$$\therefore \theta = r_{10} \sin^2 \theta = r \frac{\epsilon_{10}}{\epsilon_9} \sin^2 \theta_9$$

Ex 12 a) From Eq (10.21): $n_2 = \frac{c}{v_2} = \frac{c}{v_1} \left(\frac{v_1}{v_2}\right) = n_1 \sqrt{\frac{\epsilon_2}{\epsilon_1}}$

$$\text{If } n_1 = \frac{c}{v_1} = \frac{c}{v_2} = \frac{c}{v_3} = \frac{c}{v_4}$$

$$v_2 = v_3 = v_4 \left(\frac{\epsilon_2}{\epsilon_1}\right)^{1/2} = v_1 \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

Ex 13 $\epsilon_{10} = 10 \epsilon_0$, $n = 10$ (vacuum)

a) $|E| = \sqrt{2000} = 44.72$ (volts/meter)

$|H| = 10 \sqrt{20} = 44.72$ (volts/meter)

b) $\epsilon_{10} = 10 \epsilon_0$, $n = 10$ (vacuum)

$|E| = 44.72$ (volts/meter), $|H| = 4.472$ (volts/meter)

Ex. 12 Assume a uniformly polarized plane sheet:

$$\vec{D}(x, y) = \epsilon_0 \epsilon_p \cos(\omega t - k_x x) \hat{y} + \epsilon_0 \epsilon_p \sin(\omega t - k_x x) \hat{z}$$

$$\vec{P}(x, y) = \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \cos(\omega t - k_x x) \hat{y} - \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \sin(\omega t - k_x x) \hat{z}$$

Applying Gauss, $\vec{D} = \vec{E} + \vec{P} = \epsilon_0 \frac{\epsilon_p}{\epsilon_0} [\cos(\omega t - k_x x) \hat{y} + \sin(\omega t - k_x x) \hat{z}]$
 $= \epsilon_0 \frac{\epsilon_p}{\epsilon_0} \hat{r}$; ϵ_0 not a function independent of ϵ_p and ω

Ex. 13 $\vec{E} = E_0 \hat{x}_p + E_0 \hat{x}_y$

$$\vec{P} = \frac{1}{2} \epsilon_0 \hat{x}_p + \frac{1}{2} \epsilon_0 (\hat{x}_p \hat{x}_y - \hat{x}_y \hat{x}_p)$$

$$\vec{D}_{\text{ext}} = \frac{1}{2} \epsilon_0 (\vec{E} + \vec{P}) = \epsilon_0 \frac{1}{2} (\hat{x}_p + \hat{x}_y)$$

Ex. 14 From Gauss' law, $\vec{E} = \frac{1}{2} \frac{\rho}{\epsilon_0} \hat{r}$ where ρ is the free charge density on the inner conductor.

$$V_0 = - \int \vec{E} \cdot d\vec{s} = \frac{\rho}{2\epsilon_0} \ln\left(\frac{b}{a}\right) \implies \vec{E} = \epsilon_0 \frac{2V_0}{\ln(b/a)} \hat{r}$$

From Ampere's circuital law, $\vec{H} = \epsilon_0 \frac{2V_0}{\ln(b/a)} \hat{\phi}$

Applying vector, $\vec{D} = \vec{E} + \vec{P} = \epsilon_0 \frac{2V_0}{\ln(b/a)} \hat{r}$

Power transmitted in our cross-sectional area:

$$P = \int \vec{D} \cdot d\vec{s} = \frac{2V_0^2}{\ln(b/a)} \int_0^{2\pi} \int_a^b \left(\frac{1}{r}\right) r dr d\phi = V_0^2 \frac{2\pi}{\ln(b/a)}$$

Ex. 15 a) $\vec{E} = \frac{1}{\sqrt{2}} \frac{V_0}{r} \hat{r}$; $\vec{P} = \epsilon_0^2 \frac{V_0^2}{r^2} \hat{r}$

b) $\vec{D}(x, y) = \epsilon_0 \epsilon_p e^{-\alpha x} \cos(\omega t - \beta y) \hat{z}$

$$\vec{E}_p = (1 - \epsilon_p) \frac{V_0}{2} = (1 - \epsilon_p) \frac{V_0}{2} = \frac{\epsilon_p}{2} V_0 e^{-\alpha x}$$

$$\vec{E}(x, y) = \epsilon_0 \frac{\epsilon_p}{2} V_0 e^{-\alpha x} \cos(\omega t - \beta y) \hat{z} = \frac{V_0}{2} \hat{z}$$

c) $\vec{D}_{\text{ext}} = \frac{1}{2} \epsilon_0 (\vec{E} + \vec{P}) = \epsilon_0 \frac{1}{2} \frac{\epsilon_p}{2} V_0^2 \cos \frac{\pi}{2}$
 $= \epsilon_0 \frac{1}{2} \left(\frac{V_0^2}{2}\right)$ (value)

Ex 2.1 Given $\vec{r}_1 = r_1(\hat{r}_1 + \hat{r}_2) e^{i\omega t}$

a) Assume reduced $\vec{r}_1(t) = (r_1 \hat{r}_1 + r_2 \hat{r}_2) e^{i\omega t}$

Boundary condition at $t=0$: $\vec{r}_1(0) = \vec{r}_2(0) = 0$

$\implies \vec{r}_1(0) = r_1(-\hat{r}_1 + \hat{r}_2) e^{i\omega t} = 0$ as both have circularly polarized waves in same direction.

b) $\vec{r}_1 = (r_1 - r_2) = 0 \implies r_1 = (r_1 \cos(\omega t) + r_2 \sin(\omega t)) = 0$ ($r_1 \cos(\omega t) = r_2 \sin(\omega t)$)

$\vec{r}_1 \cos(\omega t) = r_1 \cos(\omega t) = r_2 \sin(\omega t)$, $\vec{r}_1 \sin(\omega t) = r_1 \sin(\omega t) = r_2 \cos(\omega t)$

$\vec{r}_1 \cos(\omega t) = r_2 \sin(\omega t) = r_2 \cos(\omega t)$

$r_1 = -r_2 + r_2 \cos(\omega t) = \frac{r_2}{2}(1 - \cos(\omega t))$

c) $\vec{r}_1 \sin(\omega t) = r_2 (\vec{r}_1 \cos(\omega t) + \vec{r}_1 \sin(\omega t)) e^{i\omega t}$

$= r_2 r_1 (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t} = r_2 r_1 (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$

$= r_2 r_1 (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$

$= 2 r_2 r_1 \sin(\omega t) (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t))$

Ex 2.2 Given $\vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t}$ ($r_1 \cos(\omega t)$)

a) $r_1 = 0$, $r_2 = 0 \implies r_1 = r_2 = \sqrt{r_1^2 + r_2^2} = 0$ ($r_1 \cos(\omega t)$)

$r_1 = 2r_2 \cos(\omega t) = 2r_2 \cos(\omega t) = 2r_2 \cos(\omega t)$ ($r_1 \cos(\omega t) = 2r_2 \cos(\omega t)$)

b) $\vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t} = r_1 \cos(\omega t) e^{i\omega t}$ ($r_1 \cos(\omega t)$)

$\vec{r}_1(r_1, \omega t) = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$ ($r_1 \cos(\omega t) = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$)

$= \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t} = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$

$\vec{r}_1(r_1, \omega t) = (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t} = r_1 \cos(\omega t) e^{i\omega t}$ ($r_1 \cos(\omega t)$)

c) $\cos(\omega t) = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t} \implies r_1 = 2 \cos(\omega t) e^{i\omega t} = r_1 \cos(\omega t)$

d) $\vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t} \implies \vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t}$

$\vec{r}_1(r_1, \omega t) = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$ ($r_1 \cos(\omega t) = \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$)

$= \frac{r_1}{2} (\hat{r}_1 \cos(\omega t) + \hat{r}_2 \sin(\omega t)) e^{i\omega t}$

e) $\vec{r}_1(r_1, \omega t) = \vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t} = r_1 \cos(\omega t) e^{i\omega t}$

$= r_1 \cos(\omega t) e^{i\omega t} = r_1 \cos(\omega t) e^{i\omega t}$

$\vec{r}_1(r_1, \omega t) = \vec{r}_1(r_1, \omega t) = r_1 \cos(\omega t) e^{i\omega t} = r_1 \cos(\omega t) e^{i\omega t}$ ($r_1 \cos(\omega t)$)

Ex 11.12 Given $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ (Correct)

(i) $f_1(x) = \sin x$, $f_2(x) = \cos x$ $\implies \mathcal{L} = \sqrt{100^2 + 10} = 10$ (Correct)

$\mathcal{L} = \sin 10x + \cos 10x = 10 \implies \sin 10x + \cos 10x = 10$ (Correct)

(ii) $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10} = \mathcal{L}\{f_1(x) + f_2(x)\}$ (Correct)

$$\mathcal{L}\{f(x)\} = \frac{1}{10} \mathcal{L}\{f_1(x) + f_2(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x)\} + \mathcal{L}\{f_2(x)\}) e^{200x-10}$$

$$= \frac{1}{10} (\mathcal{L}\{f_1(x)\} + \mathcal{L}\{f_2(x)\}) e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x)\} + \mathcal{L}\{f_2(x)\}) e^{200x-10} \implies \mathcal{L}\{f(x)\} = 10$$
 (Correct)

(iii) $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10} \implies \mathcal{L}\{f(x)\} = 10$ (Correct)

(iv) $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ and $\mathcal{L}\{f_1(x)\} = \mathcal{L}\{f_2(x)\} = 10$ (Correct)

$$\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10} = 10$$
 (Correct)

$$\mathcal{L}\{f(x)\} = \frac{1}{10} \mathcal{L}\{f_1(x) + f_2(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x)\} + \mathcal{L}\{f_2(x)\}) e^{200x-10}$$

$$= \frac{1}{10} (\mathcal{L}\{f_1(x)\} + \mathcal{L}\{f_2(x)\}) e^{200x-10} \implies \mathcal{L}\{f(x)\} = 10$$
 (Correct)

(v) $\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$ and $\mathcal{L}\{f_1(x)\} = \mathcal{L}\{f_2(x)\} = 10$ (Correct)

$$\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10} = 10$$
 (Correct)

Ex 11.13 (i) From Eqn (11-110) and (11-109):

$$\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x) + f_2(x)\}) e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x) + f_2(x)\}) e^{200x-10}$$

Ex 11.14 (i) From Eqn (11-110) and (11-109):

$$\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x) + f_2(x)\}) e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x) + f_2(x)\}) e^{200x-10}$$

Ex 11.15 For normal incidence: $\Gamma = 1$, where $\Gamma \leq 1$.

$$\mathcal{L}\{f(x)\} = \mathcal{L}\{f_1(x) + f_2(x)\} e^{200x-10}$$

$$\mathcal{L}\{f(x)\} = \frac{1}{10} (\mathcal{L}\{f_1(x) + f_2(x)\}) e^{200x-10}$$

Ex 11) In the decay reaction (Exercise 10):

$$E_1 = E_2 E_3 e^{-\gamma} \sqrt{1 - \beta^2}$$

where from Exercise 10, $E_2 = \frac{m_0 c^2}{\sqrt{1 - \beta^2}} \left[\sqrt{1 - \frac{v^2}{c^2}} - 1 \right]$, $E_3 = \frac{m_0 c^2}{\sqrt{1 - \beta^2}} \left[\sqrt{1 - \frac{v^2}{c^2}} + 1 \right]$

Given: $\beta = 0.7$ (initial) $\Rightarrow \gamma = 1.40028$ (initial)

So $E_2 = \frac{E_0}{\sqrt{1 - \beta^2}} = 1.40028 E_0$ \Rightarrow $E_2 = 1.40028 E_0$ (initial) $E_3 = 2.80056 E_0$ (initial)

$E_1 = \sqrt{E_2 E_3} = \sqrt{1.40028 \times 2.80056} E_0 = 2.00014 E_0$ (initial)

$E_1 = E_2 E_3 e^{-\gamma} \sqrt{1 - \beta^2}$, $E_2 = E_3 \left(\frac{E_1}{E_2 E_3} \right) = E_3 \frac{E_1}{E_2 E_3} e^{-\gamma} \sqrt{1 - \beta^2} e^{\gamma} \sqrt{1 - \beta^2}$

So $E_2 = E_3 \frac{E_1}{E_2 E_3} e^{-\gamma} \sqrt{1 - \beta^2} \Rightarrow E_2 = E_3 \frac{E_1}{E_2 E_3} e^{-\gamma} \sqrt{1 - \beta^2}$

Therefore we have: $\begin{cases} E_1 = E_2 E_3 e^{-\gamma} \sqrt{1 - \beta^2} \\ E_2 = E_3 \frac{E_1}{E_2 E_3} e^{-\gamma} \sqrt{1 - \beta^2} \end{cases}$
 $\Rightarrow E_2 = 2.00014 E_0 e^{-\gamma} \sqrt{1 - \beta^2}$, $E_3 = 2.80056 E_0 e^{-\gamma} \sqrt{1 - \beta^2}$

- $E_1(0.9) = E_2(0.9) E_3(0.9) e^{-\gamma(0.9)} \sqrt{1 - \beta(0.9)^2}$ (initial)
- $E_2(0.9) = \frac{E_1(0.9)}{E_3(0.9)} e^{\gamma(0.9)} \sqrt{1 - \beta(0.9)^2} = 1.40028 E_0 e^{\gamma(0.9)} \sqrt{1 - \beta(0.9)^2}$ (initial)
- $E_3(0.9) = \frac{E_1(0.9)}{E_2(0.9)} e^{\gamma(0.9)} \sqrt{1 - \beta(0.9)^2} = 2.80056 E_0 e^{\gamma(0.9)} \sqrt{1 - \beta(0.9)^2}$ (initial)
- $E_1(0.9) = E_2(0.9) E_3(0.9) e^{-\gamma(0.9)} \sqrt{1 - \beta(0.9)^2} = 2.00014 E_0 e^{-\gamma(0.9)} \sqrt{1 - \beta(0.9)^2}$ (initial)

• $E_1(0.7) = E_2(0.7) E_3(0.7) e^{-\gamma(0.7)} \sqrt{1 - \beta(0.7)^2} = 2.00014 E_0$ (initial)

$E_2(0.7) = E_3(0.7) \frac{E_1(0.7)}{E_2(0.7) E_3(0.7)} e^{\gamma(0.7)} \sqrt{1 - \beta(0.7)^2} = 2.00014 E_0 e^{\gamma(0.7)} \sqrt{1 - \beta(0.7)^2}$ (initial)

Ex 12) $r = \frac{E_1}{E_2} = \frac{E_3 \sqrt{E_1}}{E_2 \sqrt{E_1}} = \frac{E_3 \sqrt{E_1}}{E_2 \sqrt{E_1}}$

• $r = \left| \frac{E_3 \sqrt{E_1}}{E_2 \sqrt{E_1}} \right| = \left| \frac{1 + \beta \sqrt{E_1}}{1 - \beta \sqrt{E_1}} \right|$
 $= (1 + \beta \sqrt{E_1}) - 1 + \beta \sqrt{E_1} = 1 + 2\beta \sqrt{E_1}$

Function of power absorbed, $P = r - 1 = 2\beta \sqrt{E_1}$
 $= 2\beta \sqrt{E_1}$

• $4 = 2\beta \sqrt{E_1}$ (initial) For $\beta = 0.9$ (initial) $E_1 = 2$
 $P = 4 - 1 = 3$, or 3000 W

Ex 10.12 From Eqs. (8-104) through (8-106)

$$E_1 = E_0 \left[\frac{1}{2} (1 + \cos 2\theta) + \frac{1}{2} \cos 4\theta \right], \quad E_2 = E_0 \left[\frac{1}{2} (1 + \cos 2\theta) - \frac{1}{2} \cos 4\theta \right],$$

$$E_3 = E_0 \left[\frac{1}{2} (1 - \cos 2\theta) + \frac{1}{2} \cos 4\theta \right], \quad E_4 = E_0 \left[\frac{1}{2} (1 - \cos 2\theta) - \frac{1}{2} \cos 4\theta \right],$$

$$E_5 = E_0 \cos^2 \theta,$$

$$E_6 = E_0 \sin^2 \theta.$$

Boundary conditions: at $\theta = 0$: $E_1(0) = E_2(0)$, $E_3(0) = E_4(0)$.

at $\theta = \pi$: $E_1(\pi) = E_2(\pi)$, $E_3(\pi) = E_4(\pi)$.

Four equations to solve for unknowns E_1 , E_2 , E_3 , and E_4 in terms of E_0 .

$$(1) \quad E_1 = \frac{E_0(1 + \cos 2\theta + \cos 4\theta)}{2(1 + \cos 2\theta + \cos 4\theta)} E_0, \quad \text{where}$$

$$E_1 = \frac{E_0(1 + \cos 2\theta + \cos 4\theta)}{2(1 + \cos 2\theta + \cos 4\theta)} E_0, \quad E_2 = \frac{E_0(1 + \cos 2\theta - \cos 4\theta)}{2(1 + \cos 2\theta - \cos 4\theta)} E_0,$$

$$E_3 = \frac{E_0(1 - \cos 2\theta + \cos 4\theta)}{2(1 - \cos 2\theta + \cos 4\theta)} E_0, \quad E_4 = \frac{E_0(1 - \cos 2\theta - \cos 4\theta)}{2(1 - \cos 2\theta - \cos 4\theta)} E_0,$$

$$E_5 = \frac{E_0 \cos^2 \theta}{\cos^2 \theta} E_0, \quad E_6 = \frac{E_0 \sin^2 \theta}{\sin^2 \theta} E_0.$$

(2) If $\theta = \pi/4$: $E_1 = E_2$, $E_3 = \frac{E_0(1 + \sqrt{2})}{2} E_0$.

$\therefore P = \frac{E_0^2}{2} \cos^2 \theta$ and $\sin^2 \theta$ in brackets.

If $\theta = \pi/4$, $\cos^2 \theta = \sin^2 \theta = 1/2$, $\cos 4\theta = 1$, $\sin 4\theta = 0$.

Ex 10.13 From Example 8-11: $E_1 = \sqrt{2} E_0$, $E_2 = \sqrt{2} E_0$, $E_3 = E_0$.

(a) Maximum of reflected: $E_1 = \frac{E_0}{\sqrt{2}} = 0.707 E_0$ and $E_2 = 0.707 E_0$.

(b) For total light: $E_1 = \frac{E_0}{\sqrt{2}} = 0.707 E_0$.

$$E_2 = 0.707 E_0 \rightarrow R = 0.500 E_0.$$

From Eq. (8-103) and using boundary conditions with respect

$$\text{to } E_1 = E_2: \quad E_1(0) = E_2(0) = \frac{E_0(1 + \cos 2\theta + \cos 4\theta)}{2(1 + \cos 2\theta + \cos 4\theta)} = \frac{E_0(1 + \cos 2\theta + \cos 4\theta)}{2(1 + \cos 2\theta + \cos 4\theta)}$$

$$P = \frac{E_0^2}{2} \cos^2 \theta = 0.500 E_0^2.$$

Percentage of power reflected = $100 \times 0.500 E_0^2$
 $= 50.0\%$

$$\text{E.10.11} \quad C = \frac{2\sqrt{2} - \sqrt{2}}{2\sqrt{2} + \sqrt{2}} \cdot 2\sqrt{2} = 2 \frac{2\sqrt{2} - \sqrt{2}}{2\sqrt{2} + \sqrt{2}}$$

$$C = \frac{2 - 1}{2 + 1} \longrightarrow \frac{1}{3} = \frac{1 - \sqrt{2}}{1 + \sqrt{2}}$$

$$C = \frac{2 - \sqrt{2}}{2 + \sqrt{2}} \longrightarrow \frac{1}{3} = \frac{1 - \sqrt{2}}{1 + \sqrt{2}}$$

$$C = \frac{1 + \sqrt{2} \frac{2 - \sqrt{2}}{2 + \sqrt{2}}}{1 + \sqrt{2} \frac{2 - \sqrt{2}}{2 + \sqrt{2}}} = \frac{1}{3} \frac{(2 - \sqrt{2})(1 + \sqrt{2})}{(2 + \sqrt{2})(1 + \sqrt{2})}$$

$$= \frac{1 + \sqrt{2} \frac{2 - \sqrt{2}}{2 + \sqrt{2}}}{1 + \sqrt{2} \frac{2 - \sqrt{2}}{2 + \sqrt{2}}} = \frac{1}{3} \frac{(2 - \sqrt{2}) + (2 + \sqrt{2})}{(2 + \sqrt{2}) + (2 - \sqrt{2})}$$

$$= \frac{1(2 + 2) + \sqrt{2}(-\sqrt{2} + \sqrt{2})}{1(2 + 2) + \sqrt{2}(-\sqrt{2} + \sqrt{2})}$$

$$\text{E.10.12} \quad \vec{L}_1 = L_1 (a_1 e^{i\omega t} + c_1 e^{-i\omega t}),$$

$$\vec{L}_2 = L_2 \frac{1}{2} (a_2 e^{i\omega t} + c_2 e^{-i\omega t}),$$

$$\vec{L}_3 = L_3 (a_3' e^{i\omega t} + c_3' e^{-i\omega t}),$$

$$\vec{L}_4 = L_4 \frac{1}{2} (a_4 e^{i\omega t} + c_4 e^{-i\omega t}).$$

$$\text{At } t = 0, \vec{L}_2 = 0 \longrightarrow \vec{L}_3 = -a_3' e^{-i\omega t}.$$

$$\vec{L}_1 = L_1 a_1' (e^{i\omega t} - e^{-i\omega t + i\pi}).$$

$$\vec{L}_4 = L_4 \frac{1}{2} (a_4 e^{i\omega t} + e^{i\omega t + i\pi}).$$

$$\text{Secondary relations: } L_1 \omega = L_2 \omega \longrightarrow L_1 = L_2 = L_3' (1 - e^{-i\omega t}),$$

$$L_4 \omega = L_3 \omega \longrightarrow L_4 = L_3 = L_3' \frac{1}{2} (1 + e^{-i\omega t}).$$

$$L_3' = \frac{2L_1 L_4}{L_1 + L_4 + L_3 + L_2}.$$

$$L_3 = \frac{(L_1 - L_4)(2L_1 L_4)}{L_1 + L_4 + L_3 + L_2}.$$

$$\Rightarrow \mathcal{L}\{x_1(t)\} = \mathcal{L}\{x_2(t) + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\} = \frac{1}{s} - \frac{\omega}{2} \frac{1 + \cos 2\omega t}{s}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\{x_2(t) + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t) + \frac{\omega}{2}(1 + \cos 2\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\left\{\frac{\sin 2\omega t}{2\omega} + \frac{\cos 2\omega t}{2\omega} - \frac{\sin \omega t}{\omega} + \frac{\cos \omega t}{\omega} - \frac{\omega}{2}(1 + \cos 2\omega t) + \frac{\omega}{2}(1 + \cos 2\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\right\}$$

$$F = \mathcal{L}\left\{\frac{\sin 2\omega t}{2\omega} + \frac{\cos 2\omega t}{2\omega} - \frac{\sin \omega t}{\omega} + \frac{\cos \omega t}{\omega} - \frac{\omega}{2}(1 + \cos 2\omega t) + \frac{\omega}{2}(1 + \cos 2\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\right\}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \frac{1}{2} \mathcal{L}\{1 + \cos 2\omega t\} = 0$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = 0$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\{x_2(t) + \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)\} = 0 \Rightarrow x_2(t) = \sin(\omega t) - \frac{\omega}{2}(1 + \cos 2\omega t)$$

$$\underline{\text{Ex 2.11}} \quad x_1 = x_2 + x_3 = (1 + \cos t) \frac{1}{t}, \quad x_2 + x_3 = \frac{1}{t} + \frac{\cos t}{t}$$

$$x_2 = (1 + \cos t) \frac{1}{t} - x_3 \quad \text{or } \mathcal{L}\{x_2(t)\}$$

a) From Problem 2.8.14

$$\mathcal{L}\left\{\frac{1}{t}\right\} = \int_0^\infty \frac{e^{-st}}{t} dt = -\gamma - \ln s$$

$$\Rightarrow \mathcal{L}\left\{\frac{\cos t}{t}\right\} = \int_0^\infty \frac{e^{-st} \cos t}{t} dt = -\gamma - \ln \sqrt{s^2 + 1}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \mathcal{L}\left\{\frac{1}{t} + \frac{\cos t}{t}\right\} = -\gamma - \ln s - \gamma - \ln \sqrt{s^2 + 1}$$

$$\Rightarrow \mathcal{L}\{x_2(t)\} = \frac{1}{s} + \frac{\ln \sqrt{s^2 + 1}}{s}$$

$$= \left(1 + \frac{\ln \sqrt{s^2 + 1}}{s}\right) \frac{1}{s}$$

$$\mathcal{L}\{x_2(t)\} = \frac{1}{s} \mathcal{L}\{1 + \cos t\} = \mathcal{L}\left\{\frac{1}{t} + \frac{\cos t}{t}\right\}$$

$$= \frac{1}{s} \frac{1}{\sqrt{s^2 + 1}} (1 + \cos t)$$

$$\text{where } \frac{1}{s} \frac{1}{\sqrt{s^2 + 1}} = \frac{1}{s} \frac{1}{\sqrt{s^2 + 1}}$$

$$\mathcal{L}\{x_2(t)\} = \frac{1}{s} \frac{1}{\sqrt{s^2 + 1}} \frac{1}{s} \mathcal{L}\{1 + \cos t\} = \frac{1}{s^2} \frac{1}{\sqrt{s^2 + 1}} (1 + \cos t)$$

$$\mathcal{L}\{x_2(t)\} = \frac{1}{s^2} \mathcal{L}\{1 + \cos t\} = \frac{1}{s^2} \frac{1}{\sqrt{s^2 + 1}} \frac{1}{s} \mathcal{L}\{1 + \cos t\}$$

$$\therefore \frac{\partial \mathcal{L}}{\partial \mathcal{E}_1} = \frac{1}{2} \left(\frac{\partial}{\partial \mathcal{E}_1} \frac{1}{\sin^2 \theta_1 \cos^2 \theta_1 \mu_1 \mu_2 \cos^2 \theta_2} \right)$$

$$\mathcal{E}_1 \mathcal{L} = \frac{1}{2} \mathcal{E}_1$$

$$\frac{\partial \mathcal{L}}{\partial \mathcal{E}_2} = \frac{1}{2} \left(\frac{\partial}{\partial \mathcal{E}_2} \frac{1}{\sin^2 \theta_1 \cos^2 \theta_1 \mu_1 \mu_2 \cos^2 \theta_2} \right)$$

