

تقدم لجنة EICoM الاكاديمية

دفتر لمادة:

كهر ومغناطيسية (2)

من شرح:

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جزيل الشكر للطالبة:

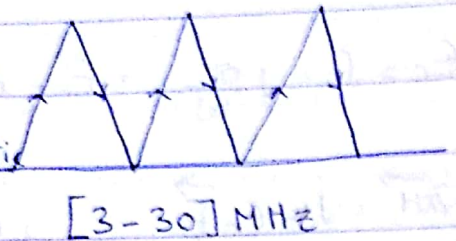
ملك الباشا



16 / 9 / 2018

CH. 7 :- Plane Wave Propagation

Time varying electric field \leftrightarrow Time varying magnetic field



- 1) free space
- 2) guided

* Propagation medium

- 1) lossless medium (ex. air) ($\sigma = 0$)
- 2) lossy medium ($\sigma > 0, \sigma \neq 0$)

7-1 :- Time harmonic fields

* Time varying $\vec{E}, \vec{H}, \vec{D}, \vec{B}, \rho_v, \vec{J}$ depends on (x, y, z) and time

$$E(x, y, z, t) = \text{Re}[\vec{E}(x, y, z) e^{j\omega t}]$$

\downarrow
 $x^* E_x + y^* E_y + z^* E_z$

$$\begin{aligned} \nabla \cdot \vec{D} &= \rho_v & \nabla \cdot \vec{B} &= 0 \\ \nabla \times \vec{E} &= -\frac{d\vec{B}}{dt} & \nabla \cdot \vec{E} &= \frac{\rho_v}{\epsilon_0} \end{aligned}$$

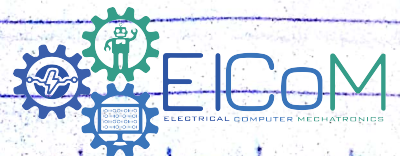
$$\nabla \cdot \vec{H} = 0 \quad \nabla \times \vec{H} = \vec{J} + j\omega \epsilon \vec{E} = \vec{J} + j\omega \vec{D}$$

Complex permittivity

$$\nabla \times \vec{H} = \sigma \vec{E} + j\omega \epsilon_0 \vec{E} = \vec{E}(\sigma + j\omega \epsilon_0) = j\omega \left(\frac{\sigma}{j\omega} + \epsilon_0 \right) \vec{E}$$

$$= j\omega (\epsilon_0 - j\sigma) \vec{E} \rightarrow \text{For lossless medium with } \sigma = 0$$

$$\nabla \times \vec{H} = j\omega \epsilon \vec{E}$$



$$E_c = E - j \frac{\sigma}{\omega} = E' - jE''$$

$$\nabla \times \vec{H} = j\omega \epsilon_c \vec{E} \quad , \quad E' = E_0 \quad , \quad E'' = \frac{\sigma}{\omega}$$

For loss less medium $\sigma = 0$, $E_c = E' = E_0$

* wave equation

$$\nabla \times (\nabla \times \vec{E}) = -j\omega M (\nabla \times \vec{H}) = -j\omega M (j\omega \epsilon_c \vec{E}) = \omega^2 M \epsilon_c \vec{E}$$

$$\nabla \times (\nabla \times \vec{E}) = \nabla (\nabla \cdot \vec{E}) - \nabla^2 \vec{E}$$

$$\nabla^2 \vec{E} = \left(\frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} \right) \vec{E}$$

$$\nabla^2 \vec{E} - \omega^2 M \epsilon_c \vec{E} = 0$$

γ^2 :- propagation constant.

$$\gamma^2 = -\omega^2 M \epsilon_c$$

$$\nabla^2 \vec{E} - \gamma^2 \vec{E} = 0 \quad (\text{wave equation of } \vec{E})$$

$$\nabla^2 \vec{H} - \gamma^2 \vec{H} = 0 \quad (\text{wave equation of } \vec{H})$$

$$\boxed{\nabla^2 \vec{E} - \gamma^2 \vec{E} = 0} \quad ; \quad \nabla^2 \vec{H} - \gamma^2 \vec{H} = 0$$

$$\gamma^2 \triangleq -\omega^2 \mu \epsilon_r$$

$$\epsilon_r = \epsilon' - j\epsilon'' = \epsilon - j\frac{\sigma}{\omega}$$

$$\nabla \cdot \vec{E} = \frac{\partial E_x}{\partial x} + \frac{\partial E_y}{\partial y} + \frac{\partial E_z}{\partial z}$$

$$\vec{E} = x' E_x + y' E_y + z' E_z$$

$$\nabla^2 T = \nabla \cdot (\nabla T) \quad ; \quad T = x' \frac{\partial T}{\partial x} + y' \frac{\partial T}{\partial y} + z' \frac{\partial T}{\partial z}$$

$$\nabla^2 T = \frac{\partial^2 T}{\partial x^2} + \frac{\partial^2 T}{\partial y^2} + \frac{\partial^2 T}{\partial z^2}$$

$$\nabla^2 \vec{E} = x' \nabla^2 E_x + y' \nabla^2 E_y + z' \nabla^2 E_z$$

$$= x' \left[\frac{\partial E_x}{\partial x} + \frac{\partial E_x}{\partial y} + \frac{\partial E_x}{\partial z} \right] + y' \left[\frac{\partial E_y}{\partial x} + \frac{\partial E_y}{\partial y} + \frac{\partial E_y}{\partial z} \right] + z' \left[\frac{\partial E_z}{\partial x} + \frac{\partial E_z}{\partial y} + \frac{\partial E_z}{\partial z} \right]$$