At $\theta = 0^\circ$ (90), $\theta = 2.25 \times 10^{-2}$ rad, $\mu_1 = 1.000293$, $\mu_2 = 1.5$

$$\frac{\partial \mathcal{L}}{\partial \mathcal{E}_2} = 0.5000735$$

Ex. 21 Given $\beta = \mu_2 / \mu_1$ and $\theta_1 = 0^\circ$

$$\text{From Eq. (19-102)} \quad \mathcal{R}_1 = \frac{2\mu_1 \sin^2 \theta_1}{\mu_1 + \mu_2} = \frac{2}{1 + \beta}, \quad \mathcal{R}_2 / \mathcal{E}_2 = 2\beta / (1 + \beta)$$

$$\text{From Eq. (19-103)} \quad \mathcal{E}_1 \mathcal{L} = \frac{1}{2} \mathcal{E}_1 \mathcal{L} = \frac{1}{2} \mathcal{E}_1, \quad \text{and } \mathcal{E}_2 \mathcal{L} = \mathcal{E}_2, \quad \text{and } \mathcal{E}_2 \mathcal{L} = \mathcal{E}_2$$

$$\text{From Eq. (19-104)} \quad \mathcal{L}_1 = \frac{\mathcal{E}_1 \mathcal{L} \cos^2 \theta_1}{\mathcal{R}_1 \mathcal{E}_1 \cos^2 \theta_1 + \mathcal{E}_2} = \frac{1}{2} \mathcal{E}_1$$

$$\text{From Eq. (19-105)} \quad \mathcal{R}_2 = \frac{\mathcal{E}_2 \mathcal{L} \cos^2 \theta_2}{\mathcal{R}_2 \mathcal{E}_2 \cos^2 \theta_2 + \mathcal{E}_1} = \frac{2\beta}{1 + \beta}$$

$$\text{From Eq. (19-106)} \quad \mathcal{L}_2 = \frac{\mathcal{E}_2 \mathcal{L} \cos^2 \theta_2}{\mathcal{R}_2 \mathcal{E}_2 \cos^2 \theta_2 + \mathcal{E}_1} = \frac{2\beta}{1 + \beta}$$

$$\text{From Eq. (19-107)} \quad \mathcal{L}_3 = \frac{\mathcal{E}_2 \mathcal{L} \cos^2 \theta_2}{\mathcal{R}_2 \mathcal{E}_2 \cos^2 \theta_2 + \mathcal{E}_1} = \frac{2\beta}{1 + \beta} \mathcal{E}_2 \mathcal{L}$$

$\mathcal{L}_1 = \mathcal{L}_2 = \mathcal{L}_3$, but the phase shift of the reflected wave depends on the polarization of the incident wave. There are standing waves in the air and exponentially decaying transmitted waves in the hemisphere.

$$\text{Ex. 22} \quad \mathcal{E}_{1n}^2 + \mathcal{E}_{1r}^2 = \mathcal{E}_1^2 = \omega^2 \mu_1 \mathcal{E}_1^2 = \mu_1 \omega^2 \mathcal{E}_1^2 \quad \text{--- (1)}$$

Continuity condition at $z=0$ for all ω and θ requires

$$\mathcal{E}_{1n} - \mathcal{E}_{1r} = \mathcal{E}_2 \sqrt{\mu_2} \sin \theta_2 = \mathcal{E}_2 \sqrt{\mu_2} \cos \theta_1 \quad \text{--- (2)}$$

$$\mathcal{H}_{1n} = \mathcal{H}_2 \sqrt{\mu_2} \quad \text{--- (3)}$$

Combining (1), (2) and (3), we eliminate for \mathcal{E}_{1n} and \mathcal{H}_{1n} in terms of ω , μ_1 , μ_2 , θ_1 , θ_2 , and \mathcal{E}_2 . And, thus

$$A_0^2 = \frac{1}{2} \rho_0 v_0^2$$

we have $v_0 = \sqrt{2A_0} = \frac{1}{\rho_0} \sqrt{2A_0} = \frac{1}{\rho_0} \sqrt{2A_0} \rho_0$

$$a) \quad v_0 = \sqrt{2A_0} = \sqrt{2 \cdot \frac{10^{-2}}{10^3}} = \sqrt{2 \cdot 10^{-5}} = \sqrt{2} \cdot 10^{-2.5} = 0.00447 \text{ m/s}$$

$$b) \quad v_0 = \frac{A_0 \omega}{\rho_0 v_0} \quad \omega = \frac{v_0}{\lambda} = \frac{v_0}{2\pi} \cdot 2\pi = \frac{v_0}{\lambda} \cdot 2\pi$$

$$= \frac{A_0 \cdot \frac{v_0}{\lambda} \cdot 2\pi}{\rho_0 v_0} = \frac{2\pi A_0}{\rho_0 \lambda} = \frac{2\pi \cdot 10^{-2}}{10^3 \cdot 0.01} = \frac{2\pi \cdot 10^{-2}}{10^1} = 2\pi \cdot 10^{-3} = 0.00628 \text{ rad/s}$$

$$c) \quad (v_0)_{\text{max}} = \frac{A_0 \omega}{\rho_0 v_0}$$

$$v_0 = \sqrt{2A_0} \quad v_0 = \frac{A_0 \omega}{\rho_0 v_0} \rightarrow (v_0)_{\text{max}} = \frac{A_0 \omega}{\rho_0 \sqrt{2A_0}}$$

$$\rightarrow \frac{(v_0)_{\text{max}}}{\sqrt{2A_0}} = \frac{A_0 \omega}{\rho_0 \sqrt{2A_0}} = \frac{A_0 \omega}{\rho_0 \sqrt{2A_0}}$$

$$d) \quad \text{for } (v_0)_{\text{max}} = 1 \text{ m/s} \rightarrow 1 = \frac{A_0 \omega}{\rho_0 \sqrt{2A_0}} = \frac{A_0 \omega}{\rho_0 \sqrt{2A_0}}$$

Ex 10



a) Small angle,

$$\frac{\sin \theta}{\sin \theta_1} = \frac{v_0}{v_1}$$

$$v_1 = v_0 \left(\frac{\sin \theta}{\sin \theta_1} \right)$$

$$b) \quad \cos \theta = \sqrt{1 - \left(\frac{v_0}{v_1} \right)^2}$$

$$A_1 = 2A_0 = 2A_0 \cos \theta = 2A_0 \sqrt{1 - \left(\frac{v_0}{v_1} \right)^2} = \frac{2A_0 \sqrt{1 - \left(\frac{v_0}{v_1} \right)^2}}{1}$$

$$c) \quad A_1 = 2A_0 = 2A_0 \cos \theta = \frac{2A_0 \sqrt{1 - \left(\frac{v_0}{v_1} \right)^2}}{1} = \frac{2A_0 \sqrt{1 - \left(\frac{v_0}{v_1} \right)^2}}{1}$$

Ex 11

$$a) \quad \sin \theta = \frac{v_0}{v_1} \rightarrow \sin \theta = \frac{v_0}{v_1} \sin \theta \quad \text{for } \theta = \theta_1$$

$$\sin \theta = \frac{v_0}{v_1} \sin \theta$$

From Eqs. (1-28) and (1-29):

$$L_1 \cos \alpha = L_2 L_3 e^{-i\alpha} e^{i\theta} e^{i\phi}$$

$$L_2 \cos \alpha = \frac{L_1}{L_3} (L_3 \cos \alpha + L_3 \sqrt{\frac{L_3^2 - L_1^2}{L_3^2}} \sin \alpha) e^{-i\alpha} e^{i\theta} e^{i\phi}$$

where $L_3 = L_1 \sin \alpha = L_3 \sqrt{\frac{L_3^2 - L_1^2}{L_3^2}} \sin \alpha$,

$$\alpha = \sin^{-1} \left(\frac{L_1}{L_3} \right) \sin \theta = \theta$$

$$L_3 = \frac{L_1 \sin \theta \cos \theta}{\cos^2 \theta - \left(\frac{L_1}{L_3} \right) \sin^2 \theta} \quad \text{from Eq. (1-28)}$$

(1) $(L_3)_{\theta=0} = \frac{L_1}{L_3} (L_3)_{\theta=0} = 0$

Ex-11 Given $\theta = \alpha$ when $\alpha = 0$, and $\alpha = 0$

(a) From Eq. (1-28): $(L_3)_{\alpha=0} = L_1$

(b) From Eq. (1-29): $(L_3)_{\alpha=0} = L_1 / L_3$

(c) $L_1 \cos \alpha = L_2 L_3 \cos \alpha \left[1 - \frac{L_3^2 - L_1^2}{L_3^2} \cos^2 \alpha - \frac{L_3^2 - L_1^2}{L_3^2} \sin^2 \alpha \right]$

$$L_1 \cos \alpha = L_2 L_3 e^{-i\alpha} \cos \alpha \left[1 - \frac{L_3^2 - L_1^2}{L_3^2} \right]$$

$$= L_2 L_3 e^{-i\alpha} \cos \alpha \left[\frac{L_3^2 + L_1^2}{L_3^2} \right]$$

where $\alpha = \sin^{-1} \left(\frac{L_1}{L_3} \right) \sin \theta = \theta$ when $\theta = \alpha$.

Ex-12 (a) $\alpha = \sin^{-1} \frac{L_1}{L_3} \sin \theta = \sin^{-1} \frac{L_1}{L_3} \sin \theta$

(b) $\alpha = \sin^{-1} \frac{L_1}{L_3} \sin \theta$, $\sin \alpha = \frac{L_1}{L_3} \sin \theta$ where $\cos \alpha = \sqrt{1 - \frac{L_1^2}{L_3^2} \sin^2 \theta}$

$$L_1 = \frac{L_1 L_3 \cos \alpha \sin \alpha}{L_3 \cos \alpha \sin \alpha} = e^{i\theta} e^{-i\alpha}$$

(c) $L_2 = \frac{L_1 L_3 \cos \alpha \sin \alpha}{L_3 \cos \alpha \sin \alpha} = L_3 e^{i\theta} e^{-i\alpha} \cos \alpha$

(d) The transmitted wave is air carrying an $e^{-i\alpha} e^{i\theta} e^{i\phi}$

where $\alpha = \sin^{-1} \left(\frac{L_1}{L_3} \right) \sin \theta = \frac{L_1}{L_3} \sin \theta$

Attenuation is air for each wavelength

$$= 2 \sin \alpha e^{-i\alpha} = 2 \sin \theta \cos \theta$$

Ex 20 When the incident light first strikes the Ag-Cu surface, $\theta_1 = \theta_2 = 0$, $\tau = \sqrt{\frac{\epsilon_2 \mu_2}{\epsilon_1 \mu_1}}$.

$$\frac{\partial R_{\text{Ag}}}{\partial \theta_1} = \frac{\partial}{\partial \theta_1} \tau^2 = \frac{\partial \left(\frac{\epsilon_2 \mu_2}{\epsilon_1 \mu_1} \right)}{\partial \theta_1}$$

Total reflection never occurs the point of both planing surfaces distance

$$\theta_2 = \sin^{-1} n_2 = \sin^{-1} \left(\frac{1}{\sqrt{2}} \right) = 45^\circ$$

On exit from the prism, $\theta_2 = \frac{\partial \theta_1}{\partial \theta_2}$.

$$\frac{\partial R_{\text{Ag}}}{\partial \theta_1} = \frac{\partial}{\partial \theta_1} \tau^2 = \frac{\partial \left(\frac{\epsilon_2 \mu_2}{\epsilon_1 \mu_1} \right)}{\partial \theta_1}$$

$$\therefore \frac{\partial R_{\text{Ag}}}{\partial \theta_1} = \left[\frac{\partial \left(\frac{\epsilon_2 \mu_2}{\epsilon_1 \mu_1} \right)}{\partial \theta_1} \right]^2 = \left[\frac{\partial \left(\frac{\epsilon_2 \mu_2}{\epsilon_1 \mu_1} \right)}{\partial \theta_2} \right]^2 = 0.278$$

Ex 21 a) $n_2 \sin \theta_2 = n_1 \sin(\theta_2' + \theta_1) = n_1 \sin \theta_1$

$$= n_1 \sqrt{1 - \cos^2 \theta_1} = n_1 \sqrt{1 - \cos^2 \theta_2} = n_1 \sin \theta_2$$

$$\sin \theta_2 = \frac{n_1}{n_2} \sqrt{1 - \cos^2 \theta_2} = \sqrt{1 - \cos^2 \theta_2} \quad (n_2 = 1)$$

$$\text{b) } n_1 \sin \theta_1 = n_2 \sin \theta_2 = \sqrt{1 - \cos^2 \theta_2} = \sin \theta_2$$

$$\theta_1 = \sin^{-1} \sin \theta_2 = \theta_2$$

Ex 22 $R_p(\theta = 0) = \frac{n_2}{n_1} = \frac{1.5}{1.0} = 1.5$

$$\text{a) } R_p(\theta = 30^\circ) \tau_s = \frac{\epsilon_2 \mu_2 \cos^2 \theta_1 \cos^2 \theta_2}{\epsilon_1 \mu_1 \cos^2 \theta_2 + \epsilon_2 \mu_2} = \frac{1.5 \cdot 1.0 \cdot \cos^2 30^\circ \cdot \cos^2 30^\circ}{1.0 \cdot 1.0 \cos^2 30^\circ + 1.5 \cdot 1.0}$$

$$= \frac{1.5 \cdot 1.0 \cdot \cos^2 30^\circ}{1.0 \cos^2 30^\circ + 1.5}$$

$$R_p(\theta = 30^\circ) \tau_s = \frac{1.5 \cdot 1.0 \cdot \cos^2 30^\circ}{1.0 \cos^2 30^\circ + 1.5}$$

$$\text{b) } R_p(\theta = 60^\circ) \tau_s = \frac{\epsilon_2 \mu_2 \cos^2 \theta_1 \cos^2 \theta_2}{\epsilon_1 \mu_1 \cos^2 \theta_2 + \epsilon_2 \mu_2} = \frac{1.5 \cdot 1.0 \cdot \cos^2 60^\circ \cdot \cos^2 60^\circ}{1.0 \cdot 1.0 \cos^2 60^\circ + 1.5 \cdot 1.0}$$

$$= \frac{1.5 \cdot 1.0 \cdot \cos^2 60^\circ}{1.0 \cos^2 60^\circ + 1.5}$$

$$R_p(\theta = 60^\circ) \tau_s = \frac{1.5 \cdot 1.0 \cdot \cos^2 60^\circ}{1.0 \cos^2 60^\circ + 1.5}$$

Ex-11) a) For perpendicular polarization and $\mu_1 = \mu_2 = \mu_0$

$$\sin \theta_{cp} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$$

Under condition of no-reflection:

$$\begin{aligned} \cos \theta &= \sqrt{1 - \frac{\epsilon_2}{\epsilon_1} \sin^2 \theta_{cp}} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}} \\ &= \sin \theta_{cp} \implies \theta_i = \theta_{cp} = \theta_r \end{aligned}$$

b) For parallel polarization and $\mu_1 = \mu_2 = \mu_0$

$$\sin \theta_{cp} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$$

$$\begin{aligned} \cos \theta &= \sqrt{1 - \frac{\epsilon_2}{\epsilon_1} \sin^2 \theta_{cp}} = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}} \\ &= \sin \theta_{cp} \implies \theta_i = \theta_{cp} = \theta_r \end{aligned}$$

Ex-12) a) $\sin \theta_i = \sqrt{\frac{\epsilon_2}{\epsilon_1}}$; $\sin \theta_r = \frac{1}{\sqrt{1 + \frac{\epsilon_2}{\epsilon_1}}}$



$$\implies \cos \theta_r = \sqrt{\frac{\epsilon_2}{\epsilon_1}}$$

$$\therefore \sin \theta_i = \cos \theta_r \quad (\theta_i > \theta_r)$$

b) Let $n_1/n_2 = n$.



Ex-13) a) For perpendicular polarization:

$$r_{\perp} = \frac{E_{i\perp} - E_{t\perp}}{E_{i\perp} + E_{t\perp}}$$

$$\sin \theta_i = \sqrt{\frac{\epsilon_2}{\epsilon_1}} \sin \theta_t \implies \cos \theta_t = \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} \cos \theta_i$$

$$r_{\perp} = \frac{E_{i\perp} - \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} E_{t\perp}}{E_{i\perp} + \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} E_{t\perp}}$$

$$r_{\perp} = \frac{E_{i\perp} \cos \theta_i - \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} E_{t\perp} \cos \theta_t}{E_{i\perp} \cos \theta_i + \sqrt{1 - \frac{\epsilon_2}{\epsilon_1}} E_{t\perp} \cos \theta_t}$$

For parallel polarization:

$$G = \frac{\frac{E_0 \cos \theta_i \cos \theta_t}{\mu_0} - \frac{E_0 \cos \theta_i \cos \theta_r}{\mu_0}}{\frac{E_0 \cos \theta_i \sin \theta_t}{\mu_0} + \frac{E_0 \cos \theta_i \sin \theta_r}{\mu_0}}$$

$$G = \frac{1 - \frac{\mu_2 \cos \theta_t}{\mu_1 \cos \theta_r}}{1 + \frac{\mu_2 \sin \theta_t}{\mu_1 \sin \theta_r}}$$

② $n_2/n_1 = 2.10$, $\theta_i = \theta_t = 45^\circ \rightarrow G = 10 \sqrt{2} \approx 14.14$



Ex-21



$$\text{Given } E_{i, \parallel} = E_{r, \parallel} + E_{t, \parallel} \quad \text{--- (1)}$$

$$E_{i, \perp} = E_{r, \perp} + E_{t, \perp} \quad \text{--- (2)}$$

$$E_{i, \parallel} = \frac{1}{2} E_{i, \perp} + E_{t, \parallel}$$

$$= \frac{1}{2} E_{i, \perp} + E_{t, \parallel} \quad \text{--- (3)}$$

$$E_{i, \parallel} = \frac{1}{2} E_{i, \perp} + \frac{2n_1 \cos \theta_i \sin \theta_t}{2n_1 \cos \theta_i \sin \theta_t + 2n_2 \sin \theta_i \cos \theta_t} E_{i, \perp}$$

a) From Eq. (3):

$$E_{r, \perp} = \frac{2n_1 \cos \theta_i \sin \theta_t - n_2 \sin \theta_i \cos \theta_t}{2n_1 \cos \theta_i \sin \theta_t + 2n_2 \sin \theta_i \cos \theta_t} E_{i, \perp}$$

$$\text{where } (n_1 \cos \theta_i) = n_2 \cos \theta_t = \text{constant}$$

$$E_{i, \parallel} = E_{r, \parallel} + E_{t, \parallel} \quad \text{--- (1)}$$

$$E_{i, \perp} = \frac{1}{2} E_{i, \parallel} + E_{t, \perp} = \frac{1}{2} E_{i, \parallel} + E_{t, \perp} \quad \text{--- (2)}$$

b) From Eq. (8-110) $\sin \theta_1 = \frac{Z_1 \sin \theta_2}{Z_2 - jZ_1}$ (Complex).

$$\cos \theta_1 = \sqrt{1 - \sin^2 \theta_1} \quad (\text{Complex}).$$

The x - and y -components of \vec{E}_2 in Eq. (8) above are different amplitudes and are out of phase, indicating that it is elliptically polarized.

8-112

$$a) \Gamma_{\parallel} = \frac{Z_2 \cos \theta_2}{Z_1 \cos \theta_1} \Big|_{\text{av}} = \frac{Z_2 \cos \theta_2}{Z_1 \cos \theta_1} = \frac{Z_2}{Z_1} = \Gamma_{\parallel} = \frac{Z_2 \cos \theta_2 - Z_1 \cos \theta_1}{Z_2 \cos \theta_2 + Z_1 \cos \theta_1}$$

$$\tau_{\parallel} = \frac{2Z_2 \cos \theta_2}{Z_2 \cos \theta_2 + Z_1 \cos \theta_1} = \tau_{\parallel} \left(\frac{\cos \theta_2}{\cos \theta_1} \right) = \frac{2Z_2 \cos \theta_2}{Z_2 \cos \theta_2 + Z_1 \cos \theta_1}$$

b) From part a) we have

$$1 + \Gamma_{\parallel}^2 = \tau_{\parallel}^2$$

This compares with

$$1 - \Gamma_{\perp}^2 = \tau_{\perp}^2 \left(\frac{\cos \theta_2}{\cos \theta_1} \right)^2 \quad \text{in Eq. (8-110).$$

Chapter 9

Theory and Application of Transmission Lines

Ex.1

$$P \cdot Z = \begin{vmatrix} Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \end{vmatrix} = Z_1 Z_2 Z_3 \longrightarrow \frac{P}{Z} = 1$$

$$P \cdot Z = \begin{vmatrix} Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \\ Z_1 & Z_2 & Z_3 \end{vmatrix} = Z_1 Z_2 Z_3 \longrightarrow \frac{P}{Z} = 1$$

Ex.2

a) $P = (Z_1 Z_2 + Z_2 Z_3) = \rho \sin(\alpha) (Z_1 Z_2 + Z_2 Z_3)$

$$\longrightarrow \begin{cases} \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (1)} \\ \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (2)} \\ \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (3)} \end{cases}$$

b) $P = (Z_1 Z_2 + Z_2 Z_3) = \rho \sin(\alpha) (Z_1 Z_2 + Z_2 Z_3)$

$$\longrightarrow \begin{cases} \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (1)} \\ \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (2)} \\ \frac{P}{Z} = \rho \sin(\alpha) & \text{--- (3)} \end{cases}$$

From (1) and (2) $\frac{P}{Z} = \rho \sin(\alpha)$ --- (4)

From (2) or (3) $\frac{P}{Z} = \rho \sin(\alpha)$ --- (5)

From (3) or (1) $\frac{P}{Z} = \rho \sin(\alpha)$ --- (6)

(i) From (4) $\frac{P}{Z} = \rho \sin(\alpha)$ --- (7)

From (4), (5), and (6) $\frac{P}{Z} = \rho \sin(\alpha) = \rho \sin(\alpha) = \rho \sin(\alpha) = \rho \sin(\alpha)$ --- (8)

Combining (7) and (8), we have $\frac{P}{Z} = \rho \sin(\alpha)$ --- (9)

Similarly, $\frac{P}{Z} = \rho \sin(\alpha)$ --- (10)

Ex.3

$Z_1 = \frac{Z_0}{\cos(\theta)}$

a) $Z_1 = \frac{Z_0}{\cos(\theta)} = \frac{Z_0}{\cos(\theta)} \longrightarrow \rho = \frac{P}{Z}$

b) $Z_1 = \frac{Z_0}{\cos(\theta)} = \frac{Z_0}{\cos(\theta)} \longrightarrow \rho = \frac{P}{Z}$

$$d) \quad \xi_2 = \frac{1-i\sqrt{3}}{2} = \frac{1}{2} - \frac{i\sqrt{3}}{2} \longrightarrow \text{w/o 1st.}$$

$$e) \quad \eta_2 = \frac{1}{\sqrt{2}} \longrightarrow \begin{array}{l} \eta_{2,1} = \eta_2/2 \text{ for } \cos \pi/4, \\ \eta_{2,2} = \eta_2/2 \text{ for } \sin \pi/4, \\ \eta_{2,3} = \eta_2 \text{ for } \tan \pi/4. \end{array}$$

Ex 2.1 Given: $\xi_1 = (1+i)^n$ (real), $\xi_2 = (1-i)^n$ (real), $\xi_3 = (1+i)^{2n}$ (real)
 Using distributive rules: $\eta_1 = 1, \eta_2 = 1, \eta_3 = (1+i)^{2n}$.
 $f = 2 \cdot (1+i)^{2n}$ (real).

$$a) \quad \xi_1 = \frac{1}{2} \sqrt{\frac{1000}{3}} = 1.7 \text{ (real)}$$

$$\xi_2 = \frac{1}{2} \sqrt{\frac{1000}{3}} = 1.7 \text{ (real)}$$

$$\xi_3 = \frac{1}{2} \sqrt{\frac{1000}{3}} = 1.7 \text{ (real)}$$

$$\xi_4 = \frac{1}{2} \sqrt{\frac{1000}{3}} = 1.7 \text{ (real)}$$

$$b) \quad \frac{\xi_1}{\xi_2} = \sqrt{\frac{1000}{3}} = 1.7 \cdot (1+i)^{2n}$$

$$c) \quad \eta_1 = 1 \text{ (real)}, \quad \eta_2 = 1 \text{ (real)}$$

$$\xi_1 = \sqrt{2} \left[1 + \frac{1}{2} \left(\frac{1}{\sqrt{2}} + \frac{1}{\sqrt{2}} \right) \right] = 1.7 \cdot (1+i)^{2n}$$

$$\xi_2 = \sqrt{2} \left[1 + \frac{1}{2} \left(\frac{1}{\sqrt{2}} - \frac{1}{\sqrt{2}} \right) \right] = 1.7 \cdot (1-i)^{2n}$$

Ex 2.2 Solving for ξ_1 (real) and ξ_2 (real)

$$\xi_1 \xi_2 = \xi_1^2 + \xi_2^2$$

$$\xi_1 \xi_2 = \xi_1^2 + \xi_2^2$$

Let $\xi_1 = \xi_2 = \xi$

$$\xi^2 = (1+i)\xi^2 + (1-i)\xi^2 = 2\xi^2$$

which implies

$$\xi^2 = (1+i)\xi^2 = 0$$

$$\xi^2 = (1-i)\xi^2 = 0$$

$$\therefore \frac{\xi_1}{\xi_2} = -\frac{\xi_1}{\xi_2} = \frac{1-i}{1+i}$$

$$\text{Ex 2.3} \quad \eta = (1+i)^2 = 1 + 2i + i^2 = 1 + 2i - 1 = 2i$$

$$\text{From Eq. 2-23, } \xi = (1+i)^2 = 2i \cdot (1+i)^2 = 2i \cdot 2i = -4$$

Squaring both sides, we obtain two equations (one
 for real and imaginary parts):

$$\begin{aligned} x^2 + y^2 &= -\frac{a^2}{2} \\ 2xy &= \frac{a^2}{2} \end{aligned}$$

From which Eqs. (17-21) and (17-22) follow.

$$\begin{aligned} \text{E21.1} \quad x &= \pm \sqrt{2} (1 - i \frac{a}{2})^n (1 + i \frac{a}{2})^n \\ &= \pm \sqrt{2} \left[(1 - i \frac{a}{2})^n + i \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) \right] \\ &= \pm \sqrt{2} \left[(1 - i \frac{a}{2})^n + i \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) \right] = \pm \sqrt{2} a^n \end{aligned}$$

$$\begin{aligned} \text{---} \quad &= \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) \\ &= \pm \sqrt{2} \left(1 + \dots + \frac{a^n}{2} \right) \end{aligned}$$

$$\begin{aligned} &= \pm \sqrt{2} (1 - i \frac{a}{2})^n (1 + i \frac{a}{2})^n \\ &= \pm \sqrt{2} \left[(1 - i \frac{a}{2})^n + i \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) \right] = \pm \sqrt{2} a^n \end{aligned}$$

$$\begin{aligned} \text{---} \quad &= \pm \sqrt{2} \left[1 + \dots + \frac{a^n}{2} \right] \\ &= \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right) \end{aligned}$$

$$x = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right)$$

$$\text{E21.2} \quad y = \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right) = \pm \sqrt{2} a^n$$

$$\text{---} \quad = \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right), \quad y = \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right)$$

$$y = \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) = \pm \sqrt{2} a^n$$

$$\text{---} \quad = \pm \sqrt{2} \left(\frac{a^n}{2} + \dots + \frac{a^n}{2} \right) = \pm \sqrt{2} a^n$$

$$y = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right)$$

$$\text{E21.3} \quad x = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right), \quad y = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right)$$

From Eqs. (17-21), (17-22) $x = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right)$, $y = \pm \sqrt{2} \left(1 - \frac{a^n}{2} \right)$

Given $Z_0 = 20 + j10 \text{ } \Omega$,
 μ is real constant (positive),
 β is a real (positive),
 $\gamma = \alpha^2 + j\omega\beta$.

$$Z_1 = Z_0 = 20 + j10 \text{ } \Omega \text{ (initial)}, \quad Z_2 = \frac{Z_0}{1 - \mu} = 20 \text{ } \Omega \text{ (parallel)},$$

$$Z_3 = \frac{Z_0}{\mu} = 10 \text{ } \Omega \text{ (series)}, \quad Z_4 = \frac{Z_0}{1 + \mu} = 10 \text{ } \Omega \text{ (parallel)}.$$

Ex. 10.10 (a) For lossy transmission line:

$$Z_1 = \sqrt{\frac{L}{C}} = \frac{1}{\beta} \sqrt{\frac{L}{C}} \tanh^2(\beta Z_0) = \frac{1}{\beta} \sqrt{\frac{L}{C}} \left(\frac{Z_0}{Z_0} - \frac{Z_0}{Z_0} \right) = 20 + j10 \text{ } \Omega$$

$$\frac{1}{\beta} = 20 \text{ } \Omega \text{ } \longrightarrow \beta = 0.05 \text{ } \text{rad/m}$$

(b) For lossless transmission line:

$$Z_1 = \frac{Z_0}{\beta} \sqrt{\frac{L}{C}} \coth(\beta Z_0) = \frac{1}{\beta} \coth(\beta Z_0) = 10 \text{ } \Omega$$

$$\frac{1}{\beta} = 10 \text{ } \Omega \text{ } \longrightarrow \beta = 0.1 \text{ } \text{rad/m}$$

$$\text{Ex. 10.11 } (P_{\text{in}})_1 = (P_{\text{in}})_2 = \frac{1}{2} \operatorname{Re} \{ V_1 I_1^* \} \quad V_1 = \sqrt{\frac{2}{3}} V_2 \sqrt{3}$$

$$= \frac{1 \text{ V} \cdot 3 \text{ A}}{\sqrt{3} \cdot \sqrt{3} \cdot \sqrt{3} \cdot \sqrt{3}} \quad I_1 = \frac{3}{\sqrt{3}} I_2$$

$$\text{To maximize } (P_{\text{in}})_1, \text{ set } \frac{d(P_{\text{in}})_1}{dR_1} = 0, \quad \left. \begin{array}{l} R_1 = R_2 = Z_0 = 10 \text{ } \Omega \\ \text{and } \frac{d(P_{\text{in}})_1}{dR_2} = 0 \end{array} \right\} R_1 = R_2 = Z_0$$

$$\text{Max. } (P_{\text{in}})_1 = \frac{1 \text{ W}}{3} = (P_{\text{in}})_2$$

\longrightarrow Max. power transfer efficiency = 50%.

Ex. 10.12 $V(0) = V_1^+ e^{j\beta z} + V_1^- e^{-j\beta z}$,

$$I(0) = I_1^+ e^{j\beta z} + I_1^- e^{-j\beta z}$$

At $z=0$, $V(0) = V_1^+ + V_1^-$, $I(0) = I_1^+ + I_1^- = \frac{1}{Z_0}(V_1^+ - V_1^-)$
 $\longrightarrow V_1^+ = \frac{1}{2}(V_1 + Z_0 I_1)$, $V_1^- = \frac{1}{2}(V_1 - Z_0 I_1)$

(a) $V(z) = \frac{1}{2}(V_1 + Z_0 I_1) e^{j\beta z} + \frac{1}{2}(V_1 - Z_0 I_1) e^{-j\beta z}$
 $I(z) = \frac{1}{2Z_0}(V_1 + Z_0 I_1) e^{j\beta z} - \frac{1}{2Z_0}(V_1 - Z_0 I_1) e^{-j\beta z}$

(b) $V(z) = V_1 \cosh \beta z + Z_0 I_1 \sinh \beta z$,
 $I(z) = I_1 \cosh \beta z + \frac{V_1}{Z_0} \sinh \beta z$

Ex. 10 From Eq. (1) and (2) $x = \frac{1}{2}z + \frac{1}{2}(y-z)$
 $= (\frac{1}{2} + \frac{1}{2})z = z$ ③

Also $y = z + (y-z) = \frac{1}{2}z + y$
 $= \frac{1}{2}z + (1 + \frac{1}{2}z)y$ ④

Substituting ③ in ④:
 $y = (1 + \frac{1}{2}z)y + z(\frac{1}{2} + \frac{1}{2}z)y$ ⑤

At $z=0$, Eq. (5) and (6) are identical and (5) reduces to (6):

$$y = 0 = (1 + 0)y + 0(\frac{1}{2} + 0)y$$
 ⑥

$$z = 0 = (\frac{1}{2} + 0)y + (1 + 0)z$$
 ⑦

Both Eqs. ⑥ & ⑦ and Eqs. ⑤ & ⑥ are of the following form: $\begin{bmatrix} A \\ B \end{bmatrix} = \begin{bmatrix} a & b \\ c & d \end{bmatrix} \begin{bmatrix} x \\ y \end{bmatrix}$ ⑧

where $A = 0$, $a = 1 + \frac{1}{2}z$, $b = 0$, ⑨

$B = z$, $c = \frac{1}{2} + \frac{1}{2}z$, $d = 1 + \frac{1}{2}z$ ⑩

and $C = \frac{1}{2}z$, $e = \frac{1}{2}$, $f = 1 + \frac{1}{2}z$ ⑪

——— $AD - BC = 0 - 0 = 0$, $eA - cB = 0 - 0 = 0$

∴ ⑧ is the required result in Eq. (5) and
 Eq. (6) can be obtained by using ⑧ in ⑤:
 $z = \frac{1}{2}(1 + z)y + (1 + \frac{1}{2}z)z$

Ex. 11 $z = \frac{1}{2}y - 2z$, $-\frac{3}{2}z = \frac{1}{2}y$

$$\frac{3}{2}z = -\frac{1}{2}y$$

$$\frac{3}{2}z = -\frac{1}{2}y$$

At $y=0$, $z = 0 = \frac{1}{2}x$

$z = 0 = \frac{1}{2}x + \frac{1}{2}y$, $x = 0$

$$\frac{3}{2}z = -\frac{1}{2}y = -\frac{1}{2} \cdot \frac{3}{2}z$$

$$\begin{aligned} \text{We have } \quad \text{VIB} &= \int (r_2 + c_2) e^{at} dt = \int (r_2 + c_2) e^{at} dt \\ \text{IIB} &= \int \left(\frac{r_2}{a} + c_2 \right) e^{at} dt = \int \left(\frac{r_2}{a} + c_2 \right) e^{at} dt \end{aligned}$$

$$\text{where } r_2 = \frac{R_2}{s_1 - s_2} r_1 \quad \text{and } c_2 = \frac{R_2}{s_1 - s_2} c_1$$

a) For an infinite line, $R_2 = R_1$:

$$\text{VIB} = \frac{R_1}{s_1 - s_2} r_1 e^{at}, \quad \text{IIB} = \frac{R_1}{s_1 - s_2} c_1 e^{at}$$

b) For a finite line of length l terminated at R_2 :

$$R_2 = R_1 \frac{R_1 + R_2 \cosh al}{R_1 + R_2 \sinh al}$$

Ex. 11 Distributed line: $R_1 = r_1 \sqrt{Z} = 20 \text{ } \Omega$, $R_2 = 20 \text{ } \Omega$

$$\text{So } \left(\frac{R_2}{R_1} \right) = \left(\frac{R_1}{R_2} \right) = 1 \text{ and}$$

$$\Rightarrow \frac{R_1}{R_1} = 1 \text{ and } \frac{R_2}{R_2} = 1 \Rightarrow \frac{R_1}{R_1} + \frac{R_2}{R_2} = 2 \Rightarrow 2 = \frac{R_1}{R_1}$$

$$1 = \frac{R_1}{R_1} = 1 \text{ and } 1 = 1, \quad c = \frac{R_1}{R_1} = 1 \text{ and } c = 1$$

$$c = \frac{R_1}{R_1} = 1 \text{ and } c = 1, \quad \frac{R_1}{R_1} = 1 \text{ and } \frac{R_2}{R_2} = 1$$

$$\Rightarrow \text{VIB} = \frac{R_1}{s_1 - s_2} e^{at} e^{bt} = \frac{R_1}{s_1 - s_2} e^{(a+b)t}, \quad \text{IIB} = \frac{R_1}{s_1 - s_2}$$

$$\therefore \text{VIB} = 20 e^{(a+b)t} \text{ and } \text{IIB} = 20 \text{ } \Omega \quad (1)$$

$$\text{IIB} = 20 e^{(a+b)t} \text{ and } \text{IIB} = 20 \text{ } \Omega \quad (2)$$

$$b) \text{ If } R_2 = 20 \text{ } \Omega \text{ then } \text{IIB} = 20 \text{ } \Omega \text{ and } \text{IIB} = 20 \text{ } \Omega \quad (3)$$

$$\text{IIB} = 20 \text{ } \Omega \text{ and } \text{IIB} = 20 \text{ } \Omega \quad (4)$$

$$c) \text{ If } R_2 = \frac{R_1}{2} = 10 \text{ } \Omega \text{ then } \text{IIB} = 10 \text{ } \Omega \text{ and } \text{IIB} = 10 \text{ } \Omega \quad (5)$$

Ex. 12 From Eq. (9-113) $Z_0 = Z_0 \text{ and } Z_0 = Z_0 \text{ } \Omega$

$$\text{From Eq. (9-113) and (9-112) } Z_0 = \sqrt{\frac{R_1 + R_2}{s_1 - s_2}}$$

$$\therefore Z_0 = (1 + j) \text{ } \Omega$$

$$d) \text{ From Eq. (9-113) } Z_0 = Z_0 \text{ and } Z_0 = \frac{R_1}{s_1 - s_2} = \frac{R_1}{s_1 - s_2}$$

Ex 2.11 a) From Eq. (9-100) $Z_{in} = Z_0$, then $\Gamma_{in} = Z_0 \frac{1 - \Gamma_{in}^{*2}}{1 + \Gamma_{in}^{*2}}$.

For $Z_0 = Z_0$, $\Gamma_{in} = \Gamma_{in}$, which is 1.

$$Z_{in} = Z_0 \frac{1 + \Gamma_{in}^{*2}}{1 - \Gamma_{in}^{*2}} = Z_0 \frac{1 + \Gamma_{in}^{*2}}{1 - \Gamma_{in}^{*2}}$$

$$= 4Z_0 / 3.$$

b) From Eq. (9-100) $Z_{in} = Z_0$, then $\Gamma_{in} = Z_0 \frac{1 - \Gamma_{in}^{*2}}{1 + \Gamma_{in}^{*2}}$.

For $Z_0 = Z_0$, $Z_{in} = Z_0 \frac{1 + \Gamma_{in}^{*2}}{1 - \Gamma_{in}^{*2}} = Z_0 \frac{1 + \Gamma_{in}^{*2}}{1 - \Gamma_{in}^{*2}}$

$$= Z_0 / 3.$$

Ex 2.12 $\beta L = \frac{2\pi}{\lambda} L = \frac{\pi}{2} = 90^\circ$

then $\beta L = 90^\circ + 180^\circ = 270^\circ$.

$$Z_L = Z_0 \frac{Z_0 + jZ_0 \tan \beta L}{Z_0 - jZ_0 \tan \beta L} = Z_0 \frac{Z_0 + jZ_0 \tan 270^\circ}{Z_0 - jZ_0 \tan 270^\circ}$$

$$= 2Z_0 - j2Z_0 \tan 270^\circ$$

Ex 2.13 a) From $Z_L = Z_0$, then $\Gamma_L = 2Z_0 \cos^2 \beta L$ (Eq. 9-101)

$Z_L = Z_0$, then $\Gamma_L = 2Z_0 \sin^2 \beta L$ (Eq. 9-102)

b) $Z_L = \sqrt{Z_0 Z_0} = 2Z_0 \cos^2 \beta L = 2Z_0 \sin^2 \beta L$ (Eq. 9-101)

then $\Gamma_L = \sqrt{\frac{Z_L}{Z_0}} = \sqrt{2} \cos \beta L = \sqrt{2} \sin \beta L = \sqrt{2} \cos \beta L \sin \beta L$

$L = 4 \text{ cm} \implies \beta L = 2.513 \text{ rad}$ (Eq. 9-101)

$\beta = 0.628 \text{ rad/cm}$ (Eq. 9-102)

c) $Z_L = \sqrt{\frac{Z_0 Z_0}{2}} = \sqrt{2} Z_0 \cos^2 \beta L$

then $\beta L \cos \beta L = \sqrt{2} Z_0$, $\beta L \sin \beta L = \frac{Z_0}{\sqrt{2}}$

$\tan \beta L = \beta L \sin \beta L / \beta L \cos \beta L = 0.707 \text{ rad/cm}$ (Eq. 9-101)

then $\beta L = 0.628 \text{ rad}$, $L = 0.314 \text{ cm}$ (Eq. 9-102)

$\beta = 0.628 \text{ rad/cm}$, $C = 0.15 \text{ cm}^{-1}$ (Eq. 9-103)

P. 2.22 (a) Since the line is very short compared to a wave length, we may use $\beta \approx \omega \sqrt{\mu_0 \epsilon_0} = \omega/c$

$$\begin{aligned} \therefore Z_0 &= \frac{\sqrt{\mu_0/\epsilon_0}}{\beta} = \sqrt{\mu_0/\epsilon_0} (c/\omega) \\ Z_0 &= \frac{\sqrt{\mu_0/\epsilon_0}}{\omega/c} = \sqrt{\mu_0/\epsilon_0} (c/\omega) \\ \beta Z_0 &= \omega \sqrt{\mu_0 \epsilon_0} \implies Z_0 = \frac{1}{\beta \sqrt{\mu_0 \epsilon_0}} = c/\omega \end{aligned}$$

(b) $\beta = \frac{\omega}{c} = 2\pi \times 10^8 \text{ rad/m} = 2\pi \times 10^8 \text{ (rad/m)}$; $\beta Z_0 = \frac{2\pi \times 10^8}{3 \times 10^8} = 2\pi$

$$\therefore Z_{in} = Z_0 \cot \beta l = -\frac{j}{2\pi} = -j37.7 \text{ } \Omega$$

$$Z_{in} = Z_0 \tanh \beta l = \tanh 2\pi = 0.996 \text{ } \Omega$$

P. 2.23 From eq. (2.10) $Z_{in} = Z_0 \cot \beta l = Z_0 \frac{\cos \beta l \sin \beta l}{\sin \beta l \cos \beta l}$

$$= Z_0 \frac{\sin 2\beta l}{\cos 2\beta l} = Z_0 \tan 2\beta l \quad \text{--- (1)}$$

For a lossless line, $\cot \beta l = Z_0 \frac{\cos \beta l \sin \beta l}{\sin \beta l \cos \beta l}$ --- (2)

At $l = \lambda/4$, $\beta l = \pi/2$ (inductance) --- (3)

When the frequency is slightly off resonance:

$$f = f_0 + \Delta f \text{ (slightly above } f_0 \text{), } \beta l = \pi/2 + \Delta \beta l \text{ (slightly above } \pi/2 \text{)}$$

$$\text{and } \Delta \beta l \text{ is very small } \Delta \beta l \approx 2\pi \times \Delta f \times \frac{l}{v}$$

--- (4) Indeed, after having set standard order small terms:

$$Z_{in} \approx \frac{j \tan 2\beta l}{\cos 2\beta l} \quad \text{--- (5)}$$

Combining (1) & (5) $\frac{\sin 2\beta l}{\cos 2\beta l} = \frac{j \tan 2\beta l}{\cos 2\beta l}$ --- (6)

Half-power point at: $\cos(2\beta l) = 1$; or $\Delta f = \pm \frac{\Delta \beta l}{2}$ --- (7)

For $l = \lambda/4$, $\beta l = \pi/2$ ($Z_{in} = j\infty$), and $\Delta f = \pm \frac{\Delta \beta l}{2}$

which, for a lossless transmission line, becomes (7)

$$\Delta f = \pm \frac{\Delta \beta l}{2} = \pm \frac{\beta l}{2} \left(\frac{1}{c} - \frac{1}{v} \right) \rightarrow \frac{\text{Half-power bandwidth } \Delta f}{f_0} = \frac{\beta l}{2} \left(\frac{1}{c} - \frac{1}{v} \right)$$

$$\Delta = \frac{\beta l}{2} = \frac{2\pi f_0 l}{2c} = \frac{\pi f_0 l}{c}$$

Q) From Eq. (b) above, $v = \beta c = \frac{c\sqrt{1-\beta^2}}{\beta}$

where $\beta = v/c = \beta_0$, and $\beta = \frac{v}{c} = \frac{\beta_0}{\gamma}$

$$\implies \beta_0 = \frac{(c - c + \beta_0 c) \gamma}{\beta_0 \gamma} = \frac{\beta_0 c \gamma}{\beta_0 \gamma}$$

$\implies 1 = \frac{1}{\gamma}$ for $v = c$ and $\beta_0 = c/c$.

Also, $\beta_0 = \frac{c\sqrt{1-\beta^2}}{\beta} \implies 1 = \frac{1}{\beta_0 \gamma} [c - c\beta_0 \sqrt{1-\beta^2} + \beta_0^2 c]$

$\beta_0 = 1$ yields equation + (a) above.

For $\beta_0 = \frac{1}{2}$, $1 = \frac{1}{\frac{1}{2} \gamma} \implies \beta_0 = \frac{1}{2} \gamma$

Use Eq. (a) above to obtain β_0 correct to the last of significant figures.

Ex. 20) a) $|r|^2 = \left| \frac{(x_1 - x_2) + i y_1}{(x_1 - x_2) + i y_1} \right| = \frac{(x_1 - x_2)^2 + y_1^2}{(x_1 - x_2)^2 + y_1^2}$

$$\frac{y_1^2}{y_1^2} = 1 \implies x_1 = \sqrt{y_1^2 - x_2^2}$$

Let $x_1 = 40 \sqrt{10000 - 25}$, $x_2 = 10 \sqrt{100}$

a) $\text{Min. } |r| = \sqrt{\frac{y_1^2 - x_2^2}{y_1^2}} = \sqrt{\frac{10000 - 25}{10000}} = \frac{1}{10}$

$\text{Min. } \beta = \frac{1/10}{1} = 0.1$

b) From Eq. (b) above, $v = \beta c = \frac{c\sqrt{1-\beta^2}}{\beta} = 0.995c$

$$\implies 1 = \frac{1}{\beta \gamma} [(1-\beta) \sqrt{1-\beta^2} + \beta^2] \quad \left(\frac{\text{Equation (a)}}{\beta \gamma} \right)$$

At voltage minimum, $\beta = \frac{1}{2} = \frac{1}{2}$

$\beta = \frac{1}{2}$ (the negative sign)

Use Eq. (a) above to obtain $\beta_0 = \frac{1}{2} \gamma \implies \beta_0 = \frac{1}{2}$

\therefore Voltage minimum correct to the last is $\left(\frac{1}{10} - \frac{1}{10}\right)$
or $0.1/10$ from the last.

Ex 11.12 (i) From Eq. (9-113a) and (9-114):

$$v(t) = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t + \phi)]$$

$$\text{where } V = \frac{V_1 - V_2}{2} = |V|e^{j\phi}, \quad \phi = \alpha_2 - \alpha_1$$

$$\text{Max } |v(t)| = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t)] \text{ for } \phi = 0$$

$$\text{min } |v(t)| = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 - \cos(2\omega t)] \text{ for } \phi = \pi$$

$$\text{So } V = \frac{\text{Max } |v(t)|}{\text{min } |v(t)|} = \frac{1 + \cos(2\omega t)}{1 - \cos(2\omega t)} \quad \left\{ \begin{array}{l} \text{Upper envelope is } V \\ \text{Lower envelope is } -V \end{array} \right.$$

(ii) From Eq. (9-113): $Z_1(\omega) = \frac{1 - \cos(\omega t)}{1 - \cos(\omega t)} \frac{1 - \cos(\omega t)}{1 - \cos(\omega t)} Z_2$

$$\text{At a voltage max, } \phi = 0, \quad Z_1(\omega) = \frac{Z_2}{1 - \cos(\omega t)}$$

$$\text{(c) At a voltage min, } \phi = \pi, \quad Z_1(\omega) = \frac{Z_2}{1 + \cos(\omega t)}$$

Ex 11.13 From Eq. (9-113): $Z_1 = Z_2 \frac{1 - \cos(\omega t)}{1 - \cos(\omega t)} \rightarrow Z_1 \cos^2 \left(\frac{\omega t}{2} \right) = \frac{1 + \cos(\omega t)}{2} Z_2$

$$Z_1 = Z_2 \frac{1 + \cos(\omega t)}{2 \cos^2 \left(\frac{\omega t}{2} \right)} \rightarrow Z_1 = Z_2 \frac{1 + \cos(\omega t)}{1 + \cos(\omega t)}$$

Now $Z_1 = 20 \Omega$ and $Z_2 = 40 \Omega$ (see Ex. 11.12), we have

$$40 \cos^2 \left(\frac{\omega t}{2} \right) = Z_2 \frac{1 + \cos(\omega t)}{1 + \cos(\omega t)} \rightarrow \begin{cases} 40 \cos^2 \omega t = 20 Z_1 \\ 20 Z_1 = 20 \cos^2 \omega t \end{cases}$$

$$\therefore Z_1 = 20 \cos^2 \omega t, \quad \phi = \cos^2 \omega t = 0.5 \cos^2 \omega t \rightarrow Z = 20 \cos^2 \omega t$$

Ex 11.14 (a) $|v| = \frac{1}{2}|v| = \frac{1}{2}|v| = \frac{1}{2}$

$$\text{Eq. (9-113a): } v(t) = \frac{1}{2}(V_1 + V_2)e^{j\omega t} [1 + \cos(2\omega t)]$$

$$\text{Eq. (9-114): } V = \frac{V_1 - V_2}{2} = |V|e^{j\phi}, \quad \phi = \alpha_2 - \alpha_1$$

$$\text{Perhaps } \alpha_1 \text{ is a minimum when } \phi = \pi \rightarrow Z_1 = Z_2 \frac{1 - \cos(\omega t)}{1 - \cos(\omega t)}$$

$$\therefore Z = \frac{1}{2} e^{j\omega t}$$

$$\text{(b) } Z_1 = Z_2 \left[\frac{1 + \cos(\omega t)}{1 + \cos(\omega t)} \right] = 40 \cos^2 \omega t = 20 \cos^2 \omega t$$

$$\text{(c) Terminating resistance } Z_1 = \frac{1}{2} = \frac{1}{2} \Omega = 100 \text{ m}\Omega$$

$$Z_2 = \frac{1}{2} \cos^2 \omega t = 20 \cos^2 \omega t = 20 \Omega$$

$$\text{Another set of solutions: } Z_1 = 20 \cos^2 \omega t \text{ and } Z_2 = 40 \cos^2 \omega t$$

Ex 2.21 $\lim_{x \rightarrow \infty} (x^2 - 2x) = \infty$ $\frac{\infty}{\infty}$ Use L'Hopital's.

Let $u = \frac{1}{x}$, $v = \frac{1}{x^2}$, $w = \frac{1}{x^3}$, and substitute.

$$x^2 - 2x = \frac{1}{u^2} - \frac{2}{u} = \frac{1 - 2u}{u^2} \implies \begin{cases} \lim_{u \rightarrow 0} (1 - 2u) = 1 \\ \lim_{u \rightarrow 0} u^2 = 0 \end{cases}$$

We have

$$L_1 = \lim_{x \rightarrow \infty} \left[(1 - 2x^{-1}) \sqrt{x^2 - 2x + 1} \right]$$

$$= \lim_{x \rightarrow \infty} \left[(1 - 2x^{-1}) \sqrt{x^2 - 2x + 1} \right]$$

$$L_2 = \lim_{x \rightarrow \infty} x^2$$

Ex 2.22 $L_1 = L_2 = \frac{1}{2}$

$$f = x^2 e^{2x}, \quad f' = \frac{d}{dx} x^2 e^{2x}, \quad g = \frac{1}{2} x^2 e^{2x}$$

$$\therefore L_1 = \lim_{x \rightarrow \infty} \frac{f'(x) g(x) - f(x) g'(x)}{g(x)^2 - g'(x)^2}$$

$$= \lim_{x \rightarrow \infty} \frac{2x e^{2x} \cdot \frac{1}{2} x^2 e^{2x} - x^2 e^{2x} \cdot x e^{2x}}{\left(\frac{1}{2} x^2 e^{2x}\right)^2 - (x e^{2x})^2}$$

$$= \lim_{x \rightarrow \infty} \frac{x^3 e^{4x} - x^3 e^{4x}}{\frac{1}{4} x^4 e^{4x} - x^2 e^{4x}}$$

Ex 2.23 (i) Given: $u_1 = 2x^2 - 3x$, $u_2 = 2x^2 + 4x - 3$, $u_3 = 2x^2 - 2x + 3$

$$v_1 = \frac{1}{x^2} e^{-x}, \quad v_2 = \frac{1}{x^2} e^x$$

$$\text{where } u_1 = \frac{d}{dx} \left(\frac{1}{x^2} e^{-x} \right) = \frac{d}{dx} \left(\frac{1}{x^2} e^{-x} \right)$$

$$\therefore v_1 = \frac{1}{x^2} e^{-x} \implies v_1 = \frac{1}{x^2} e^{-x} \implies \text{or}$$

$$v_2 = \frac{1}{x^2} e^x \implies v_2 = \frac{1}{x^2} e^x \implies \text{or}$$

Putting $u_1 = v_1$ and $u_2 = v_2$ in the first and second

$$\text{we have } v_1 = 2x^2 - 3x = \frac{1}{x^2} e^{-x} \implies v_1 = \frac{1}{x^2} e^{-x}$$

$$= \frac{1}{x^2} e^{-x} \implies \text{or}$$

$$v_2 = 2x^2 + 4x - 3 = \frac{1}{x^2} e^x \implies v_2 = \frac{1}{x^2} e^x \implies \text{or}$$

$$\text{d) } \beta = \frac{1+2\beta^2}{1-\beta^2} = 2.$$

$$\begin{aligned} \text{e) } (R_{2n})_1 &= \int_0^1 R_n(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^n + \frac{1}{2} \left(\frac{1}{2} - x \right)^n \right) dx \\ &= \frac{1}{2} \int_0^1 \left(\frac{1}{2} + x \right)^n dx + \frac{1}{2} \int_0^1 \left(\frac{1}{2} - x \right)^n dx \\ &= \frac{1}{2} \left[\frac{1}{n+1} \left(\frac{1}{2} + x \right)^{n+1} \right]_0^1 + \frac{1}{2} \left[-\frac{1}{n+1} \left(\frac{1}{2} - x \right)^{n+1} \right]_0^1 \\ &= \frac{1}{2(n+1)} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right) - \frac{1}{2(n+1)} \left(\left(\frac{1}{2} \right)^{n+1} - \left(\frac{3}{2} \right)^{n+1} \right) \\ &= \frac{1}{n+1} \left(\frac{3}{2} \right)^{n+1} = \frac{3^{n+1}}{2^{n+1}(n+1)}. \end{aligned}$$

$$\text{E.2.21} \quad \begin{aligned} \text{a) } \int_0^1 x^2 dx &= \frac{1}{3} x^3 \Big|_0^1 = \frac{1}{3}, \quad \int_0^1 x^3 dx = \frac{1}{4} x^4 \Big|_0^1 = \frac{1}{4}, \\ \int_0^1 x^4 dx &= \frac{1}{5} x^5 \Big|_0^1 = \frac{1}{5}, \quad \int_0^1 x^6 dx = \frac{1}{7} x^7 \Big|_0^1 = \frac{1}{7}. \end{aligned}$$

$$\text{b) } R_{2n} = \int_0^1 R_n(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^n + \frac{1}{2} \left(\frac{1}{2} - x \right)^n \right) dx.$$

$$\begin{aligned} \text{c) } R_n &= \int_0^1 R_{n-1}(x) dx = \int_0^1 \left(\frac{1}{2} \left(\frac{1}{2} + x \right)^{n-1} + \frac{1}{2} \left(\frac{1}{2} - x \right)^{n-1} \right) dx \\ &= \frac{1}{2} \int_0^1 \left(\frac{1}{2} + x \right)^{n-1} dx + \frac{1}{2} \int_0^1 \left(\frac{1}{2} - x \right)^{n-1} dx \\ &= \frac{1}{2(n)} \left(\left(\frac{3}{2} \right)^n - \left(\frac{1}{2} \right)^n \right) - \frac{1}{2(n)} \left(\left(\frac{1}{2} \right)^n - \left(\frac{3}{2} \right)^n \right) \\ &= \frac{1}{n} \left(\frac{3}{2} \right)^n = \frac{3^n}{2^n n}. \end{aligned}$$

$$\text{d) } \frac{R_n}{n} = \frac{1}{n} \left(\frac{3}{2} \right)^n = \frac{1}{n} \left(\frac{3}{2} \right)^{n-1} = \frac{1}{n} \frac{R_{n-1}}{n-1}.$$

$$\begin{aligned} \text{e) } R_{2n} &= \frac{1}{2(n+1)} \left(\left(\frac{3}{2} \right)^{n+1} - \left(\frac{1}{2} \right)^{n+1} \right) - \frac{1}{2(n+1)} \left(\left(\frac{1}{2} \right)^{n+1} - \left(\frac{3}{2} \right)^{n+1} \right) \\ &= \frac{1}{n+1} \left(\frac{3}{2} \right)^{n+1} = \frac{3^{n+1}}{2^{n+1}(n+1)}, \quad R_n = \frac{3^n}{2^n n} \\ &= \frac{1}{n} \left(\frac{3}{2} \right)^n = \frac{3^n}{2^n n}, \quad R_{2n} = \frac{3^{2n}}{2^{2n}(2n)} = \frac{3^{2n}}{2^{2n} 2n} \\ &= \frac{1}{2} \left(\frac{3}{2} \right)^{2n} = \frac{1}{2} \left(\frac{3}{2} \right)^n \left(\frac{3}{2} \right)^n = \frac{1}{2} \left(\frac{3}{2} \right)^n \frac{3^n}{2^n} = \frac{1}{2} \left(\frac{3}{2} \right)^n \frac{R_n}{n} \end{aligned}$$

$$\text{E.2.22} \quad \begin{aligned} \text{a) } \int_0^1 x^2 dx &= \frac{1}{3} x^3 \Big|_0^1 = \frac{1}{3}, \quad \int_0^1 x^3 dx = \frac{1}{4} x^4 \Big|_0^1 = \frac{1}{4}, \\ \int_0^1 x^4 dx &= \frac{1}{5} x^5 \Big|_0^1 = \frac{1}{5}, \quad \int_0^1 x^6 dx = \frac{1}{7} x^7 \Big|_0^1 = \frac{1}{7}. \end{aligned}$$

$$\begin{aligned} \text{b) } \int_0^1 x^2 dx &= \frac{1}{3} x^3 \Big|_0^1 = \frac{1}{3}, \quad \int_0^1 x^3 dx = \frac{1}{4} x^4 \Big|_0^1 = \frac{1}{4}, \\ \int_0^1 x^4 dx &= \frac{1}{5} x^5 \Big|_0^1 = \frac{1}{5}, \quad \int_0^1 x^6 dx = \frac{1}{7} x^7 \Big|_0^1 = \frac{1}{7}. \end{aligned}$$

$$\begin{aligned} \text{c) } \int_0^1 x^2 dx &= \frac{1}{3} x^3 \Big|_0^1 = \frac{1}{3}, \quad \int_0^1 x^3 dx = \frac{1}{4} x^4 \Big|_0^1 = \frac{1}{4}, \\ \int_0^1 x^4 dx &= \frac{1}{5} x^5 \Big|_0^1 = \frac{1}{5}, \quad \int_0^1 x^6 dx = \frac{1}{7} x^7 \Big|_0^1 = \frac{1}{7}. \end{aligned}$$

d) At the limit, $n \rightarrow \infty$.

$$\begin{aligned} \int_0^1 x^2 dx &= \frac{1}{3} x^3 \Big|_0^1 = \frac{1}{3}, \\ \int_0^1 x^3 dx &= \frac{1}{4} x^4 \Big|_0^1 = \frac{1}{4}, \\ \int_0^1 x^4 dx &= \frac{1}{5} x^5 \Big|_0^1 = \frac{1}{5}, \\ \int_0^1 x^6 dx &= \frac{1}{7} x^7 \Big|_0^1 = \frac{1}{7}. \end{aligned}$$

$$\int_0^1 x^2 dx = \frac{1}{3}, \quad \int_0^1 x^3 dx = \frac{1}{4},$$

$$\int_0^1 x^4 dx = \frac{1}{5}, \quad \int_0^1 x^6 dx = \frac{1}{7}.$$

2.2-22 $E_1 = 0$, $E_2 = 1$

(a) $0 < t < T/2$



(b) $T/2 < t < T$



(c) $T < t < 3T/2$



(d) $3T/2 < t < 2T$



$$V_1^+ = V_1, \quad V_1^- = V_1^+ = V_1$$

$$V_2^+ = V_1^+ = V_1$$

$$V_2^- = V_1^- = V_1$$

$$V_1^+ = V_1 V_2, \quad V_1^- = -V_1^+ = -V_1 V_2$$

$$V_2^+ = V_1^- = -V_1 V_2$$

$$V_2^- = V_2^+ = V_1 V_2$$

At $t=2T$, both V_1 and V_2 revert back to the initial state of 0, and the cycle repeats itself with a period $2T$.



At the connecting points of two transmission lines with different characteristic impedances Z_0 and Z_0'

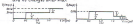
$$\begin{array}{l} \frac{V_1}{Z_0} = \frac{V_2}{Z_0} \\ Z_0 = Z_0' \end{array} \quad \begin{array}{l} V_1 = V_2 \\ Z_0 = Z_0' \end{array} \quad \begin{array}{l} V_1' + V_2' = V_1 \\ Z_0' = Z_0 \end{array} \quad \begin{array}{l} V_1' = V_2' = V_1 \\ Z_0' = Z_0 \end{array}$$

$$\begin{array}{l} \text{Solving } V_1' = \frac{Z_0' - Z_0}{Z_0' + Z_0} V_1 \\ Z_0' = Z_0 \end{array} \quad \begin{array}{l} V_2' = \frac{2Z_0}{Z_0' + Z_0} V_1 \\ Z_0' = Z_0 \end{array}$$

$$\begin{array}{l} a) \quad V_1' = \frac{50 - 50}{50 + 50} V_1 = 0 \text{ (no)} \\ Z_0' = 50 \\ V_2' = \frac{2 \cdot 50}{50 + 50} V_1 = 200 \text{ (no)} \\ Z_0' = 50 \\ V_1' = \frac{50 - 50}{50 + 50} V_1 = 0 \text{ (no)} \\ Z_0' = 50 \\ V_2' = \frac{2 \cdot 50}{50 + 50} V_1 = 200 \text{ (no)} \\ Z_0' = 50 \end{array}$$

No transient waves on the parallel cable after V_1' and V_2' reach the input terminated at $t_1 = 200/c = 1 \text{ ns}$, and no transient waves on the series line after V_1' and V_2' reach the load $t_2 = 1 \text{ ns}$, $t_3 = 1 \text{ ns} + 200/c = 2 \text{ ns}$, and $t_4 = 2 \text{ ns} + 200/c = 3 \text{ ns}$.

b) On the parallel cable (1) after $t_1 = 1 \text{ ns}$ for V_1' and V_2' to reach the output ($x = 400$). The reflected waves V_1' and V_2' arrive at the input at $t = 1 \text{ ns} + 200/c = 2 \text{ ns}$. There are no changes after that.



On the series line, steady state is reached at $t = 2 \text{ ns}$.



Beispiel 1 $R_1 = R_2 = 1 \rightarrow R_1 = 1 \rightarrow R_2 = 1 \rightarrow R_1 = 1 \rightarrow R_2 = 1$ $T = 1$ s

a) Voltage collector diagram



$$V_1 = \frac{R_1}{R_1 + R_2} V_{in}$$

$$= \frac{1}{2} V_{in}$$

$$V_2 = \frac{R_2}{R_1 + R_2} V_{in}$$

$$= \frac{1}{2} V_{in}$$

Current collector diagram



Beispiel 2 $R_1 = 1 \rightarrow R_2 = 1, R_1 = 1 \rightarrow R_2 = 1$ $T = 1$ s

a) Voltage collector diagram



Current collector diagram



c)



Ex 2.12 The current reflection diagram for Example 9-17 is



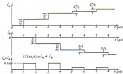
$$\Gamma_L = \frac{1}{2} = \Gamma_S = 1$$

$$\Gamma = 2/3 \text{ (red)}$$

Indices on the directed lines are normalized with respect to

$$\Gamma^2 = \frac{1^2}{2^2} = \frac{1}{4}$$

$$= 4/16 \text{ (red)}$$



Ex 2.11 Use the equivalent circuit in Fig. 2.11(b) to study transient voltage and currents:



(a) Amplitude of first current wave (starting from

$$t=0 \text{ to } t=0.1) \quad I_1^* = \frac{V}{R_1 + R_2} = \frac{10}{10} = 1$$

Refer to Fig. 2.11(a) $L_1 = -L_2$ $V_1(t) = 10u(t)$

$$L = 10L_1 \rightarrow \tau = \frac{10L_1}{R_1 + R_2} = \frac{1}{2} \quad \tau = 1 \quad \Gamma = 0.5$$



$$L = \frac{1}{2} \rightarrow \tau = \frac{10L_1}{R_1 + R_2} = \frac{1}{2} \quad \tau = 1 \quad \Gamma = 0.5$$

Ex 2.12 (a) Governing equation of the load for $t > 0$

$$L_2 \frac{di(t)}{dt} + (R_2 + R_1)i(t) = 10V$$

$$\text{Solution: } i(t) = \frac{10V}{R_1 + R_2} \left[1 - e^{-\frac{R_1 + R_2}{L_2}(t - \tau)} \right] \quad t > \tau$$

For the present problem, $V = 10V$, $R_1 = R_2 = 10\Omega$, $L_2 = 10mH$

$$L_2 = 10 \times 10^{-3} \text{ (H)} \quad \tau = L_2 / R_2 = 100 \times 10^{-3} / 10 = 10 \text{ (ms)}$$

$$i(t) = \frac{10}{20} \left[1 - e^{-\frac{20}{10 \times 10^{-3}}(t - 10 \times 10^{-3})} \right] \quad t > 10 \text{ (ms)}$$

$$i(t) = 0.5 \left[1 - e^{-200(t - 10 \times 10^{-3})} \right] \quad t > 10 \text{ (ms)}$$



$$\text{At } t = 10 \text{ (ms)} \\ i(t) = \frac{10}{20} \left[1 - e^{-200(t - 10 \times 10^{-3})} \right] \\ = 0.5 \text{ (A)}$$



At $x = \pi^+$ point
 $\lim_{x \rightarrow \pi^+} f(x) = 1 - e^{-2(\pi - \pi)} = 1$
 $\neq 1 = f(\pi)$

Ex-10 From Eq (1)-(2) $\lim_{x \rightarrow 0} (2x^2 - 3x) = 0$ (1)

At the limit $\lim_{x \rightarrow 0} \frac{2x^2}{x^2} = \lim_{x \rightarrow 0} \frac{2x^2}{x^2}$ (2)

Substituting (2) in (1) $\lim_{x \rightarrow 0} \frac{2x^2}{x^2} = \left(\frac{2}{x^2} - \frac{3}{x^2}\right) \lim_{x \rightarrow 0} x^2 = \frac{1}{x^2} \lim_{x \rightarrow 0} x^2$ (3)

(3) Solution of (3): $\lim_{x \rightarrow 0} (2x^2 - 3x) = \lim_{x \rightarrow 0} \frac{2x^2}{x^2} \left[1 - e^{-2(\pi - \pi)}\right]$ (4)

For this problem $\lim_{x \rightarrow 0} \frac{2x^2}{x^2} = 2$ (5) $\lim_{x \rightarrow 0} (1 - e^{-2(\pi - \pi)}) = 1$

$T = 2\pi$, $\lim_{x \rightarrow 0} \frac{2x^2}{x^2} = 2$, $\lim_{x \rightarrow 0} (1 - e^{-2(\pi - \pi)}) = 1$

$\lim_{x \rightarrow 0} (2x^2 - 3x) = 2 \lim_{x \rightarrow 0} \left[1 - e^{-2(\pi - \pi)}\right] = 2 \lim_{x \rightarrow 0} 1 = 2$

From (5): $\lim_{x \rightarrow 0} (2x^2 - 3x) = 2 \lim_{x \rightarrow 0} (1 - e^{-2(\pi - \pi)}) = 2$ (6)

(6) At $x = \pi^+$, $\lim_{x \rightarrow \pi^+} (2x^2 - 3x) = 2 \lim_{x \rightarrow \pi^+} (1 - e^{-2(\pi - \pi)}) = 2$



Ex-11 (a) $\lim_{x \rightarrow 0} (2x - 3x) = \lim_{x \rightarrow 0} (2x) e^{2x} = 0$ since $\lim_{x \rightarrow 0} 2x = 0$, $\lim_{x \rightarrow 0} e^{2x} = 1$

$\lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} = \lim_{x \rightarrow 0} \frac{2x}{x^2} e^{2x} = \lim_{x \rightarrow 0} \frac{2}{x} e^{2x}$ since $\lim_{x \rightarrow 0} \frac{2}{x} = \infty$, $\lim_{x \rightarrow 0} e^{2x} = 1$

$\therefore \lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} = \lim_{x \rightarrow 0} \frac{2}{x} e^{2x} = \lim_{x \rightarrow 0} \frac{2}{x} e^{2x} = \lim_{x \rightarrow 0} \frac{2}{x} e^{2x}$ since $\lim_{x \rightarrow 0} \frac{2}{x} = \infty$, $\lim_{x \rightarrow 0} e^{2x} = 1$

(b) $\lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} = \lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} e^{2x} = \lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} e^{2x} = \lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} e^{2x}$ since $\lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} = \infty$, $\lim_{x \rightarrow 0} e^{2x} = 1$

(c) $\lim_{x \rightarrow 0} \frac{2x - 3x}{x^2} = 0$ since $\lim_{x \rightarrow 0} (2x - 3x) = 0$, $\lim_{x \rightarrow 0} \frac{1}{x^2} = \infty$
 $\lim_{x \rightarrow 0} (2x - 3x) = 0 = \lim_{x \rightarrow 0} (2x - 3x) = 0$

$$P(2) = \frac{1}{2} \cos^2(180^\circ) = \frac{1}{2} \cos^2(0^\circ)$$

a) Quarter-circular line, $d = 100 \text{ mm}$, $d/2 = 50 \text{ mm}$

Start at the extreme-left point P_1 on the extreme circle, rotate clockwise one complete revolution (360° or 2π) and continue on P_2 an additional 180° to arrive on the "unwrapped" straight generator "line". Read $d = 100 \text{ mm} \rightarrow \xi = 2\pi \times (100/2) = 314.16 \text{ mm}$.

Draw a straight line from the (0,0,0) point through the center and intercept at (0,314,0) on the opposite side of the sheet $\rightarrow \xi = \frac{1}{2} \times (314.16) = 157.08 \text{ mm}$.

b) Semi-circular line, $d = 100 \text{ mm}$, $d/2 = 50 \text{ mm}$

Start from the extreme-left point P_1 , rotate clockwise two complete revolutions and continue on the arc another 180° to read $d = 100 \text{ mm} \rightarrow \xi = 2\pi \times (100/2) = 314.16 \text{ mm}$.

Draw a straight line from the (0,0,0) point through the center and intercept at (0,314,0) on the opposite side of the sheet $\rightarrow \xi = \frac{1}{2} \times (314.16) = 157.08 \text{ mm}$.

Example



$$\xi = \frac{1}{2} (314.16) = 157.08 \text{ mm}$$

a) 1. Locate $\xi = 157.08 \text{ mm}$ on the sheet (Point P_1)

2. Draw center of P above a 90° line through P_1 , intercepting ξ_1 at 157. $\rightarrow \xi = 157$.

$$P = \frac{1}{2} \cos^2(90^\circ) = \frac{1}{2} \cos^2(0^\circ)$$

a) 1. Draw line ξ_1 , intercepting the periphery of P' .

Read 0.000 on "unwrapped" straight generator "line".

2. Move clockwise by ξ_1 to start (Point ξ').

3. Draw ξ and ξ' , intercepting the 0° circle of P' .

4. Read $\xi = 157.08 \text{ mm}$ of ξ_1 .

$$\xi_1 = 2\pi \xi = 2\pi \times 157.08 \text{ mm}$$

4) External line $Q_1'Q_2'$ to Q_2 . Arcs $Q_1'Q_2' = 2.000 - 2.000 \sin 20^\circ$.

$$Q_1'Q_2' = \frac{1}{2} Q_1'Q_2' = 0.000 - 2.000 \sin 20^\circ$$

Q There is no voltage minimum on this line, but $Q_1'Q_2'$.

Example



$$Z_L = j100 (1 - j100) = 0.000 - j100$$

4) Draw $Q_1'Q_2'$ at $2.000 - j100$ on Smith chart (Point Q_1'). With center at O' draw a circle through Q_1' , intersecting line 0.25λ at Q_2' . ——— $d = 0.25\lambda$.

$$d) P = 0.125 \sin^2 20^\circ$$

4) Draw line 0.25λ , intersecting the periphery at Q_1' . Arcs 0.25λ to "normalize the forward power" scale.

1. Draw clockwise by 0.25λ to 0.000 (Point Q_1').

2. Draw 0.25λ and Q_1' , intersecting the (V) scale at Q_2' .

3. Arcs $d_1 = 0.000 - j100$ at Q_1' .

$$Q_1'Q_2'Q_1 = \sin^2 \alpha \sin^2 \beta \sin^2 \gamma$$

4) External line $Q_1'Q_2'$ to Q_2 . Arcs $Q_1'Q_2' = 2.000 - 2.000 \sin 20^\circ$.

$$Q_1'Q_2' = \frac{1}{2} Q_1'Q_2' = 0.000 - 2.000 \sin 20^\circ$$

Q There is a voltage minimum at $Q_1'Q_2' = 0.000 - 2.000$.

Example $\lambda/4 = 20$, $\lambda = 80$ mm.

First voltage minimum occurs at $Q_1'Q_2' = \frac{1}{2} Q_1'Q_2' = 0.125$.



4) Start from Q_1' and rotate counterclockwise 0.25λ toward the load to Q_1' .

1. Draw the (V) circle, intersecting line 0.25λ at Q_2' (Point).

2. Draw 0.25λ , intersecting the (V) circle at Q_1' .

4. Draw $\alpha_1 = \cos t - j \sin t$.

$$\alpha_1 = \cos t - j \sin t = e^{-j t} \quad \text{or}$$

5) $P = \frac{d^2}{dt^2} P = \alpha_1 e^{j 200 t}$

6) If $\alpha_2 = 0$, the first voltage minimum would be at $\alpha_2 = \alpha_1 \alpha_1 = 1$ (not from the short-circuit).

Example



a) $\alpha_1 = \frac{1}{\sqrt{2}} (\cos - j \sin)$
 $= \cos - j \sin$

1. Draw α_1 as a unit vector (Point P).

2. Take α and β , and extend to K.

3. Draw an isosceles triangle formed geometrically under α_1 and β .

$$\beta = \alpha_1 \alpha_1 = 1 \quad \text{and} \quad \alpha = \beta \alpha_1 = \alpha_1 \quad \text{or} \quad \alpha = \alpha_1$$

$$\frac{d^2}{dt^2} = 1 \quad \text{or} \quad \alpha = \frac{1}{\sqrt{2}} (\cos - j \sin) = \alpha_1 \quad \text{or} \quad \alpha = \alpha_1$$



b) 1. Draw α_1 as a unit vector (Point P).

2. Draw the line α through P to Q. Draw an isosceles triangle geometrically under α_1 and β .

3. Draw a vector α_2 as a unit vector (Point P).

4. Take α_2 intersecting the (1)° circle through Q at K.

5. Mark point P on the (1)° circle that $\frac{d^2}{dt^2} = \alpha_1$.

6. Draw at P: $\alpha = \alpha_1 \alpha_1 = 1$ or $\alpha = \alpha_1$ (Point P).

Q) 1. Show clockwise from \mathbb{R}^2 an "orthogonal" vector generator" leads to \mathbb{R}^2 , say P'

2. Give 90°

3. Show point P on line \mathbb{R}^2 such that

$$\vec{OP} = e^{i\pi/2} \vec{OP} = i \text{rot } \vec{OP}$$

4. Show at P : $\vec{v} = e^{i\pi/2} \vec{v} \text{rot} \rightarrow \vec{v} = e^{i\pi/2} \text{rot}(\vec{v})$.

Ex: $\vec{v} = 2 + 3i \text{ rot}$, $\vec{v} = 3 + 2i \text{ rot} \rightarrow \vec{v} = \frac{1}{\sqrt{2}} \text{rot}(2+3i) \text{ rot}$

$$\vec{v} = \frac{1}{\sqrt{2}} (2+3i) \text{ rot}$$

For $\vec{v} = 2 + 3i \text{ rot}$ then $\vec{v} = \frac{1}{\sqrt{2}} \text{rot}(2+3i) \text{ rot}$

$$\vec{v} = 2 + 3i \text{ rot} \rightarrow \vec{v} = 3 + 2i \text{ rot}$$

Ex: 2



$$\vec{v} = 2 + 3i \text{ rot}$$

$$\vec{v} = 3 + 2i \text{ rot}$$

Q) For $\vec{v} = 2 + 3i \text{ rot}$

$$\vec{v} = 2 + 3i \text{ rot}$$

$$\vec{v} = 3 + 2i \text{ rot} = \vec{v} \text{ rot}$$

$$\vec{v} = 2 + 3i \text{ rot}$$

$$\vec{v} = 3 + 2i \text{ rot} = \vec{v} \text{ rot}$$

$$\vec{v} = 2 + 3i \text{ rot} = \vec{v} \text{ rot}$$

$$\vec{v} = 3 + 2i \text{ rot} = \vec{v} \text{ rot}$$

Q) For $\vec{v} = 2 + 3i \text{ rot}$, $\vec{v} = 3 + 2i \text{ rot}$

The required vectors are $\vec{v} = \frac{1}{\sqrt{2}} \text{rot}(2+3i) \text{ rot}$

	$\vec{v} = \frac{1}{\sqrt{2}} \text{rot}(2+3i) \text{ rot}$	$\vec{v} = \frac{1}{\sqrt{2}} \text{rot}(3+2i) \text{ rot}$
$\vec{v} = 2 + 3i \text{ rot}$	$\vec{v} = 2$, $\vec{v} = 3i \text{ rot}$	$\vec{v} = 3$, $\vec{v} = 2i \text{ rot}$
$\vec{v} = 3 + 2i \text{ rot}$	$\vec{v} = 3i \text{ rot}$, $\vec{v} = 2 \text{ rot}$	$\vec{v} = 2i \text{ rot}$, $\vec{v} = 3 \text{ rot}$

Ex. 10.11 $\alpha = \beta = \gamma = \pi/2$

Use Jacobi about an independent orbit. Some restriction as that in problem 10-10 except R_{22} would be in the column \mathbb{R}^3 (instead of \mathbb{R}^2) and \mathbb{R}^3 is the same as \mathbb{R}^2 .

$$\xi_1: \xi_1 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3, \quad \xi_2: \xi_2 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ with } \alpha = \beta = \gamma = \pi/2.$$

$$\xi_3: \xi_3 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ with } \alpha = \beta = \gamma = \pi/2.$$

To obtain a matrix with a vector that having $\xi_1 = \mathbb{R}^3$, we need a normalized orbit decomposition of $\mathbb{R}^3 = \mathbb{R}^3$ for the solution corresponding to ξ_1 . From Jacobi about an orbit the required orbit length $R_{22} = \alpha \mathbb{R}^3$.

Similarly, for solution corresponding to ξ_2 , a orbit with a normalized decomposition of \mathbb{R}^3 is needed, which requires a orbit length $R_{22} = \alpha \mathbb{R}^3$.

Ex. 10.12



$$\xi_1 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3.$$

$$\xi_2: \xi_2 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ (with } \alpha = \beta \text{)}$$

$$\xi_3: \xi_3 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ (with } \alpha = \beta \text{)}$$

$$\xi_4: \xi_4 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ (with } \alpha = \beta \text{)}$$

$$\xi_5: \xi_5 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ (with } \alpha = \beta \text{)}$$

$$\xi_6: \xi_6 = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3 \text{ (with } \alpha = \beta \text{)}$$

of \mathbb{R}^3 (with $\alpha = \beta$)

	of \mathbb{R}^3 (with $\alpha = \beta$)	of \mathbb{R}^3 (with $\alpha = \beta$)
$(\xi_1) = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$
$(\xi_2) = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$
$(\xi_3) = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$
$(\xi_4) = \alpha \mathbb{R}^3 + \beta \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$	$R_{22} = \alpha \mathbb{R}^3$

Ex-20



$$R_1 = \frac{Rr}{R+r} = R \cos \alpha \cos \beta$$

Point Q_1 on small circle
(radius of r)

Since the rotated gear circle is tangent to the gear circle, an added line length d_1 is needed to connect Q_1 (radius r),

moving from Q_1 along the (r)-circle to Q_2 (radius of the gear circle) (radius of R). Note that Q_2 is different from Q , the point of tangency between the gear and rotated gear circles.

$$d_1 = R_1 \cos \alpha + r \cos \beta + R \cos \alpha$$

$$R_2 = r_2 = r \cos \beta \quad (\text{radius of } r)$$

$$R_3 = R_2 = R \cos \alpha \quad (\text{radius of } R)$$

$$R_4 = R_2 - R_3 = (1 - \beta^2) - (1 - \alpha^2) = \alpha^2 - \beta^2 = R_2 \cos \alpha$$

$$R_5 = r_2 \cos \beta = R_2 \cos \beta$$

Ex-21 Let $d = \beta d_1 = \frac{Rr}{R+r} d_1$

Report: $d_1 = \frac{d}{\cos \alpha \cos \beta}$ (Analytical solution)

d_1	d	no. solutions
$r/2$	2.11°	0.00
$r/3$	4.1°	0.00
$r/4$	6.1°	2.00
$2r/3$	19.1°	2.00
$r/2.5$	11.1°	0.00

¹ See B. P. Chang and C. H. Liang, "Computer Solution of Double-Link Inverse Kinematic Problems," *IEEE Transactions on Education*, vol. E-31, pp. 107-111, November 1988.

Chapter 11

Waveguides and Cavity Resonators

Ex 11-1 $\nabla \times \mathbf{E} = -j\omega\mu\mathbf{H}$ (1)
 $\nabla \times \mathbf{H} = j\omega\epsilon\mathbf{E}$ (2)

From (1) $(\mathbf{E}_x + \mathbf{E}_y \frac{\partial}{\partial y}) \times (\mathbf{E}_z + \mathbf{E}_y \frac{\partial}{\partial y}) = -j\omega\mu(\mathbf{E}_x + \mathbf{E}_y \frac{\partial}{\partial y})$

———— $\mathbf{E}_y \times (\mathbf{E}_y \frac{\partial}{\partial y}) = \nabla \times \mathbf{E}_y = -j\omega\mu\mathbf{E}_y$

———— $\mathbf{E}_x \times \mathbf{E}_y = \nabla \times \mathbf{E}_x = -j\omega\mu\mathbf{E}_x$ (3)
($\because \mathbf{E}_x \times (\mathbf{E}_y \frac{\partial}{\partial y}) = \mathbf{E}_x \times \mathbf{E}_y$)

Similarly from (2) we obtain

$\mathbf{E}_x \times \mathbf{E}_y = \nabla \times \mathbf{E}_y = j\omega\epsilon\mathbf{E}_y$ (4)

Combining (3) and (4), we have

$j\omega\mu\mathbf{E}_x = (\mathbf{E}_x \times \mathbf{E}_y) = \nabla \times \mathbf{E}_y = j\omega\epsilon\mathbf{E}_y$

———— $\mathbf{E}_y = -\frac{\mu}{\epsilon} (\nabla \times \mathbf{E}_x) = -\frac{\mu}{\epsilon} j\omega\mu\mathbf{E}_x$ (5)

Similarly, $\mathbf{E}_x = -\frac{\epsilon}{\mu} (\nabla \times \mathbf{E}_y) = -\frac{\epsilon}{\mu} j\omega\epsilon\mathbf{E}_y$ (6)

Ex 11-2



From Eq. (11-28a):

$$\left(\frac{\beta}{k_0}\right)^2 = \left(\frac{\omega}{\omega_c}\right)^2 - 1$$

From Eq. (11-28c):

$$\left(\frac{\alpha}{k_0}\right)^2 = \left(\frac{\omega}{\omega_c}\right)^2 - 1$$

Both are equivalent to a unit circle.

(b)



From Eq. (11-28a):

$$\left(\frac{\beta}{k_0}\right)^2 = 1 - \frac{1}{\left(\frac{\omega}{\omega_c}\right)^2}$$

From Eq. (11-28c):

$$\left(\frac{\alpha}{k_0}\right)^2 = 1 - \frac{1}{\left(\frac{\omega}{\omega_c}\right)^2}$$



From Eq. (10-11):

$$\left(\frac{d\phi}{dr}\right) = \frac{(\partial\phi/\partial r)}{(r/r_0)^2} = f$$

$$\partial \text{At } r/r_0 = 1.00:$$

$$\phi_0/r_0 = 1.00,$$

$$\phi_0/r_0 = 1.00,$$

$$\phi_0/r_0 = 1.00,$$

$$\phi_0/r_0 = 1.00.$$

Ex. 10-11 a) For parallel-plate arrangement:



$$\phi(r) - \phi_0 = \left(\frac{q\sigma}{\epsilon_0}\right)r$$

$$\phi_0 = \frac{q\sigma}{\epsilon_0}r_0$$

$$\phi_0 = \frac{q\sigma}{\epsilon_0}r_0$$

$$\phi_0 = \frac{q\sigma}{\epsilon_0}r_0$$

Displacement parameter \$r\$ and \$r_0\$ must both be used, and the slope of the \$\phi(r)\$ curves, is directly \$q\$.

But not the slope of \$\phi(r)\$ depends on \$q\$.

Ex. 10-11 b) Field equations for 3D media, from Eqs. (10-12) and (10-13):

$$\nabla^2 \phi = -\rho_0 \text{ (charge)}$$

$$\nabla^2 \phi = -\frac{\rho_0}{\epsilon_0} \text{ (charge)}$$

$$\nabla^2 \phi = -\frac{\rho_0}{\epsilon_0} \text{ (charge)}$$

Surface charge densities:

$$\phi_0 = \phi_0 \cdot \delta \int_{\text{area}} = \phi_0 \cdot \delta \int_{\text{area}} = -\frac{\rho_0}{\epsilon_0} \delta \int_{\text{area}}$$

$$\phi_0 = \phi_0 \cdot \delta \int_{\text{area}} = -\rho_0 \delta \int_{\text{area}} = -\rho_0 \delta \int_{\text{area}}$$

Surface current densities:

$$\phi_0 = \phi_0 \cdot \delta \int_{\text{area}} = \phi_0 \cdot \delta \int_{\text{area}} = -\rho_0 \delta \int_{\text{area}}$$

$$\phi_0 = \phi_0 \cdot \delta \int_{\text{area}} = -\rho_0 \cdot \delta \int_{\text{area}} = \rho_0 \cdot \delta \int_{\text{area}} \frac{\rho_0}{\epsilon_0} \delta \int_{\text{area}}$$

Problem 8.101 Find expressions for \mathcal{H}_z inside, from Eq. (8.33) and (8.34):

$$\mathcal{H}_z^{\text{top}}(y) = \mathcal{H}_z \text{ inside top slab,}$$

$$\mathcal{H}_z^{\text{mid}}(y) = \frac{1}{2} \mathcal{H}_z \text{ inside middle slab,}$$

$$\mathcal{H}_z^{\text{bot}}(y) = -\frac{1}{2} \mathcal{H}_z \text{ inside bottom slab,}$$

$$\mathcal{H}_z = \mathcal{H}_y + \mathcal{H}(z) = \mathcal{H}_y \mathcal{H}_z,$$

$$\mathcal{H}_z = -\mathcal{H}_y + \mathcal{H}(z) = \mathcal{H}_y \cos^2 \theta \mathcal{H}_z \text{ if } \mathcal{H}_z \text{ dir. is odd,}$$

$$\mathcal{H}_z = \mathcal{H}_y \text{ dir. is even.}$$

Problem 8.102 Repeat part (a) of the field expressions in problem 8.101.



Repeat part (a) of the field expressions in problem 8.101.



Problem 8.103 Using the field expressions in problem 8.102, show:

$$\mathcal{E}_z = \int \mathcal{H}_z (1 + \epsilon^2) = \int \mathcal{H}_z (\mathcal{H}_z \mathcal{E}_z^2 - \mathcal{H}_y \mathcal{E}_z^2)$$

$$\mathcal{E}_z = \mathcal{E}_y = \int \mathcal{H}_z (\mathcal{E}_z^2 - \mathcal{E}_y^2) = \frac{1}{2} \mathcal{H}_z \mathcal{E}_z^2 \cos^2 \theta (\mathcal{E}_z^2)$$

$$(\mathcal{E}_z)_z = \int \mathcal{E}_z \mathcal{H}_z \mathcal{E}_z = \frac{1}{2} \mathcal{H}_z \mathcal{E}_z^2 \mathcal{E}_z^2 \quad (\text{per unit pole, inside})$$

$$(\mathcal{E}_z)_z = \frac{1}{2} \mathcal{H}_z (\mathcal{E}_z^2 - \mathcal{E}_y^2) = \frac{1}{2} \mathcal{H}_z \mathcal{E}_z^2 \mathcal{E}_z^2 \cos^2 \theta (\mathcal{E}_z^2)$$

$$(\mathcal{E}_z)_z = \int \mathcal{E}_z \mathcal{H}_z \mathcal{E}_z = \frac{1}{2} \mathcal{H}_z \mathcal{E}_z^2 \mathcal{E}_z^2 = \frac{1}{2} \mathcal{H}_z \mathcal{E}_z^2 \quad (\text{per unit pole, inside})$$

$$\text{From Eq. (8.33): } \mathcal{H}_z = \frac{\mathcal{E}_z \mathcal{H}_z \mathcal{E}_z}{\mathcal{E}_z \mathcal{H}_z \mathcal{E}_z} = \frac{1}{2} \mathcal{H}_z = \frac{1}{2} \sqrt{1 - \epsilon^2} \mathcal{H}_z$$

which is the same as Eq. (8.33)

Ex. 10 Given: $\beta = 2.00 \times 10^8 \text{ cm/s}$, $\nu = 1.00 \times 10^{14} \text{ s}^{-1}$, $\mu = 1$,
 $\rho = 1.00 \times 10^3 \text{ kg/m}^3$, $\lambda = 3.00 \times 10^8 \text{ cm}$, $f = 10^8 \text{ cm/s}$.

(i) Part (a)

$$\begin{aligned} \beta &= \omega \sqrt{\mu} = 2.00 \times 10^8 \text{ cm/s} \\ \omega_1 &= \frac{f}{\lambda} \sqrt{\mu} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_2 &= \frac{f}{\lambda} \sqrt{\frac{\mu}{\rho}} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_3 \omega_4 &= \frac{f}{\lambda} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_5 \omega_6 &= \frac{f}{\lambda} = 1.00 \times 10^8 \text{ cm/s} \end{aligned}$$

(ii) Part (b) — $C_1 \lambda_1 = \frac{f}{\rho} = 1.00 \times 10^8 \text{ cm} \cdot \text{s}$

$$C_1 = \sqrt{1 - \frac{C_2^2}{C_1^2}} = 0.9999$$

$$\begin{aligned} \beta &= \omega \sqrt{\mu} = 2.00 \times 10^8 \text{ cm/s} \\ \omega_1 &= \frac{f}{\lambda} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_2 &= \frac{f}{\lambda} = \frac{f}{\lambda} \sqrt{\frac{\mu}{\rho}} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_3 &= \omega_4 = 1.00 \times 10^8 \text{ cm/s} \\ \omega_5 &= \omega_6 = 1.00 \times 10^8 \text{ cm/s} \\ \omega_7 &= \omega_8 = 1.00 \times 10^8 \text{ cm/s} \end{aligned}$$

(iii) Part (c) — $C_2 \lambda_2 = \frac{f}{\rho} = 1.00 \times 10^8 \text{ cm} \cdot \text{s}$

$$C_2 = \sqrt{1 - \frac{C_1^2}{C_2^2}} = 0.9999$$

$$\begin{aligned} \beta &= \omega \sqrt{\mu} = 2.00 \times 10^8 \text{ cm/s} \\ \omega_1 &= \frac{f}{\lambda} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_2 &= \frac{f}{\lambda} \sqrt{\frac{\mu}{\rho}} = 1.00 \times 10^8 \text{ cm/s} \\ \omega_3 &= \omega_4 = 1.00 \times 10^8 \text{ cm/s} \\ \omega_5 &= \omega_6 = 1.00 \times 10^8 \text{ cm/s} \\ \omega_7 &= \omega_8 = 1.00 \times 10^8 \text{ cm/s} \end{aligned}$$

Ex. 11 (i) **Part (a)** — $C_1 \lambda_1 = C_2 \lambda_2 = 1.00 \times 10^8 \text{ cm} \cdot \text{s}$

All required quantities are the same as above for the Part (a) in problem 10. If $\rho = 10^3 \text{ kg/m}^3$, using ω_1 using f/λ , (100%), we have

$$\omega_1 = \frac{f}{\lambda} \sqrt{\frac{\mu}{\rho}} = \frac{f}{\lambda} = 1.00 \times 10^8 \text{ cm/s}$$

6) TE₁₀ mode ——— $(\lambda_c)_{TE_{10}} = (\lambda_c)_{TE_{10}} = 2a \sin^2 \theta \cos \theta < 2a$.

All required quantities are the same as those for the TE₁₀ mode in problem 5 but λ_c except α_1 .

$$\alpha_1 = \frac{1}{2a} \sqrt{\frac{2a^2}{\lambda_c} - \left(\frac{a}{\lambda_c}\right)^2} = \frac{1}{2a} \sin^2 \theta \quad (\text{Eqn. 6})$$

Ex. 10 For TE₁₀ mode in a parallel-plate waveguide,

$$\alpha_1 = \frac{1}{2a} \sqrt{\frac{2a^2}{\lambda_c} - \frac{a^2}{\lambda_c^2}} \\ = \frac{1}{2a} \sqrt{\frac{2a^2}{\lambda_c} - \frac{1}{\lambda_c^2}}$$

where $f(\lambda_c) = a - a^2$, $a = \lambda_c/2$.

6) To find minimum α_1 , set

$$\frac{df(\lambda_c)}{d\lambda_c} = 1 - 2a = 0 \quad \text{---} \quad a = \frac{\lambda_c}{2}$$

$$\therefore f = \sqrt{2} \lambda_c$$

8) At $\lambda_c/2 = 0.5 \lambda_c$, $\frac{1}{\lambda_c^2} = \frac{1}{(0.5 \lambda_c)^2}$

$$\text{and } \alpha_1 = \frac{1}{2a} \sqrt{\frac{2a^2}{\lambda_c} - \frac{1}{\lambda_c^2}}$$

9) For $\alpha_1 = 0.001 \text{ rad}$, $b = 0.001 \lambda_c$, $\mu_r = 1$, $\epsilon_r = 1$, and $f = 100 \text{ GHz}$,

$$(\lambda_c)_{TE_{10}} = \frac{1}{\alpha_1} \sin^2 \theta = 0.001 \lambda_c$$

$$\text{min. } \alpha_1 = 0.001 \text{ rad} \quad (\text{Eqn. 7})$$

Ex. 11 Parallel-plate waveguide: incident for $\mu_r = 1$, $\epsilon_r = 1$.

6) TE₁₀ mode

From Eqs. (1) and (2):

$$\begin{cases} H_1^2 = E_1^2 \\ H_2^2 = \frac{E_2^2}{\mu_r} \end{cases}$$

$$P_{\text{av}} = \frac{1}{2} \int_0^a (H_1^2 + H_2^2) dy = \frac{1}{2} \frac{E_1^2}{\mu_r} a$$

Substituting strength of air: $\text{Max. } E_1 = 3 \text{ kV/cm}$ (100V)

$$\text{Max. } \left(\frac{P_{\text{av}}}{a}\right) = \frac{1}{2} \frac{E_1^2}{\mu_r} = \frac{1}{2} (3000)^2 = 4.5 \text{ MW/cm} \quad (\text{max.}) = 4.5 \text{ MW/cm} \quad (\text{Eqn. 8})$$

ii) TM₁₀ mode

From Eq. (20-10) and (20-11)

$$\begin{cases} E_z^0(x,y) = E_0 \cos\left(\frac{\pi x}{a}\right) \\ H_z^0(x,y) = -\frac{E_0}{\eta_0 \sqrt{1-\beta_{TM}^2}} \sin\left(\frac{\pi x}{a}\right) \end{cases}$$

$$k_z = \frac{\beta_{TM}}{\eta_0} = \beta_0 \cos\theta \quad \text{and}$$

$$E_{\text{av}} = \frac{1}{2} \int_V E_z^0 \cdot H_z^0 \cos\theta \, dV = \frac{E_0^2 \beta_{TM}}{2 \eta_0 \sqrt{1-\beta_{TM}^2}}$$

$$\text{Max. } \left(\frac{E_{\text{av}}}{E_0}\right) = \frac{\beta_{TM}^2 \cos\theta}{2 \eta_0 \sqrt{1-\beta_{TM}^2}} = 1 \quad \text{at } \theta = 0 \text{ (total internal reflection)}$$

iii) TE₁₀ mode

From Eq. (20-14) and (20-15)

$$\begin{cases} E_z^0(x,y) = E_0 \sin\left(\frac{\pi x}{a}\right) \\ H_z^0(x,y) = \frac{E_0}{\eta_0 \sqrt{1-\beta_{TE}^2}} \cos\left(\frac{\pi x}{a}\right) \end{cases}$$

$$E_{\text{av}} = \frac{1}{2} \int_V E_z^0 \cdot H_z^0 \sin\theta \, dV = \frac{E_0^2 \beta_{TE}}{2 \eta_0 \sqrt{1-\beta_{TE}^2}}$$

$$\text{Max. } \left(\frac{E_{\text{av}}}{E_0}\right) = \frac{\beta_{TE}^2 \sin\theta}{2 \eta_0 \sqrt{1-\beta_{TE}^2}} = 1 \quad \text{at } \theta = 90^\circ \text{ (total internal reflection)}$$

Ex. 20-12 a) TM₁₀ mode



b) TE₁₀ mode



— Electric Field Lines
- - - Magnetic Field Lines

Ex. 20-13 If you derive (20-124) through (20-127) for TM₁₀ mode

$$E_z(x,y) = \frac{2E_0}{\pi} \left(\frac{\pi}{a}\right) E_0 \cos\left(\frac{\pi x}{a}\right) \sin\left(\frac{\pi y}{b}\right)$$

$$E_y^0(x,y) = \frac{2E_0}{\pi} \left(\frac{\pi}{a}\right) E_0 \sin\left(\frac{\pi x}{a}\right) \cos\left(\frac{\pi y}{b}\right)$$

$$I_y^c(x,y) = I_y(x,y) + A \left(\frac{b^2}{12} + y^2 \right)$$

$$I_y^c(x,y) = \frac{b^3}{12} \left(\frac{b}{3} \right) + A \left(\frac{b^2}{12} + y^2 \right) = \frac{b^3}{36}$$

$$I_y^c(x,y) = \frac{b^3}{36} \left(\frac{b}{3} \right) + A \left(\frac{b^2}{12} + y^2 \right) = \frac{b^3}{36}$$

a) *Parallel axis theorem:*

$$\begin{aligned} I_y^c(x,y) &= I_y + A d^2 = I_y + A \left(\frac{b}{2} \right)^2 = I_y + A \left(\frac{b^2}{4} \right) = I_y + A \left(\frac{b^2}{12} + \frac{b^2}{6} \right) \\ &= I_y + A \left(\frac{b^2}{12} + \frac{b^2}{6} \right) = I_y + A \left(\frac{b^2}{12} + \frac{2b^2}{12} \right) = I_y + A \left(\frac{3b^2}{12} \right) \\ &= I_y + A \left(\frac{b^2}{4} \right) \end{aligned}$$

$$\begin{aligned} I_y^c(x,y) &= I_y + A d^2 = I_y + A \left(\frac{b}{2} \right)^2 = I_y + A \left(\frac{b^2}{4} \right) = I_y + A \left(\frac{b^2}{12} + \frac{b^2}{6} \right) \\ &= I_y + A \left(\frac{b^2}{12} + \frac{b^2}{6} \right) = I_y + A \left(\frac{b^2}{12} + \frac{2b^2}{12} \right) = I_y + A \left(\frac{3b^2}{12} \right) \\ &= I_y + A \left(\frac{b^2}{4} \right) \end{aligned}$$



Ex-14 Rectangular area: $a = 4$ cm, $b = 3$ cm.

$$I_p = \frac{ab^3}{12} + \frac{a^3b}{12} = \frac{4 \cdot 3^3}{12} + \frac{4^3 \cdot 3}{12}$$

Moments with the centroid I_y, I_x are:

Moments	I_y	I_x	I_{xy}	I_p
At Cent	14.4	27	4.8	41.9

a) For $A = 12$ cm², the only propagating value is I_y .

b) For $A = 12$ cm², the propagating values are:

$$I_y, I_x, I_{xy}, I_{yz}, \text{ and } I_{xz}$$

Solution: $\alpha_{200} = \alpha_0 \sqrt{1 - \frac{1}{200}}$

For the TC_{200} mode, $\lambda_0 = \frac{1}{200}$

$\therefore \alpha_{200} = \alpha_0 \sqrt{1 - \frac{1}{200}} = \frac{1}{\sqrt{200}} \sqrt{1 - \frac{1}{200}}$

Q.10.11 $Q(L) = \frac{1}{\sqrt{2\pi\sigma^2}} \sqrt{\frac{2\pi\sigma^2}{L}} e^{-\frac{1}{2\sigma^2} L} = \frac{1}{\sqrt{2\pi\sigma^2}} f(L, \sigma)$

a) $\sigma = 1$, $f(L, \sigma) = \sqrt{\frac{2\pi}{L}} e^{-\frac{1}{2}L}$

Mode	Area
TC_{10}	1
TC_{10}, TC_{20}	1
$TC_{10}, TC_{20}, TC_{30}$	1/2
$TC_{10}, TC_{20}, TC_{30}, TC_{40}$	1/4
$TC_{10}, TC_{20}, TC_{30}, TC_{40}, TC_{50}$	1/8
$TC_{10}, TC_{20}, TC_{30}, TC_{40}, TC_{50}, TC_{60}$	1/16

b) $\sigma = 2$, $f(L, \sigma) = \sqrt{\frac{2\pi}{L}} e^{-\frac{1}{4}L}$

Mode	Area
TC_{10}, TC_{20}	1
$TC_{10}, TC_{20}, TC_{30}, TC_{40}, TC_{50}, TC_{60}$	1
$TC_{10}, TC_{20}, TC_{30}, TC_{40}, TC_{50}, TC_{60}, TC_{70}, TC_{80}$	1/2
$TC_{10}, TC_{20}, TC_{30}, TC_{40}, TC_{50}, TC_{60}, TC_{70}, TC_{80}, TC_{90}, TC_{100}$	1/4

Q.10.12 $f = 10 \times 10^6 \text{ (bit/s)}$, $\lambda = 10^6 \text{ (1/s)}$

Let $\sigma = 10$, i.e. $Q(L) = \frac{1}{\sqrt{2\pi\sigma^2}} \sqrt{\frac{2\pi\sigma^2}{L}} e^{-\frac{1}{2\sigma^2} L}$

a) $Q(L) = \frac{1}{\sqrt{2\pi\sigma^2}} \sqrt{\frac{2\pi\sigma^2}{L}} e^{-\frac{1}{2\sigma^2} L}$ for the desired TC_{20} mode.

For $f > 10$ ($Q(L) = 1$) assumed.

The next higher mode is TC_{30} with $Q(L) = \frac{1}{\sqrt{2\pi\sigma^2}}$

For $f < 10$ ($Q(L) = 1$) assumed.

We choose $a = 10 \text{ (bit)}$ and $b = 10 \text{ (bit)}$.

b) $\alpha_0 = \frac{1}{\sqrt{2\pi\sigma^2}} \sqrt{\frac{2\pi\sigma^2}{L}} = 10 \times 10^6 \text{ (bit)}$

$\lambda_0 = \frac{1}{\sqrt{2\pi\sigma^2}} = 10 \times 10^6 \times 10^6 \text{ (bit)}$

$\beta = \frac{1}{\sqrt{2\pi}} \times 10 \text{ (bit)}$

$(TC_{20})_0 = \frac{1}{\sqrt{2\pi\sigma^2}} = 10 \text{ (bit)}$

Ex 10.19 Given: $a = 1.2 \times 10^2$ (cm), $b = 2.5 \times 10^2$ (cm), $p = 0.5 \times 10^2$ (cm).

a) $A = \frac{1}{2} ab = \frac{1}{2} (1.2 \times 10^2)(2.5 \times 10^2)$

$A = \sqrt{1.5 \times 10^4} = 122.47$

$A_1 = A \cos \theta = 122.47 \cos 30^\circ = 106.07$ (cm²)

$A_2 = A \sin \theta = 122.47 \sin 30^\circ = 61.235$

$A_3 = A \cos^2 \theta = 1.2 \times 10^2 \times 0.75$

$A_4 = A \sin^2 \theta = 0.5 \times 10^2 \times 0.75$

$(A_1 A_2 A_3 A_4) = 1.2 \times 10^2 \times 0.5 \times 10^2 = 3 \times 10^4$ (cm⁴)

b) $A = \frac{1}{2} ab = \frac{1}{2} (1.2 \times 10^2)(2.5 \times 10^2)$

$A = \sqrt{1.5 \times 10^4} = 122.47$

$A_1 = A \cos \theta = 122.47 \cos 30^\circ = 106.07$ (cm²)

$A_2 = A \sin \theta = 122.47 \sin 30^\circ = 61.235$

$A_3 = A \cos^2 \theta = 1.2 \times 10^2 \times 0.75$

$A_4 = A \sin^2 \theta = 0.5 \times 10^2 \times 0.75$

$(A_1 A_2 A_3 A_4) = \frac{1}{2} ab = 1.5 \times 10^4$ (cm⁴)

Ex 10.20 Given: $a = 3 \times 10^2$ (cm), $b = 4 \times 10^2$ (cm), $p = 2 \times 10^2$ (cm)

a) $A_1 = 3 \times 10^2 \times 4 \times 10^2$ (cm²)

$A_2 = \frac{1}{2} ab = 6 \times 10^4$ (cm²)

b) $A = \frac{1}{2} ab = 6 \times 10^4$ (cm²) $\sqrt{1 - \left(\frac{2}{3}\right)^2} = 0.866$

$A_1 = \frac{1}{2} ab \cos^2 \theta = 4.09 \times 10^4$ (cm²)

c) $A_2 = \frac{1}{2} ab \sin^2 \theta = 1.91 \times 10^4$ (cm²) $\frac{1}{2} \left(\frac{1.91 \times 10^4}{4.09 \times 10^4} \right) = 0.231 \times 10^4$ (cm²)

d) $A_3 = \frac{1}{2} ab = 6 \times 10^4$ (cm²) $\frac{1}{2} \left(\frac{6 \times 10^4}{4.09 \times 10^4} \right) = 0.731 \times 10^4$ (cm²)

Ex 10.21 Given: $a = 3 \times 10^2$ (cm), $b = 4 \times 10^2$ (cm), $p = 10^2$ (cm)

a) $A = \frac{1}{2} ab = 6 \times 10^4$ (cm²) $A_1 = 3 \times 10^2 \times 10^2 = 3 \times 10^4$ (cm²)

$\sqrt{1 - \left(\frac{1}{3}\right)^2} = \frac{2}{3}$ $\sqrt{1 - \left(\frac{1}{4}\right)^2} = \frac{3}{4}$

$A_2 = 3 \times 10^2 \times 10^2 \left(\frac{2}{3} \right)^2 = \frac{4}{9} \left(\frac{3 \times 10^4}{4} \right) \left[1 - \frac{1}{9} \left(\frac{16}{9} \right) \right]$

$= 1.29 \times 10^4$ (cm²)

b) From Eqs. (20-43), (20-44) and (20-45):

$$E_0 = E_0 \sin\left(\frac{\pi y}{b}\right)$$

$$H_0 = -\hat{z} \sqrt{\frac{\mu_0}{\epsilon_0}} \cos\left(\frac{\pi y}{b}\right)$$

$$E_0 = \hat{z} \left(\frac{b}{2}\right) \frac{dH_0}{dy} = \hat{z} \cos\left(\frac{\pi y}{b}\right)$$

$$E_0 = \hat{z} \left(\frac{b}{2}\right)^2 \left(-\frac{\pi}{b}\right)^2 H_0 \sin\left(\frac{\pi y}{b}\right) = \frac{\pi^2 b^2}{4} \sqrt{\frac{\mu_0}{\epsilon_0}} \sin\left(\frac{\pi y}{b}\right)$$

For $E_0 = 0$ at the end surfaces, assuming average matched conditions:

$$\sin\left(\frac{\pi y}{b}\right) = E_0 = 0, \text{ for } y=0, \quad \sin\left(\frac{\pi y}{b}\right) = E_0 = 0, \text{ for } y=b$$

The two possible solutions — The field distribution is higher at the center and by a factor of $2^{1/2} = 0.707$

$$\therefore \text{Max. } |E_0| = 0.707 \text{ (vol.)}$$

$$\text{Max. } |H_0| = 2.23 \text{ (amp.)}$$

$$\text{Max. } |E_0| = 199.2 \text{ (vol.)}$$

$$\text{a) } P_{\text{inc}} = E_0 \cdot H_0 = (E_0 H_0) \hat{z} = -E_0 H_0 \hat{z} = -E_0 \left(\frac{E_0}{\sqrt{\mu_0/\epsilon_0}}\right) \hat{z}$$

$$|P_{\text{inc}}| = 199.2 \times 2.23 \text{ (vol.)}$$

$$P_{\text{ref}} = E_0 \cdot H_0 = (E_0 H_0) \hat{z} = -E_0 H_0 \hat{z} = -E_0 \left(\frac{E_0}{\sqrt{\mu_0/\epsilon_0}}\right) \hat{z}$$

$$|P_{\text{ref}}| = (199.2)^2 + (2.23)^2 = \frac{1}{2} \left[\left(\frac{E_0}{\sqrt{\mu_0/\epsilon_0}}\right)^2 + \left(\frac{E_0}{\sqrt{\mu_0/\epsilon_0}}\right)^2 \right]$$

total dissipation of energy

$$\text{At the } \hat{z} \text{ boundary and } |P_{\text{ref}}| = \frac{1}{2} \left(\frac{E_0}{\sqrt{\mu_0/\epsilon_0}}\right)^2 = 199.2 \times 2.23 \text{ (vol.)}$$

d) Total amount of average power dissipated in load of waveguide:

$$E_0 = 199.2 (1^{1/2} - 0) = 199.2 (1^{1/2} - 0) = 223.2 \text{ (vol.)}$$

Solve From problem 2 (c) (ii), we have

$$E_{00} = \frac{2.23 \times 199.2}{2} \sqrt{1 - \left(\frac{f}{f_c}\right)^2} = \frac{2.23 \times 199.2}{2} = 223.2$$

$$\therefore \text{Max. } P_{\text{av}} = \frac{(223.2)^2 - (2.23 \times 199.2)^2}{2 \times 223.2} = 223.2 \text{ (vol.)} = 2 \text{ (amp.)}$$

Ex. 11.11 a)

TM₁



b) TE₁₁



———— Electric Field Lines

..... Magnetic Field Lines

$$\text{Ex. 11.11 c) } E_p \text{ (TM)}: E_z = \frac{h}{\gamma_{z0}^2} = \frac{j\omega\mu^2}{\gamma_{z0}^2}$$

$$\text{For TM}_{10} \text{ mode, } \gamma_{z0} = \frac{2.405}{a} \rightarrow \omega_{c10} = \frac{2.405}{a} c$$

$$\text{For TE}_{10} \text{ mode, } \gamma_{z0} = \frac{2.405}{a} \rightarrow \omega_{c10} = \omega_{c10} \text{ (Degenerate mode)}$$

Ex. 11.12 $\beta^2 = \beta^2 - \beta^2 = \omega^2 \mu \epsilon - \beta^2$

$$\text{For TE}_{10} \text{ mode, } \beta = \frac{\omega^2 \mu \epsilon - (2.405/a)^2}{\omega^2 \mu \epsilon} \rightarrow \omega_{c10} = 2.405/a$$

$$\text{For TM}_{10} \text{ mode, } \beta = \frac{\omega^2 \mu \epsilon - (2.405/a)^2}{\omega^2 \mu \epsilon} \rightarrow \omega_{c10} = 2.405/a$$



a) If a is doubled, ω_c and ω_{c10} in diagram (a) are halved but diagram (b) will remain the same.

b) If the dielectric medium is changed then ω_c and ω_{c10} (for $\omega_c > \omega_{c10}$) both ω_c and ω_{c10} are reduced by a factor of $\sqrt{\epsilon}$ and the slope of the asymptotic line is changed from $\sqrt{\mu_0 \epsilon_0}$ to $\sqrt{\mu_0 \epsilon}$, diagram (b) remains unchanged.

Ex 21 of Problems: $L_2^2 = C_2 L_2(\text{Re}) \neq 0$.

Boundary conditions $L_2^2 = 0$ at both ends and $\mu \neq 0$ are satisfied when n is an integer.

There are no TM_{0n} modes.

(i) TE modes: $H_z^2 = C_2 L_2(\text{Re}) \cos \mu z$, where $\mu z = \frac{n\pi z}{a}$, n an integer.

TM_{0n} modes do exist.

(ii) For TE modes, L_2^2 at $z=0$ requires that $L_2(\text{Re})=0$
 \rightarrow Eigenvalues $(TM_{0n}) = \mu_{0n}^2/a^2$, $n=1, 2, \dots$

For TM modes, L_2^2 at $z=0$ requires that $L_2'(\text{Re})=0$.

\rightarrow Eigenvalues $(TM_{0n}) = \mu_{0n}^2/a^2$, $n=1, 2, \dots$

Ex 22 From Eqs (10-101) and (10-102)

Inside the slab: $\beta^2 = \mu_1^2 \mu_2^2 + k_0^2 = \mu_1^2 \mu_2^2$.

Outside the slab: $\beta^2 = \mu_1^2 \mu_2^2 + k^2 = \mu_1^2 \mu_2^2$.

$$\therefore \mu_1 \mu_2 \alpha_1 + \beta = \mu_1 \mu_2 \alpha_2,$$

$$\text{and } \frac{\beta}{\mu_1 \mu_2} = \mu_1^2 \frac{\beta}{\mu_1^2} = \frac{\beta}{\mu_1^2}.$$

Ex 23 From Eqs (10-125) and (10-126)

$$\left(\frac{\beta}{\mu_1}\right)^2 = \left(\frac{\beta}{\mu_2}\right)^2 = \left(\frac{\beta}{\mu_1}\right)^2 \left(\frac{\mu_1^2}{\mu_2^2}\right) \quad \text{--- (i)}$$

$$\frac{\beta}{\mu_1} = \left(\frac{\beta}{\mu_2}\right) \mu_1 \left(\frac{\mu_1}{\mu_2}\right) \quad \text{--- (ii)}$$

Let $X = \mu_1 \alpha_1$, $Y = \mu_2 \alpha_2$, $\alpha = \mu_1 \mu_2$, and $\beta = \frac{\mu_1^2 \mu_2^2}{\mu_1 \mu_2} = \alpha$.

$$\text{Eqs. (i) and (ii) become } \begin{cases} X^2 + Y^2 = \alpha^2 & \text{--- (i)} \\ Y = \alpha X \text{ and } & \text{--- (ii)} \end{cases}$$

(i) $\beta = \alpha$ and $\alpha = \mu_1 \mu_2 = \alpha$ and

$$\mu_1 \alpha_1 = \mu_2 \alpha_2 = \alpha \text{ and } \alpha = \mu_1 \mu_2 = \alpha \text{ and } \alpha = \alpha \text{ and } \alpha = \alpha \text{ and } \alpha = \alpha.$$



Graphical solution:

$$x_2^2 = 0.0005, \quad y_2^2 = 0.000100^2$$

$$x = 0.0224, \quad y = 0.0100 \quad (\text{approx.})$$

$$y_2^2 = 0.000100^2 = 0.00001000$$

$$\text{From Eq. (2-102): } \beta = \beta_0 \sqrt{1 - \frac{y_2^2}{\beta_0^2}} = 0.9999 \quad (\text{approx.})$$

4) $f = 0.000100^2$ (rad), $\lambda = 0.000100$ (rad), $\lambda_0 = 0.000100$

$$x = 0.0224, \quad y = 0.0100$$

$$x_2 = 0.0005, \quad y_2 = 0.000100^2$$

We obtain $x = 0.0224$ (approx.)

$$\beta = 0.9999 \quad (\text{approx.})$$

Ex. 11. From Eq. (2-101):

$$\left(\frac{dx}{dy}\right)^2 = -\frac{1}{2} \left(\frac{dy}{dx}\right)^2 \cos^2 \left(\frac{dy}{dx}\right) \quad (2)$$

Using the notation in problem 2-10(1), we obtain two equations from (2) in 2-10(1) and (2) above:

$$\begin{cases} X^2 + Y^2 = R^2, & (1) \\ Y^2 = 2R^2 \cos^2 X, & (2) \end{cases}$$

4) $f = 0.000100^2$ (rad), $\lambda = 0.000100$ (rad), $\lambda_0 = 0.000100$

$$x = 0.0224,$$

$$y = 0.0100,$$

$$\lambda = 0.0224,$$

$$\lambda_0 = 0.0100.$$

There are no intersections for curves representing Eqs. (1) and (2); hence, even the nodes do not exist at the given frequencies.

Ex. 12. Use Eqs. (2-101) and (2-102):

$$x_2^2 = -\frac{dy}{dx} \frac{dy}{dx}, \quad x_2^2 = \frac{dy}{dx} \frac{dy}{dx}$$

$$f(x, y_2) = \beta_0 [f^2(x) e^{2i\pi y_2 / \lambda_0}]$$

$$f(x, y_2) = \beta_0 [2f^2(x) e^{2i\pi y_2 / \lambda_0}]$$

Table 1:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) &= \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \\
\mathcal{L}_1^4(\alpha) &= \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \longrightarrow \mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha) \otimes \mathcal{L}_1(\alpha)
\end{aligned}$$

Table 2:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1^2(\alpha) &= \mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^2(\alpha) \longrightarrow \mathcal{L}_1^6(\alpha) \\
\mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1^2(\alpha) &= \mathcal{L}_1^5(\alpha) \otimes \mathcal{L}_1^3(\alpha) \longrightarrow \mathcal{L}_1^8(\alpha) \\
\mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^2(\alpha) &= \mathcal{L}_1^6(\alpha) \otimes \mathcal{L}_1^4(\alpha) \longrightarrow \mathcal{L}_1^{10}(\alpha)
\end{aligned}$$

Table 3:

$$\begin{aligned}
\mathcal{L}_1^2(\alpha) \otimes \mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^5(\alpha) \otimes \mathcal{L}_1^3(\alpha) \longrightarrow \mathcal{L}_1^8(\alpha) \\
\mathcal{L}_1^3(\alpha) \otimes \mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^6(\alpha) \otimes \mathcal{L}_1^3(\alpha) \longrightarrow \mathcal{L}_1^9(\alpha) \\
\mathcal{L}_1^4(\alpha) \otimes \mathcal{L}_1^3(\alpha) &= \mathcal{L}_1^7(\alpha) \otimes \mathcal{L}_1^3(\alpha) \longrightarrow \mathcal{L}_1^{10}(\alpha)
\end{aligned}$$

Table 4) From Table 1 and 2, it is seen that $\mathcal{L}_1^6(\alpha)$ for \mathcal{L}_1 mode, which is the dominant mode.

From Eq. (10-11):

$$\alpha = \frac{\beta^2}{\beta_0} \beta_0 \text{ for } \beta^2 = \frac{\beta_0^2}{\beta_0} \beta_0, \text{ for } \beta^2 > \beta_0$$

Neglecting the α^2 term in Eq. (10-11):

$$\beta^2 - \alpha \beta_0 \beta_0 = \alpha^2 \text{ or } \alpha = \beta_0 \sqrt{\beta^2 - \beta_0^2}$$

From Eq. (10-11): $\beta_0^2 \alpha \beta_0 \beta_0 = \beta^2 - \beta_0^2$

$$\therefore \alpha = \frac{\beta^2 - \beta_0^2}{\beta_0^2} = \beta_0^2$$

2) at $\beta = \beta_0$, $\beta_0 = \beta_0$, $\beta_0 = \beta_0$, $\beta = \beta_0$ and $\beta_0 = \beta_0$, $\beta_0 = \beta_0$:

$$\alpha = \frac{\beta^2}{\beta_0} \beta_0 = 1 \text{ or } \alpha = 1 \text{ (Eq. 10)}$$

$$\beta^2 - \beta_0^2 = \alpha \beta_0^2, \quad \alpha \beta_0 - \beta_0^2 = 1$$

$$\longrightarrow \left(\beta - \frac{\beta_0}{2} \right) = 1.041 \text{ cm}$$

Table 1 See Eqs. (19-21) and (22-23)

$$\begin{aligned} \epsilon_1^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \omega} \\ \epsilon_{1, \text{odd}}^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \omega} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \omega} \\ \epsilon_{1, \text{even}}^* &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \omega} & \epsilon^* &= \frac{\partial \epsilon}{\partial \omega} \frac{\partial \omega}{\partial \omega} \end{aligned}$$

(19-21)

$$\begin{aligned} \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \end{aligned}$$

(22-23)

$$\begin{aligned} \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \end{aligned}$$

(24-25)

$$\begin{aligned} \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \\ \epsilon_1^* \omega &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega & \epsilon^* \omega &= \frac{\partial \epsilon}{\partial \omega} \omega \end{aligned}$$

Setting $\omega = \omega_0$ in $\epsilon_1^*(\omega) = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega$ and
in $\epsilon^*(\omega) = \frac{\partial \epsilon}{\partial \omega} \omega$ we obtain

and expanding, we obtain

$$\begin{aligned} \epsilon_1^*(\omega_0) &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega_0 \\ &= \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega_0 \end{aligned}$$

Table 2 a) Table 2(a) and Table 2(b) are the propagator loop modes. Using the ϵ_1^* and ϵ^* in the formulas in Table 1(a), (b), (c), we have

$$\epsilon_1^* = \frac{1}{2} \frac{\partial \epsilon}{\partial \omega} \omega_0 \quad \text{for odd TE modes,}$$

$$\epsilon^* = \frac{\partial \epsilon}{\partial \omega} \omega_0 \quad \text{for even TE modes,}$$

4) Dielectric Dielectric — From Eq. (2) and (3):

$$\text{for } r < a: \quad \epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

Find charge density in conductor $\rho = \epsilon_0 \cdot E \Big|_{r=a}$

$$\rho = -\epsilon_0 \left. \frac{d\psi}{dr} \right|_{r=a} = -\epsilon_0 \frac{d}{dr} \left(\frac{C}{r} \right) \Big|_{r=a}$$

Find charge density in conductor $\rho = \epsilon_0 \cdot E \Big|_{r=a}$

$$\rho = \epsilon_0 \left. \frac{d\psi}{dr} \right|_{r=a} = \frac{2C}{a^2} \epsilon_0$$

Dielectric Dielectric — From problem 4 (a-c):

$$\text{for } r < a: \quad \epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\therefore \rho = \epsilon_0 \left(\frac{d}{dr} \left(\frac{C}{r} \right) \right) \Big|_{r=a} = \frac{2C}{a^2} \epsilon_0$$

$$\rho = \epsilon_0 \cdot \frac{d}{dr} \left(\frac{C}{r} \right) \Big|_{r=a}$$

4.11.11 Dielectric Dielectric — From problem 4 (a-c):

for $r < a$: $\epsilon_1 \nabla^2 \psi = -\frac{\rho}{\epsilon_0}$ and $\psi = 0$ at $r = a$

$$\left[\epsilon_1 \frac{d}{dr} \left(\frac{C}{r} \right) \right]_{r=a} = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \quad (1)$$

$$\left[\epsilon_1 \frac{d}{dr} \left(\frac{C}{r} \right) \right]_{r=a} = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \quad (2)$$

$$\text{From (1) & (2): } \left[\epsilon_1 \frac{d}{dr} \left(\frac{C}{r} \right) \right]_{r=a} = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\left[\epsilon_1 \frac{d}{dr} \left(\frac{C}{r} \right) \right]_{r=a} = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0$$

$$\left[\epsilon_1 \frac{d}{dr} \left(\frac{C}{r} \right) \right]_{r=a} = -\frac{\rho}{\epsilon_0}$$

Similarly, for $r > a$:

$$\epsilon_2 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \quad \text{at } r = a$$

$$\epsilon_2 \nabla^2 \psi = -\frac{\rho}{\epsilon_0} \quad \text{and } \psi = 0 \quad \text{at } r = a$$

Boundary conditions: $\psi = 0$ at $r = a$ and $r = b$

$$\frac{\rho}{\epsilon_0} \rightarrow \text{Characteristic equation for cylindrical dielectric: } \frac{d^2 \psi}{dr^2} + \frac{1}{r} \frac{d\psi}{dr} = -\frac{\rho}{\epsilon_0}$$

Ex. 10.12 From Eq. (10-103): $k_{\text{app}} = \frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}}$

$$k_{\text{app}} = 0.0017 \text{ s}^{-1} \text{ from (1)}, \quad k_{\text{app}} = \frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}}$$

Let us tabulate k_{app} for various values of k_1/k_2 :

k_1/k_2	k_{app}	$(k_{\text{app}} - 0.0017)$
0.0001	0.0001	-0.0016 s^{-1}
0.001	0.0010	-0.0007 s^{-1}
0.01	0.0040	+0.0023 s^{-1}
0.1	0.0200	+0.0183 s^{-1}
1	0.1000	+0.0983 s^{-1}
10	0.3244	+0.3227 s^{-1}
100	0.9798	+0.9781 s^{-1}
1000	0.9999	+0.9982 s^{-1}
10000	0.9999	+0.9982 s^{-1}

Ex. 10.13 (a) From (10-104), the quadratic equation made in k_{app} reads:

$$k_{\text{app}}^2 - \frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}} + 0.0017 = 0 \quad (2)$$

(b) From Eq. (10-104):

$$\begin{aligned} k_{\text{app}} &= \frac{0.0017 \pm \sqrt{0.0017^2 - 0.0017 \left(\frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}} \right)}}{2 \left(\frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}} \right)} \\ &= \frac{0.0017 \pm 0.0017 \left(\frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}} \right)}{\frac{1}{2} \sqrt{\frac{2k_1}{k_2} + \frac{2k_1^2}{k_2^2} - \frac{k_1}{k_2}}} = 0.0017 \end{aligned}$$

From Eqs. (10-103) and (10-104):

$$k_1 = \frac{1}{2} k_2 \left(k_{\text{app}}^2 + 0.0017 \right) = 0.001125 k_2 \quad (3)$$

$$k_2 = \frac{1}{2} k_1 \left(\frac{1}{k_{\text{app}}} + \frac{1}{0.0017} \right) = 0.00055 k_1 \quad (4) = 0.00061 k_2$$

Solve (i) $(V_{\text{ext}})_z = \frac{7}{2} \sqrt{\frac{1}{2} + \frac{1}{2}} = \frac{7}{2} (V_{\text{ext}})_z = 4.9 \text{ eV} \approx 0.180 \text{ eV}$

(ii) $(V_{\text{ext}})_z = \frac{7}{2} \sqrt{\frac{1}{2} - \frac{1}{2}} = 0$

(iii) $(V_{\text{ext}})_z - (V_{\text{ext}})_z = 4.9 \text{ eV} \approx 0.180 \text{ eV} = 4.9 \text{ eV} \approx 0.180 \text{ eV}$
 $= (V_{\text{ext}})_z$

Solve (a) Considering Eqs. (29-29) and (29-30)

$$V_{\text{ext}} = \frac{qV}{R^2} \frac{b \cos^2 \theta + d \sin^2 \theta}{(2b \cos^2 \theta + d \sin^2 \theta)^2}$$

→ V_{ext} has a symmetrical dependence on θ and θ' . It will be maximum when $\theta = \theta'$, which gives a max. value-to-surface ratio

(b) When $\theta = \theta'$, $V_{\text{ext}} = \frac{qV}{R^2} \frac{1}{2(1+b/a)}$

Solve (i) $V_{\text{ext}} = \frac{qV \cos^2 \theta + d \sin^2 \theta}{2a \cos^2 \theta + d \sin^2 \theta}$

For $\theta = 45^\circ$, $V_{\text{ext}} = \frac{qV}{2(a+d)} \sqrt{\frac{1}{2} + \frac{1}{2}} = 0.707 qV \left(\frac{1}{2} \right)$

$V_{\text{ext}} = 0.354 qV$

(ii) For $V_{\text{ext}} = 0.354 V_{\text{ext}}$, $\cos^2 \theta = 0.354$

Solve (i) From the field configuration in the cavity we see that the V_{ext} made with respect to a is the same as the V_{ext} made with respect to b . Thus, $(V_{\text{ext}})_{\text{ext}}$ can be obtained from $(V_{\text{ext}})_z$ in Eq. (29-30) by changing b to a and a to b .

(ii) V_{ext} for the V_{ext} can be derived from the field expression in Eqs. (29-29), (29-30), (29-31) by setting $\theta = \theta'$, and using Eq. (29-30).

$$V = 2V_0 = \frac{qV}{R^2} \left(\frac{2b^2}{2a^2} \right) \cos^2 \theta \quad \text{or } V_{\text{ext}}$$

$$V_{\text{ext}} = \frac{qV}{R^2} \left(\frac{b^2}{a^2} \right) \cos^2 \theta = \frac{qV}{R^2} \left(\frac{b^2}{a^2} \right) \cos^2 \theta$$

$$\begin{aligned}
 R_1 &= R_1 \left(\int_0^1 \frac{1}{x} dx \right)^2 = R_1 \ln^2 2 = R_1 \ln^2 2 \\
 &= R_1 \left(\int_0^1 \frac{1}{x} dx + \int_0^1 \frac{1}{x} dx \right) dx dy \\
 &= R_1 \left(\frac{1}{2} \ln^2 2 + \frac{1}{2} \ln^2 2 + \frac{1}{2} \ln^2 2 + \frac{1}{2} \ln^2 2 \right) \\
 &= R_1 \left(\frac{1}{2} \ln^2 2 + \frac{1}{2} \ln^2 2 \right)
 \end{aligned}$$

$$R_{\text{eff}} = \frac{R_1 \ln^2 2}{2} = \frac{R_1 \ln^2 2 (1 + \ln^2 2)}{2(1 + \ln^2 2)} = R_1 \frac{\ln^2 2}{2}$$

Example 11 TM_{long} mode:

$$R_1^2 = C_m R_1 \left(\frac{R_1}{2} \right) \ln^2 2 + R_1 \left(\frac{R_1}{2} \right)$$

$$m = 0, 1, 2, \dots; n = 0, 1, 2, \dots; p = 0, 1, 2, \dots$$

$$C_{\text{TM}_{\text{long}}} = \frac{R_1 \ln^2 2}{2(1 + \ln^2 2)}$$

$$C_{\text{TM}_{\text{long}}} = \frac{R_1 \ln^2 2}{2(1 + \ln^2 2)}$$

$$\text{TM}_{\text{long}} \text{ mode: } R_1^2 = C_m R_1 \left(\frac{R_1}{2} \right) \ln^2 2 + R_1 \left(\frac{R_1}{2} \right)$$

$$m = 0, 1, 2, \dots; n = 0, 1, 2, \dots; p = 0, 1, 2, \dots$$

$$C_{\text{TM}_{\text{long}}} = \frac{R_1 \ln^2 2}{2(1 + \ln^2 2)}$$

$$C_{\text{TM}_{\text{long}}} = \frac{R_1 \ln^2 2}{2(1 + \ln^2 2)}$$

1) For $\ln 2$, the dominant mode is TM_{long} / C_{TM_{long}} = $\frac{R_1 \ln^2 2}{2(1 + \ln^2 2)} = \frac{R_1 \ln^2 2}{2}$

The first seven modes with equivalent circuit for it

Mode	TM ₀₀	TM ₁₀	TM ₂₀	TM ₀₁	TM ₁₁	TM ₂₁	TM ₃₀
$\frac{R_1 \ln^2 2}{2}$	1.00	1.07	1.17	1.30	1.45	1.64	1.76

$$\text{Example 12} \quad C = \frac{R_1 \ln^2 2}{2} = \frac{R_1 \ln^2 2}{2}, \quad L = \frac{R_1 \ln^2 2}{2} \ln \left(\frac{2}{1} \right)$$

$$\text{a) } R_1 = \frac{R_1 \ln^2 2}{2} = \frac{R_1 \ln^2 2}{2} \ln \left(\frac{2}{1} \right)$$

$$\text{b) } R_1 = \frac{R_1 \ln^2 2}{2} = \frac{R_1 \ln^2 2}{2} \ln \left(\frac{2}{1} \right)$$

Chapter II

Antennas and Radiating Systems

Ex 1 Maxwell equations for dipole antenna:

$$\nabla \times \mathbf{E} = -\mu_0 \frac{\partial \mathbf{H}}{\partial t} \quad \text{--- (1)}$$

$$\nabla \times \mathbf{H} = \mathbf{J} + \frac{\partial \mathbf{D}}{\partial t} \quad \text{--- (2)}$$

$$\nabla \cdot \mathbf{E} = \frac{\rho}{\epsilon_0} \quad \text{--- (3)}$$

$$\nabla \cdot \mathbf{H} = 0 \quad \text{--- (4)}$$

$$\begin{aligned} \text{a) } \nabla \times \text{--- (2): } \nabla \times (\nabla \times \mathbf{E}) &= \mu_0 \frac{\partial}{\partial t} (\nabla \times \mathbf{H}) \\ &= \mu_0 \frac{\partial}{\partial t} \left(\mathbf{J} + \frac{\partial \mathbf{D}}{\partial t} \right) \end{aligned} \quad \text{--- (5)}$$

$$\begin{aligned} \text{But } \nabla \times (\nabla \times \mathbf{E}) &= \nabla(\nabla \cdot \mathbf{E}) - \nabla^2 \mathbf{E} \\ &= \frac{1}{\epsilon_0} \nabla \rho - \nabla^2 \mathbf{E} \end{aligned} \quad \text{--- (6)}$$

Combining (5) and (6), we obtain

$$\nabla^2 \mathbf{E} - \mu_0 \frac{\partial^2 \mathbf{D}}{\partial t^2} = \frac{1}{\epsilon_0} \nabla \rho - \mu_0 \frac{\partial \mathbf{J}}{\partial t}$$

$$\text{b) Similarly, we have } \nabla^2 \mathbf{H} - \mu_0 \frac{\partial^2 \mathbf{J}}{\partial t^2} = -\nabla \times \mathbf{J}$$

Ex 2 \mathbf{A}_D (vector) $\mathbf{A} = -\nabla \psi - \mu_0 \int \mathbf{J}_D \cdot \mathbf{e}_r / r^2 + \mu_0 \mathbf{e}_z \int \mathbf{J}_D \cdot \mathbf{e}_z / r^2$

$$\begin{aligned} \mathbf{e}_x &= -\frac{\partial \psi}{\partial x} - \mu_0 \mathbf{J}_x \\ \mathbf{e}_y &= -\frac{\partial \psi}{\partial y} - \mu_0 \mathbf{J}_y \\ \mathbf{e}_z &= -\frac{\partial \psi}{\partial z} - \mu_0 \mathbf{J}_z \end{aligned} \quad \begin{array}{l} \text{The expansion of } \mathbf{A}_D, \mathbf{A}_z \\ \text{and } \mathbf{A}_y \text{ are given in Ex.} \\ \text{--- (1), (2), (3) respectively.} \end{array}$$



$$\psi = \frac{1}{4\pi\epsilon_0} \left[\frac{q_{\text{total}}}{r} - \frac{q_{\text{total}}}{r} \right]$$

$$\mathbf{e}_x = \mathbf{e}_x - \mu_0 \mathbf{J}_x$$

$$\mathbf{e}_y = \mathbf{e}_y - \mu_0 \mathbf{J}_y$$

$$\mathbf{e}_z = \mathbf{e}_z - \mu_0 \mathbf{J}_z$$

$$\begin{aligned} \psi &= \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\mathbf{r}')}{r} dV' \\ &= \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\mathbf{r}')}{r} dV' - \frac{1}{4\pi\epsilon_0} \int \frac{\rho(\mathbf{r}')}{r} dV' \end{aligned}$$

$$V = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (1 + 2z) dz dy dx$$

$$= \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2z + 1) dz dy dx$$

Using A_1, A_2, A_3 and V in E_1, E_2 and E_3 , we obtain the same results as given in Eqs. (1)-(3).

Ex. 1



$$\text{Sol. } E_1 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2z + 1) dz dy dx$$

$$= \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_2 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$= \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_3 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_4 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_5 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_6 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_7 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_8 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_9 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$= \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

In the same manner, we have

$$E_{10} = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$\therefore E_{11} = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$E_{12} = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

Let $m = \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$

$$E_1 = \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$= \frac{1}{2} \int_{-\frac{a}{2}}^{\frac{a}{2}} \int_{-\frac{b}{2}}^{\frac{b}{2}} (2x + 2y + 1) dz dy dx$$

$$\begin{aligned} \text{a) } \vec{D} &= \frac{1}{\sqrt{\epsilon_0}} \vec{D}, \vec{E} = \vec{E}_p + \vec{E}_b = \vec{E}_p + \vec{E}_b \quad \text{Expression for } \vec{E}_p, \vec{E}_b \\ \text{b) } \vec{E} &= \frac{1}{\sqrt{\epsilon_0}} \vec{E} = \vec{E}_p + \vec{E}_b \quad \text{and } \vec{E}_b \text{ same as above} \\ & \quad \text{plus a } \vec{E}_b \text{ (Strahlungsfeld)} \end{aligned}$$

In the far zone, $r \gg \lambda$, $\vec{E}(\vec{r}, t)$ and $\vec{B}(\vec{r}, t)$ terms can be neglected. We have the following instantaneous expressions according to (10.10.10):

$$\begin{aligned} \vec{E}(\vec{r}, t) &= -\vec{E}_p \frac{d^2 \vec{p}(t - r/c)}{dt^2} \\ \vec{B}(\vec{r}, t) &= \vec{E}_p \frac{d^2 \vec{p}(t - r/c)}{dt^2} \times \hat{r} \\ \vec{H}(\vec{r}, t) &= \vec{E}_p \frac{d^2 \vec{p}(t - r/c)}{dt^2} \times \hat{r} \end{aligned}$$

10.11 Far-zone electric field of elemental electric dipole

$$\vec{E}_p(\vec{r}, t) = \frac{1}{4\pi\epsilon_0} \left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) \frac{1}{r^2} \hat{r} - \frac{1}{4\pi\epsilon_0} \left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) \frac{1}{r^3} \hat{r} \times \hat{r}$$

For the elemental magnetic dipole:

$$\vec{E}_p(\vec{r}, t) = \frac{1}{4\pi\epsilon_0} \left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) \frac{1}{r^2} \hat{r} - \frac{1}{4\pi\epsilon_0} \left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) \frac{1}{r^3} \hat{r} \times \hat{r}$$

$$\text{a) Thus, } \frac{\frac{d^2 \vec{p}(t - r/c)}{dt^2}}{\left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) / r^2} = \frac{\frac{d^2 \vec{p}(t - r/c)}{dt^2}}{\left(\frac{d^2 \vec{p}(t - r/c)}{dt^2} \right) / r^2} = 1$$

— — — Dipole polarization

b) Circular polarization if $\vec{p} = p \hat{e}_\theta$

$$\text{10.12} \quad \text{Equation of continuity: } \vec{\nabla} \cdot \vec{J} + \dot{\rho} = 0 \quad \rightarrow \quad \vec{J} = \frac{1}{4\pi} \frac{d^2 \vec{p}}{dt^2}$$

$$\text{a) } \vec{\nabla} \cdot \vec{J} = \vec{\nabla} \cdot \left(\frac{1}{4\pi} \frac{d^2 \vec{p}}{dt^2} \right) = \frac{1}{4\pi} \frac{d^2}{dt^2} \vec{\nabla} \cdot \vec{p} = \frac{1}{4\pi} \frac{d^2}{dt^2} \rho$$

$$\text{b) } \vec{\nabla} \cdot \vec{J} = \vec{\nabla} \cdot \left(\frac{1}{4\pi} \frac{d^2 \vec{p}}{dt^2} \right) = \frac{1}{4\pi} \frac{d^2}{dt^2} \vec{\nabla} \cdot \vec{p} = \frac{1}{4\pi} \frac{d^2}{dt^2} \rho$$

$$\text{10.13} \quad \lambda = \frac{2\pi c}{\omega} \text{ wavelength, } \frac{d^2}{dt^2} = \frac{d^2}{dt^2} = \frac{d^2}{dt^2} \text{ (harmonic dipole)}$$

$$\text{a) Radiation resistance, } R_r = 80\pi^2 \left(\frac{\lambda}{2\pi} \right)^2 = 80\pi^2 \left(\frac{\lambda}{2\pi} \right)^2$$

$$\text{b) } R_r(\lambda=10) = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 = 80\pi^2 \left(\frac{10}{2\pi} \right)^2$$

$$R_r(\lambda=10) = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 \rightarrow R_r = \frac{80\pi^2}{\lambda^2} = 80\pi^2 \lambda^{-2}$$

$$\text{c) } R_r(\lambda=10) = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 = 80\pi^2 \left(\frac{10}{2\pi} \right)^2 \rightarrow R_r(\lambda=10) = \frac{80\pi^2}{\lambda^2} = 80\pi^2 \lambda^{-2}$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \left[\sin^{-1} x \right]_0^1 = \frac{\pi}{2}$$

$$= \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx$$

$$= \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \frac{\pi}{2} = \frac{1}{\sqrt{1-x^2}} dx$$

$$dx = \frac{1}{\sqrt{1-x^2}} dx$$

In case of $\sin^{-1} x = \frac{1}{\sqrt{1-x^2}}$ and $\cos^{-1} x = \frac{-1}{\sqrt{1-x^2}}$

$$\therefore R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$, over continuity graph.
- From $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$ and $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$

$$R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$$

- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$, $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$
- $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$

From problem $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$ and $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$

- For $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$ and $R_2 = \int_0^1 \frac{1}{\sqrt{1-x^2}} dx = \frac{1}{\sqrt{1-x^2}} dx$

Ex. 10-20 a) $L_1 = \int_{-\pi/2}^{\pi/2} (\frac{d}{dt} \sin t)^2 dt = 2 \int_{-\pi/2}^{\pi/2} \cos^2 t dt$

b) $L_2 = \int_{-\pi/2}^{\pi/2} \cos^2 t dt$, $L_3 = \int_{-\pi/2}^{\pi/2} \sin^2 t dt$

c) For $\beta = \pi/2$ ($\sin \beta = 1$), $L_1 = L_2 = \int_{-\pi/2}^{\pi/2} \cos^2 t dt$ and $L_3 = \int_{-\pi/2}^{\pi/2} \sin^2 t dt$
 $\beta = 0$ ($\sin \beta = 0$), $L_1 = \int_{-\pi/2}^{\pi/2} \cos^2 t dt$ and $L_3 = \int_{-\pi/2}^{\pi/2} \sin^2 t dt$
 $L_2 = \int_{-\pi/2}^{\pi/2} \cos^2 t dt$, $L_3 = \int_{-\pi/2}^{\pi/2} \sin^2 t dt$
 $\rightarrow L_1 = 2 \int_{-\pi/2}^{\pi/2} \cos^2 t dt$

Ex. 10-21 $R = \int_{-\pi/2}^{\pi/2} L_1 \cdot dt = \int_{-\pi/2}^{\pi/2} \int_{-\pi/2}^{\pi/2} L_1 \cdot L_2 \cdot dt \cdot d\theta$

$$= \int_{-\pi/2}^{\pi/2} \int_{-\pi/2}^{\pi/2} \cos^2 t \cdot \sin^2 \theta \cdot \left(\left[\frac{d}{dt} \sin t \right] \left[\frac{d}{d\theta} \cos \theta \right] \right) dt \cdot d\theta$$

$$= \int_{-\pi/2}^{\pi/2} \left[\cos^2 t \right] dt \cdot \int_{-\pi/2}^{\pi/2} \sin^2 \theta \cdot \left[\frac{d}{d\theta} \cos \theta \right] d\theta$$

Ex. 10-22 $\text{proj} = \frac{\sin^2 \theta \cos \theta}{\sqrt{1 + \sin^2 \theta}}$



For $\theta = \pi/4$,

$$\text{proj} = \left| \frac{\sin^2(\pi/4) \cos(\pi/4)}{\sqrt{1 + \sin^2(\pi/4)}} \right|$$

Width of path from between the first walls
 $= 2 \sin^2(\pi/4) \cos(\pi/4)$

Ex. 10-23 $L(t) = L_1 \left(1 - \frac{t}{T} \right)$

$$\begin{aligned} \text{From Ex. 10-20: } L_1 &= \frac{d}{dt} \int_{-\pi/2}^{\pi/2} \cos^2 t \cdot dt \\ &= \frac{d}{dt} \int_{-\pi/2}^{\pi/2} \cos^2 t \cdot dt \\ &= \frac{d}{dt} \int_{-\pi/2}^{\pi/2} \left(1 - \sin^2 t \right) dt \\ &= \frac{d}{dt} \left[\int_{-\pi/2}^{\pi/2} (1 - \sin^2 t) dt \right] \end{aligned}$$

Maximum L_1 occurs at $t = \frac{\pi}{2}$, where
 $L_1 \left(\frac{\pi}{2} \right) = 2 - \frac{\pi}{2}$

Ex 11-10 a) $V_{\text{ext}} = -k_1 x_1 = -\frac{d}{2} \left[\frac{\cos(\frac{1}{2} \pi \cos^2 \theta)}{\sin \theta} \right]$

b) $E = \frac{1}{2} \left(\frac{d^2 V_{\text{ext}}}{dx_1^2} \right) x_1 = \frac{d^2 V_{\text{ext}}}{2 dx_1^2} = \frac{d^2}{2 dx_1^2} \left[\frac{\cos(\frac{1}{2} \pi \cos^2 \theta)}{\sin \theta} \right]$

which has a maximum value $(2 \times 10^3)^2 (10^{-10})^2$ at $\theta = \frac{\pi}{4}$

c) For $\lambda = \frac{h^2}{2mE} = 2.00 \text{ nm}$ and $E_0 = 10 \text{ eV}$ (eV) (eV)

$\theta = \frac{\pi}{4} \rightarrow x_1(\frac{\pi}{4}) = \frac{d}{2} = 1.0 \text{ nm}$, $V_{\text{ext}} = -1.0 \text{ eV}$

$E_1 = \frac{d^2}{2} \left(\frac{d^2 V_{\text{ext}}}{dx_1^2} \right) = 0.4 \text{ eV}$ (eV) = 1.7 eV (eV)

$\theta = \frac{3\pi}{4} \rightarrow x_1(\frac{3\pi}{4}) = \frac{d}{2} \left[\frac{\cos(\frac{1}{2} \pi \cos^2 \theta)}{\sin \theta} \right] = 0.5 \text{ nm}$,

$V_{\text{ext}} = -0.5 \text{ eV}$, $E_1 = 0.4 \text{ eV}$ (eV)

Ex 11-11



if $V(x) = 0$ (eV)

so $E_1 = \frac{h^2}{8ma^2} \psi^2 \text{ (eV)}$

(for $\psi = \sin \frac{\pi x}{a}$)

$= \frac{h^2}{8ma^2} \left(\frac{\pi^2}{a^2} \right) \psi^2 \text{ (eV)}$

where $\psi(x) = \sin \frac{\pi x}{a}$ (eV)

a) if $a = 0.5 \text{ nm}$

$[E_1] = \{0.10 \text{ eV (eV)}\}$



c) if $a = 1 \text{ nm}$

$[E_1] = \{0.025 \text{ eV (eV)}\}$



Ex-10



From Ex-9 we have
 $L_1 = \frac{y - y_1}{m_1} = \frac{y - y_2}{m_2}$ and $L_2 = \frac{y - y_1}{m_2} = \frac{y - y_2}{m_1}$

where

$$\begin{aligned} \sin \theta &= |m_1 - m_2| \\ &= \left| \frac{y_2 - y_1}{x_2 - x_1} - \frac{y_2 - y_1}{x_1 - x_2} \right| \\ &= \left| \frac{y_2 - y_1}{x_2 - x_1} + \frac{y_2 - y_1}{x_2 - x_1} \right| \\ &= \frac{2(y_2 - y_1)}{\sqrt{(x_2 - x_1)^2 + (y_2 - y_1)^2}} \end{aligned}$$

$$L_1 \perp L_2 \Rightarrow L_1 \perp L_2 \Rightarrow \frac{y_2 - y_1}{x_2 - x_1} \cdot \frac{y_2 - y_1}{x_1 - x_2} = -1$$

Further, distance between two lines $\frac{|c_1 - c_2|}{\sqrt{a^2 + b^2}}$

- In the complex plane: $a=0, L_1 \perp L_2 \Rightarrow a=0$
- In the complex plane: $a=0, L_1 \perp L_2 \Rightarrow a=0$ (no points)
- In the complex plane: $a=0, L_1 \perp L_2 \Rightarrow a=0$ (no points and)
- $a=0$, points



$L_1 \perp L_2 = \{0,0\}$



$L_1 \perp L_2 = \emptyset$

Ex 217 From Eq. (20.11) $M(\psi) = \frac{d^2}{dx^2} [\psi(x) \cos \psi(x)]$, where

$$\psi = \beta \sin \alpha x \cos \alpha x + \gamma$$

In the absence of a dipole, $\alpha = 0$, $M(\psi) = 0$,
 $\psi(x) = \frac{\beta}{2} - \frac{\gamma}{2}$

$$M(\psi) = \cos^2 \left[\frac{\beta - \gamma}{2} \right]$$



$$M(\psi) = \frac{\beta}{2} - \frac{\gamma}{2}$$

$$M(\psi) = \cos^2 \left[\frac{\beta - \gamma}{2} \right]$$



Ex 218 a) Relative distribution amplitude: $\frac{1}{2} \cos^2 \alpha x$

b) Array factor: $M(\psi) = \cos^2 \left[\frac{\beta - \gamma}{2} \right]$



$$a) \cos^2 \left[\frac{\beta - \gamma}{2} \right] = \left(\frac{1}{2} \right)^2$$

$$= \frac{1}{4} \cos^2 \alpha x$$

Half-power beamwidth

$$= 2 \cos^{-1} \cos \alpha x$$

$$= 2\alpha x$$

For uniform array, from Eq. (20.11):

$$\frac{1}{2} \left(\frac{\cos \frac{\beta - \gamma}{2}}{\cos \frac{\beta + \gamma}{2}} \right)^2 = \frac{1}{4} \cos^2 \alpha x$$

Half-power beamwidth for 2-element uniform array
 with $\beta = \gamma$ is $2 \cos^{-1} \cos \alpha x = 2\alpha x$

Ex 219 a) From Eq. (20.11) the array: $M(\psi) = \frac{d^2}{dx^2} \left[\frac{\cos^2 \psi(x)}{\cos^2 \psi(x)} \right]$



1) Resonance Operation $\phi = \phi_{res} = 0$

$$|Z_{eq}| = \frac{1}{\omega} \left| \frac{R + j\omega L}{1 - \omega^2 LC} \right| = \left| \frac{R}{\omega} \right| \quad \text{for } \phi = 0$$

where $Z = R + j\omega L$

At half-power points: $\left| \frac{R}{\omega} \right| = \frac{1}{\sqrt{2}} R \implies \omega = 0.707 R$

(For each bandwidth, another operating)

For bandwidth operation, the half-power bandwidth

$$\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$$

$$= 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$$

For $\omega = 1$, $\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$

From Eq. (1) & (2): $\Delta \omega_{BW} = 0.707 \left(\frac{R}{L} \right) \text{ rad/s}$

2) Resonance Operation $\phi = \phi_{res} = 0$

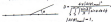
$$|Z_{eq}| = \frac{1}{\omega} \left| \frac{R + j\omega L}{1 - \omega^2 LC} \right| = \left| \frac{R}{\omega} \right| \text{ rad/s}$$

For $\omega = 1$, $|Z_{eq}| = \left| \frac{R}{\omega} \right| \text{ rad/s}$

From Eq. (1) & (2): $|Z_{eq}| = \left| \frac{R}{\omega} \right| \text{ rad/s}$

3) Resonance Operation $\phi = \phi_{res} = 0$

Resonance operation: $\phi = \phi_{res} = 0$



$$\int_0^{\infty} \left(\frac{R}{\omega} \right)^2 d\omega = \frac{R^2}{\omega^2} \int_0^{\infty} \frac{1}{\omega^2} d\omega = \frac{R^2}{\omega^2} \left(\frac{1}{\omega} \right) = \frac{R^2}{\omega^3}$$

$$\therefore \omega = \frac{R}{\omega^3} = \frac{R}{\omega^3} \quad \text{where } \omega = \text{angular frequency}$$

4) Resonance Operation $\phi = \phi_{res} = 0$



ϕ is at $\phi = 0$

Radius of circle is

$$R = \frac{R}{2} \left(\frac{1}{R} \right) = \frac{1}{2}$$

Ex-21 a) $A(x) = 1 + x$, $B = x^2$ ψ at π

b) $A(x) = (1+x)^2$ A double zero $\psi_1 = \psi_2 = 0$

c) $A(x) = \frac{x^2+1}{x-1}$ A zero at ψ_1 , $\psi_2 = 1, 1, 1, 1$

d) $A(x) = (1+x+x^2) = \left(\frac{x^3-1}{x-1}\right)$

Double zero $(\psi_1 = \psi_2 = 1, 1)$

$(\psi_3 = \psi_4 = 1, 1)$

- e) The zeros of an array polynomial specify the nulls in the array pattern as it changes from 0 to π . If π ($\psi = \pi$) gives the main beam. The regions between the nulls (except the main beam region) are sidelobe. The double zero of the array is nulls are more widely spaced leading to a wider beamwidth and lower sidelobe level. $\psi = \pi$ is lower than that of a three-element uniform array.

Ex-22 From Ex. 11-10 and 11-11:

$$|E_{\theta}| = \frac{1}{\sqrt{2}} \left[\cos^2 \left(\frac{\psi}{2} \right) + \cos^2 \left(\frac{3\psi}{2} \right) \right]$$

where $A_1(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(\psi/2)}{\psi/2}$, $\psi_1 = \frac{\pi}{2}$ and $\psi_2 = \frac{3\pi}{2}$

$A_2(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(3\psi/2)}{3\psi/2}$, $\psi_1 = \frac{\pi}{3}$ and $\psi_2 = \frac{2\pi}{3}$

$$|E_{\theta}(\psi)| = \frac{1}{\sqrt{2}} \left[\left(\frac{\sin(\psi/2)}{\psi/2} \right)^2 + \left(\frac{\sin(3\psi/2)}{3\psi/2} \right)^2 \right]$$

Ex-23 From Ex. 11-10 and 11-11 $A_1(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(\psi/2)}{\psi/2}$, $\psi_1 = \frac{\pi}{2}$ $\psi_2 = \frac{3\pi}{2}$ $\psi_3 = \frac{5\pi}{2}$ $\psi_4 = \frac{7\pi}{2}$ $\psi_5 = \frac{9\pi}{2}$ $\psi_6 = \frac{11\pi}{2}$ $\psi_7 = \frac{13\pi}{2}$ $\psi_8 = \frac{15\pi}{2}$ $\psi_9 = \frac{17\pi}{2}$ $\psi_{10} = \frac{19\pi}{2}$ $\psi_{11} = \frac{21\pi}{2}$ $\psi_{12} = \frac{23\pi}{2}$ $\psi_{13} = \frac{25\pi}{2}$ $\psi_{14} = \frac{27\pi}{2}$ $\psi_{15} = \frac{29\pi}{2}$ $\psi_{16} = \frac{31\pi}{2}$ $\psi_{17} = \frac{33\pi}{2}$ $\psi_{18} = \frac{35\pi}{2}$ $\psi_{19} = \frac{37\pi}{2}$ $\psi_{20} = \frac{39\pi}{2}$ $\psi_{21} = \frac{41\pi}{2}$ $\psi_{22} = \frac{43\pi}{2}$ $\psi_{23} = \frac{45\pi}{2}$ $\psi_{24} = \frac{47\pi}{2}$ $\psi_{25} = \frac{49\pi}{2}$ $\psi_{26} = \frac{51\pi}{2}$ $\psi_{27} = \frac{53\pi}{2}$ $\psi_{28} = \frac{55\pi}{2}$ $\psi_{29} = \frac{57\pi}{2}$ $\psi_{30} = \frac{59\pi}{2}$ $\psi_{31} = \frac{61\pi}{2}$ $\psi_{32} = \frac{63\pi}{2}$ $\psi_{33} = \frac{65\pi}{2}$ $\psi_{34} = \frac{67\pi}{2}$ $\psi_{35} = \frac{69\pi}{2}$ $\psi_{36} = \frac{71\pi}{2}$ $\psi_{37} = \frac{73\pi}{2}$ $\psi_{38} = \frac{75\pi}{2}$ $\psi_{39} = \frac{77\pi}{2}$ $\psi_{40} = \frac{79\pi}{2}$ $\psi_{41} = \frac{81\pi}{2}$ $\psi_{42} = \frac{83\pi}{2}$ $\psi_{43} = \frac{85\pi}{2}$ $\psi_{44} = \frac{87\pi}{2}$ $\psi_{45} = \frac{89\pi}{2}$ $\psi_{46} = \frac{91\pi}{2}$ $\psi_{47} = \frac{93\pi}{2}$ $\psi_{48} = \frac{95\pi}{2}$ $\psi_{49} = \frac{97\pi}{2}$ $\psi_{50} = \frac{99\pi}{2}$

Using Ex. 11-10 and 11-11 $A_2(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(3\psi/2)}{3\psi/2}$ $\psi_1 = \frac{\pi}{3}$ $\psi_2 = \frac{2\pi}{3}$ $\psi_3 = \frac{4\pi}{3}$ $\psi_4 = \frac{5\pi}{3}$ $\psi_5 = \frac{7\pi}{3}$ $\psi_6 = \frac{8\pi}{3}$ $\psi_7 = \frac{10\pi}{3}$ $\psi_8 = \frac{11\pi}{3}$ $\psi_9 = \frac{13\pi}{3}$ $\psi_{10} = \frac{14\pi}{3}$ $\psi_{11} = \frac{16\pi}{3}$ $\psi_{12} = \frac{17\pi}{3}$ $\psi_{13} = \frac{19\pi}{3}$ $\psi_{14} = \frac{20\pi}{3}$ $\psi_{15} = \frac{22\pi}{3}$ $\psi_{16} = \frac{23\pi}{3}$ $\psi_{17} = \frac{25\pi}{3}$ $\psi_{18} = \frac{26\pi}{3}$ $\psi_{19} = \frac{28\pi}{3}$ $\psi_{20} = \frac{29\pi}{3}$ $\psi_{21} = \frac{31\pi}{3}$ $\psi_{22} = \frac{32\pi}{3}$ $\psi_{23} = \frac{34\pi}{3}$ $\psi_{24} = \frac{35\pi}{3}$ $\psi_{25} = \frac{37\pi}{3}$ $\psi_{26} = \frac{38\pi}{3}$ $\psi_{27} = \frac{40\pi}{3}$ $\psi_{28} = \frac{41\pi}{3}$ $\psi_{29} = \frac{43\pi}{3}$ $\psi_{30} = \frac{44\pi}{3}$ $\psi_{31} = \frac{46\pi}{3}$ $\psi_{32} = \frac{47\pi}{3}$ $\psi_{33} = \frac{49\pi}{3}$ $\psi_{34} = \frac{50\pi}{3}$ $\psi_{35} = \frac{52\pi}{3}$ $\psi_{36} = \frac{53\pi}{3}$ $\psi_{37} = \frac{55\pi}{3}$ $\psi_{38} = \frac{56\pi}{3}$ $\psi_{39} = \frac{58\pi}{3}$ $\psi_{40} = \frac{59\pi}{3}$ $\psi_{41} = \frac{61\pi}{3}$ $\psi_{42} = \frac{62\pi}{3}$ $\psi_{43} = \frac{64\pi}{3}$ $\psi_{44} = \frac{65\pi}{3}$ $\psi_{45} = \frac{67\pi}{3}$ $\psi_{46} = \frac{68\pi}{3}$ $\psi_{47} = \frac{70\pi}{3}$ $\psi_{48} = \frac{71\pi}{3}$ $\psi_{49} = \frac{73\pi}{3}$ $\psi_{50} = \frac{74\pi}{3}$ $\psi_{51} = \frac{76\pi}{3}$ $\psi_{52} = \frac{77\pi}{3}$ $\psi_{53} = \frac{79\pi}{3}$ $\psi_{54} = \frac{80\pi}{3}$ $\psi_{55} = \frac{82\pi}{3}$ $\psi_{56} = \frac{83\pi}{3}$ $\psi_{57} = \frac{85\pi}{3}$ $\psi_{58} = \frac{86\pi}{3}$ $\psi_{59} = \frac{88\pi}{3}$ $\psi_{60} = \frac{89\pi}{3}$ $\psi_{61} = \frac{91\pi}{3}$ $\psi_{62} = \frac{92\pi}{3}$ $\psi_{63} = \frac{94\pi}{3}$ $\psi_{64} = \frac{95\pi}{3}$ $\psi_{65} = \frac{97\pi}{3}$ $\psi_{66} = \frac{98\pi}{3}$ $\psi_{67} = \frac{100\pi}{3}$

a) Substituting $\psi = \frac{\pi}{2}$ in $A_1(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(\psi/2)}{\psi/2}$ $A_1(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(\pi/4)}{\pi/4} = \frac{1}{\sqrt{2}} \frac{\frac{\sqrt{2}}{2}}{\pi/4} = \frac{1}{\sqrt{2}} \frac{\sqrt{2}}{\pi} = \frac{1}{\pi}$

b) Also value of $A_2(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(3\psi/2)}{3\psi/2}$

For $\psi = \frac{\pi}{2}$ $A_2(\psi) = \frac{1}{\sqrt{2}} \frac{\sin(3\pi/4)}{3\pi/4} = \frac{1}{\sqrt{2}} \frac{\frac{\sqrt{2}}{2}}{3\pi/4} = \frac{1}{\sqrt{2}} \frac{\sqrt{2}}{3\pi} = \frac{1}{3\pi}$

Ex:12 a) $E_p = 10^{-12} \text{ J}$, $E_p = \frac{1}{2} m v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right)$ (1)

$$E_p = \frac{1}{2} m_0 v^2 \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) = \frac{1}{2} m_0 v^2 \gamma$$

$$= \frac{1}{2} m_0 v^2 \left(\frac{1 + \frac{v^2}{c^2}}{\sqrt{1 - \frac{v^2}{c^2}}} \right) \quad (2)$$

$$\therefore E_p = \frac{1}{2} m_0 v^2 = \frac{1}{2} m_0 v^2 \left(\frac{1 + \frac{v^2}{c^2}}{\sqrt{1 - \frac{v^2}{c^2}}} \right) = \frac{1}{2} m_0 v^2 \gamma$$

Ex:13 From eq (1) & (2): $\frac{E_p}{m_0 c^2} = \left(\frac{1}{\sqrt{1 - \frac{v^2}{c^2}}} \right) \frac{1}{2} \frac{v^2}{c^2}$

a) For half mass at poles: $E_p = E_{p0} = 1.0 \text{ J}$

$$E = \frac{1}{2} m_0 v^2 \gamma, \quad m_0 = 1 \text{ kg}, \quad c = 3 \times 10^8 \text{ m/s}, \quad \gamma = \frac{1}{\sqrt{1 - \frac{v^2}{c^2}}}$$

$$E_p = 1.0 \text{ J} = \frac{1}{2} m_0 v^2 \gamma = 0.5 \text{ J} = 0.5 \text{ J} \quad (3) = 0.5 \text{ J} \text{ (given)}$$

b) For kinetic at poles: $E_p = E_{k0} = 1.0 \text{ J}$

$$E_p = 1.0 \text{ J} \text{ (given)}$$

Ex:14 From given: Earth radius = 6370 km

Geostationary orbit radius = 42,164 km



$$\psi = \cos^{-1} \left(\frac{2 \times 6370}{42164} \right) = \cos^{-1} 0.302 = 72.5^\circ$$

$$\psi = \cos^{-1} 0.302 = 72.5^\circ$$

a) Two satellites would cover only

$$2 \times (72.5^\circ) = 145.0^\circ < 120^\circ$$

Thus three satellites in this
geostationary plane

$$3 \times (72.5^\circ) = 217.5^\circ > 120^\circ$$

$\psi = 72.5^\circ$ cannot cover the
polar regions.

b) Let E_1 = Power transmitted by satellite antenna

$$E_2 = \text{Power density within the cone} = \frac{E_1}{4\pi r^2} \theta$$

$$\text{Area of cone cap on earth} = \pi \int_0^{\theta} r^2 \sin \theta d\theta$$

$$= 2\pi r^2 (1 - \cos \theta) \text{ (see Ex:11)} = 2\pi r^2 \theta$$

$$\therefore E_2 = 2\pi r^2 \theta \frac{E_1}{4\pi r^2} = \frac{1}{2} E_1 \theta = \frac{1}{2} E_1 \theta$$

$$\text{Major lobe diameter} = 2\theta = \theta_1 \theta_2$$

Ex-11 (i) From Eq (1) and (2) $A_1 = \frac{2.5 \times 10^4}{2.5 \times 10^4} A_2$

$$A_1 = \frac{2.5 \times 10^4}{2.5 \times 10^4} = 1.0 \text{ m}^2 \text{ (Ans)}, \quad Q_1 = v_1 A_1 = 1.0 \times 1.0 = 1.0 \text{ m}^3/\text{s}$$

$$A_2 = \frac{2.5 \times 10^4}{2.5 \times 10^4} = 1.0 \text{ m}^2 \text{ (Ans)}, \quad Q_2 = v_2 A_2 = 1.0 \times 1.0 = 1.0 \text{ m}^3/\text{s}$$

$$F = 1.0 \times 1.0 = 1.0 \text{ N (Ans)}, \quad A_3 = 1.0 \text{ m}^2 \text{ (Ans)}$$

$$\text{--- } A_4 = 1.0 \text{ m}^2 \text{ (Ans)}$$

(ii) From Eq (1) and (2) $A_1 = \frac{2.5 \times 10^4}{2.5 \times 10^4} A_2$

$$A_1 = \frac{2.5 \times 10^4}{2.5 \times 10^4} = 1.0 \text{ m}^2 \text{ (Ans)}, \quad \text{--- } A_2 = 1.0 \text{ m}^2 \text{ (Ans)}$$

Ex-12 (i) From Eq (1) and (2) $A_1 = A_2 \left(\frac{v_2}{v_1} \right)^2$

where, from Eq (1) and (2), $v_2 = 4 \left[\frac{2.5 \times 10^4}{2.5 \times 10^4} \right] v_1$
 $= 4 v_1$
 $= 4 \times 1.0 = 4.0 \text{ m/s}$

Using (i) $A_1 = 1.0 \times (4.0)^2 = 16.0 \text{ m}^2$

(ii) $A_1 = A_2 \left(\frac{v_2}{v_1} \right)^2 = 16.0 \text{ m}^2$

$$\text{--- } A_2 = 1.0 \text{ m}^2 \text{ (Ans)}$$

Ex-13



$$F_{\text{net}} = \rho g h A_2$$

$$= \rho g h \frac{\pi r_2^2}{4} = \frac{\rho g h \pi r_2^2}{4}$$

In the horizontal section,
 $F_{\text{net}} = 0$

$$F_{\text{net}} = \rho g h \frac{\pi r_2^2}{4} = \frac{\rho g h \pi r_2^2}{4}$$

$$= \rho g h \frac{\pi r_2^2}{4} = \frac{\rho g h \pi r_2^2}{4}$$

$$\text{where } F_{\text{net}} = \frac{\rho g h \pi r_2^2}{4}$$

$$(i) A_1 = A_2 \left(\frac{v_2}{v_1} \right)^2, \quad A_2 = A_1 \left(\frac{v_1}{v_2} \right)^2$$

$$E = \int_0^{\infty} E_r dr = \int_0^{\infty} \frac{1}{r} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$\text{In the far zone, } \frac{1}{r} \approx \frac{1}{R} \rightarrow E = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$E_{\text{far}} = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$E_{\text{far}} = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

Q)



Radiation pattern
for linear
antenna at distance R

$$\text{Ex: 11.1} \quad \text{From Eq. (11-11a)} \quad E_{\text{far}} = \frac{1}{R} (E_{\theta} - E_{\phi})$$

$$\text{a) } E_{\theta} = E_0 (E_{\theta} + E_{\phi}) \sin^2 \theta$$

$$\text{From Eq. (11-11a)} \quad |E_{\theta}| = E_0 \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right)$$

$$\text{b) For } E_{\theta} = E_0 (E_{\theta} - E_{\phi}) \sin^2 \theta$$

$$|E_{\theta}| = E_0 \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right)$$

$$\text{c) For } E_{\theta} = E_0 (E_{\theta} + E_{\phi}) \sin^2 \theta, \quad |E_{\theta}| = \frac{1}{2} E_0$$

$$\text{Ex: 11.2} \quad \text{Eq. (11-11a)} \quad E = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$E = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$\text{For circular polarization, } r = R \sin \theta$$

$$\text{a) } E = E_0 \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$= \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$E = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$\therefore E_{\theta} = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

$$\text{b) } E_{\theta} = \frac{1}{R} \int_0^{\infty} \left(\frac{1}{2} \cos \theta - \frac{1}{2} \sin^2 \theta \right) dr$$

Ex-13 Assume $L_2(x, y) = x^2 y$.

$$\begin{aligned}
 R_2(x) &= \int_{x_0}^x \int_{y_0}^y L_2(x, y) dy dx \\
 &= \int_{x_0}^x \left[\frac{xy^2}{2} \Big|_{y_0}^y \right] dx \\
 L_2(x, x) &= \frac{1}{2} x^3 - \frac{1}{2} x_0^3 = \frac{1}{2} x^3 \Big|_{x_0}^x
 \end{aligned}$$

Ex-14 From Ex-13 we have

$$R_2(x) = \int_{x_0}^x L_2(x, y) dy = \frac{1}{2} x^3 \Big|_{x_0}^x$$

(a) In the xy -plane, $\phi = x^3$:

$$\begin{aligned}
 L_2(x) &= \int_{x_0}^x \int_{y_0}^y \phi(x, y) dy dx \\
 &= \int_{x_0}^x \left(\frac{1}{2} x^3 \Big|_{y_0}^y \right) dx \\
 &= \frac{1}{2} x^3 \Big|_{x_0}^x \quad \text{Let } \phi = \frac{1}{2} x^3 \text{ over } \mathbb{R}^2 \\
 L_2(x) &= \frac{1}{2} \left[\frac{1}{2} x^6 \Big|_{x_0}^x \right].
 \end{aligned}$$

(b) Let $\left[\frac{1}{2} x^3 \right]' = \frac{1}{2} x^2$ over $\mathbb{R}^2 = \mathbb{R}^2$.

Half-power integrals: $(x^2)_x = \frac{1}{3} x^3 \Big|_{x_0}^x$

For $x/x_0 = 1/2$, $(x^2)_x = \frac{1}{3} x^3 \Big|_{x_0}^x$
 $= \frac{1}{3} x_0^3 \log 2$.

(c) Let $\frac{1}{2} x^2 = 0 \implies x_0 = \sqrt{2} \frac{1}{2} = \frac{1}{\sqrt{2}}$ over \mathbb{R}^2
 $= \frac{1}{\sqrt{2}} \frac{1}{2} \log 2$.

(d) First solution over $\mathbb{R}^2 = \mathbb{R}^2$,

$$\text{where } L_2 = \frac{1}{2} \left[\frac{1}{2} x^6 \right]$$

\therefore Level of first solution, $L_1 = \frac{1}{2} \log_2 \left[\frac{1}{2} x^6 \right] = \frac{1}{2} \log 2$.

Comparison of Results:

	Uniform State	Triangular State
Active Period	$ab \left(\frac{2a^2b^2}{a^2+b^2} \right)$	$\frac{ab^2}{2} \left(\frac{2a^2b^2}{a^2+b^2} \right)$
Half-power bandwidth	$2\pi \frac{1}{2} \text{ rad}$	$2.14 \frac{1}{2} \text{ rad}$
Location of first null	$2\pi \frac{1}{2} \text{ rad}$	$2\pi \times 1.5 \text{ rad}$
First-null-to-first-zero	11.3 dB	24.7 dB

EX-11 a) In the rectangle, $\phi = \pi/2$:

$$E_{\text{avg}} = 1.1 \int_0^{\pi/2} \cos(\phi) \sin(\phi) d\phi \quad (\text{value of } ab)$$

$$= \frac{1.1}{2} \left[\frac{\sin^2(\phi) + \cos^2(\phi)}{2} \right]_0^{\pi/2} \quad \phi = \frac{\pi}{2} \text{ rad} = \frac{90^\circ}{180^\circ} \times 360^\circ$$

b) Let $\frac{\left(\frac{E^2}{2} \right) \cos \phi}{\left(\frac{E^2}{2} \right) - E^2} = \frac{1}{2} \implies \phi = 1.107 \text{ rad}$

Half-power bandwidth $(2ab)_{\text{HP}} = 2 \times 10^{-3} \cos(1.107 \text{ rad})$

For $a/b = 0.1$, $(2ab)_{\text{HP}} = 2.14 \times \frac{1}{2} \text{ rad}$
 $= 2.14 \frac{1}{2} \text{ rad}$

c) Let $\phi = \frac{\pi}{2} \implies a_n = 2 \times 10^{-3} \left[\frac{1.1^2}{2} \right] = \frac{1}{2} \times \frac{1}{2} \text{ rad}$
 $= 24.7 \frac{1}{2} \text{ rad}$

d) At first null, $\phi = \pi \text{ rad}$:

$$L_n = -20 \log_{10} \frac{\left(\frac{E^2}{2} \right)}{\left(\frac{E^2}{2} \right) - E^2} = 20 \log_{10} 2 = 23.0 \text{ dB}$$

	Uniform State	Circle State
Active Period	$ab \left(\frac{2a^2b^2}{a^2+b^2} \right)$	$\frac{ab^2}{2} \left(\frac{2a^2b^2}{a^2+b^2} \right)$
Half-power bandwidth	$2\pi \frac{1}{2} \text{ rad}$	$2.14 \frac{1}{2} \text{ rad}$
Location of first null	$2\pi \frac{1}{2} \text{ rad}$	$2.14 \frac{1}{2} \text{ rad}$
First-null-to-first-zero	11.3 dB	24.7 dB

The following corrections should be made to PHYSICS AND CHEMISTRY by David H. Sharp. An explanation for any correction should say WHY?

INDEX

Table of New Corrections to David H. Sharp (Ind. No.)

- P. vii, 1st paragraph, 2nd line: Physics — Chem.
- P. 5, Eq. (1-10): Add the first square under the first square—see in 14-1100.
- P. 11, Fig. 1-1: The dashed lines for the bottom face of the object are not 100°. The dot for μ_{21} should be put at the center of the bottom face. The dot for μ_{12} should be put at the center of the upper face.
- P. 46, Eq. (1-100): $\mu_2 \rightarrow \mu_1$
- P. 75, problem 1-1-10: $2n_1^2 \rightarrow 2n_2^2$
- P. 111, Fig. 1-10: Add a short arrow.
- P. 146, problem 1-1-10: Insert $\mu_1 \rightarrow \mu_2$.
- P. 157, Eq. (1-100): In the denominator: $\mu_1 \rightarrow \mu_2$.
- P. 167, problem 1-1, 2nd line: Add μ_1 radius μ_2 and "After turning".
- P. 167, problem 1-1 (1-10) \rightarrow (1-11).
- P. 168, Eq. (1-100): Insert " μ_1 " after the " μ_2 " sign.
- P. 168, Table 1-1, 1st line: The letter E in the 1st row may.
- P. 211, Eq. (2-100): $n_1 \rightarrow n_2$; $\mu_1 \rightarrow \mu_2$; $\mu_2 \rightarrow \mu_1$; $\mu_1 \rightarrow \mu_2$; $\mu_2 \rightarrow \mu_1$; $\mu_1 \rightarrow \mu_2$; $\mu_2 \rightarrow \mu_1$
- P. 221, line 1: Insert $\frac{\mu_1 \mu_2}{\mu_1 + \mu_2}$ after the word $\frac{\mu_1 \mu_2}{\mu_1 + \mu_2}$.
- P. 221, problem 2-1-10: Insert $\mu_1 \mu_2$ before the word " μ_1 ".
- P. 221, problem 2-1-10, 11: $1.0 \rightarrow 1.5$.
- P. 247, problem 2-1, line 1: Change "the wave impedance is μ_1 " to "the constant magnetic permeability and a wave impedance equal to".
- P. 251, line 1: $\mu_1 \rightarrow \mu_2$; problem 2-1-10, line 1: reflection \rightarrow refraction.
- P. 251, Fig. 2-100: Make the segment between E and μ_2 —(see Fig. 2-10, p. 107.)
- P. 251, 2nd line from bottom: Fig. 2-100 \rightarrow Fig. 2-100A.
- P. 251, problem 2-1-10: 1st line: $\mu_2 \rightarrow \mu_1$; E referring to Fig. 100.
- P. 251, Fig. 2-10: Insert $\frac{\mu_1 \mu_2}{\mu_1 + \mu_2}$ before "Electric field lines" and at the bottom "Magnetic field lines".

P. 224, eq. (20-11): $\vec{E} \rightarrow \vec{E}'$ (2 added); P. 22-224: $\vec{E}' \rightarrow \vec{E}$

P. 224, problem 20-44: last line Change "to the field" to "to";
and last before "axis of support".

P. 224, problem 20-44(a): Change "vector" to "moment".

P. 224, paragraph 4, line 4: $\vec{E} \rightarrow \vec{E}'$

P. 224, last. sentence, line 2: 20-224 \rightarrow 20-224.

P. 224, line 10: $\vec{E} \rightarrow \vec{E}'$

P. 224, problem 20-47 (a): \vec{E}' (more space)

P. 224, problem 20-47 (b): 20-47 \rightarrow 20-47.

P. 224, problem 20-48: 20-48 \rightarrow 20-48.

Book references: table \vec{E}' and \vec{E} (see eq. 20-11, p. 22 and eq. 20-12, p. 224)
Energy density

Request to check the explanation about energy and work

P. 21, Fig. 1-1: Molecular line should be centered on the center.

P. 24, Fig. 1-4

P. 24, Fig. 1-5: \vec{v} should be on the center.

P. 24, Fig. 1-6

P. 24, Fig. 1-7: \vec{v} should be on the center.

P. 25, Fig. 1-8: \vec{v} should be centered; the arrow should be
drawn up to touch \vec{v} ; the point on \vec{v} should be \vec{v} (see Fig. 1-7 or 1-8).

P. 25, Fig. 1-9 (the vertical) vertical line should coincide with the \vec{v} of
axis. (see Fig. 1-7 or 1-8).

P. 25, Fig. 1-10: The arrow should pass through the center of the circle.

P. 25, Fig. 1-11: P. 25, Fig. 1-12: P. 25, Fig. 1-13: P. 25, Fig. 1-14.