* Plane wave propagation in a "lossless medium"

Propagation property $\begin{cases} \rightarrow \text{phase voltage (up)} \\ \rightarrow \text{wave length } (\lambda) \end{cases}$

goverend by = $\left. \begin{array}{l} \omega ; \text{ angular frequency} \\ \epsilon ; \mu ; \sigma \end{array} \right\}$ depends on

$$\neq \sigma = 0 \rightarrow \epsilon_r = \epsilon - j\frac{\sigma}{\omega} = \epsilon$$

$$\text{Wave number} \Rightarrow k^2 = -\gamma^2 = -(-\omega^2 \mu \epsilon) = \omega^2 \mu \epsilon$$

$$k = \omega \sqrt{\mu \epsilon} \quad ; \quad \nabla^2 \vec{E} + k^2 \vec{E} = 0$$

$$\nabla^2 \vec{E} + k^2 \vec{E} = 0$$

$$\Rightarrow x^{\wedge} \left(\frac{\partial^2 E_x}{\partial x^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2} \right) + y^{\wedge} \left(\frac{\partial^2 E_y}{\partial x^2} + \frac{\partial^2 E_y}{\partial y^2} + \frac{\partial^2 E_y}{\partial z^2} \right) + z^{\wedge} \left(\frac{\partial^2 E_z}{\partial x^2} + \frac{\partial^2 E_z}{\partial y^2} + \frac{\partial^2 E_z}{\partial z^2} \right) + k^2 (x^{\wedge} E_x + y^{\wedge} E_y + z^{\wedge} E_z) = 0$$

$$\rightarrow x^{\wedge} \left(\frac{\partial^2 E_x}{\partial x^2} + \frac{\partial^2 E_x}{\partial y^2} + \frac{\partial^2 E_x}{\partial z^2} \right) + x^{\wedge} E_x k^2 = 0$$

$$\rightarrow y^{\wedge} \left(\frac{\partial^2 E_y}{\partial x^2} + \frac{\partial^2 E_y}{\partial y^2} + \frac{\partial^2 E_y}{\partial z^2} \right) + y^{\wedge} k^2 E_y = 0$$

$$\rightarrow z^{\wedge} \left(\frac{\partial^2 E_z}{\partial x^2} + \frac{\partial^2 E_z}{\partial y^2} + \frac{\partial^2 E_z}{\partial z^2} \right) + z^{\wedge} k^2 E_z = 0$$

$$k = \omega \sqrt{\mu \epsilon}$$

\therefore for uniform plane wave \vec{E} and \vec{H} have uniform properties over cell points of an infinite plane (x-y plane)

$$\frac{\partial E_x}{\partial x} = 0 \quad \frac{\partial E_x}{\partial y} = \frac{\partial E_x}{\partial z} = \frac{\partial E_y}{\partial x} = \frac{\partial E_y}{\partial y}$$

$$= \frac{\partial H_x}{\partial x} = \frac{\partial H_x}{\partial y} = \frac{\partial H_y}{\partial x} = \frac{\partial H_y}{\partial y} = 0$$

$$E_z = H_z = 0$$

$$\nabla \times \vec{H} = j\omega \epsilon \vec{E}$$

x^{\wedge}	y^{\wedge}	z^{\wedge}
$\frac{\partial}{\partial x}$	$\frac{\partial}{\partial y}$	$\frac{\partial}{\partial z}$
H_x	H_y	H_z

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

$$z^{\wedge} \frac{\partial E_x}{\partial x} - \frac{\partial E_y}{\partial y} = -j\omega \mu z^{\wedge} H_z$$

* where we have similar 3 equations

consider z^{\wedge} -component

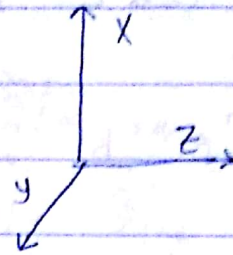
$$z^{\wedge} \left(\frac{\partial H_y}{\partial x} - \frac{\partial H_x}{\partial y} \right) = z^{\wedge} j\omega \epsilon E_z$$

$$\frac{d^2 E_x}{dz^2} + k^2 E_x = 0$$

$$\frac{d^2 E_y}{dz^2} + k^2 E_y = 0$$

$$\frac{d^2 H_x}{dz^2} + k^2 H_x = 0$$

$$\frac{d^2 H_y}{dz^2} + k^2 H_y = 0$$



ASSUME

↑ We have only E_x ($E_y = 0$); we have wave in the z direction only

$$E(z) = \hat{x} E_x = \hat{x} E_{x0} e^{-jkz}, \quad E_y = E_z = 0$$

$$\therefore \text{Re} [\hat{x} E_{x0} e^{-jkz} e^{j\omega t}]$$

$$E(z) = \hat{x} E_{x0} e^{-jkz}$$

$$k^2 = \omega^2 \mu \epsilon$$

$$\text{Re} [\vec{E}(z) e^{j\omega t}]$$

time domain / المجال الزمني

$$kz = 2\pi \Rightarrow \lambda = \frac{2\pi}{k}$$

← H field / المجال المغناطيسي

$$\nabla \times \vec{E} = j\omega \mu \vec{H}$$

\hat{x}	\hat{y}	\hat{z}	
$\frac{d}{dx}$	$\frac{d}{dy}$	$\frac{d}{dz}$	$= -j\omega \mu (\hat{x} H_x + \hat{y} H_y + \hat{z} H_z)$
\vec{E}_x	0	0	

$$\hat{x}(0) + \hat{y} \frac{d\vec{E}_x}{dz} = \hat{z} \frac{d\vec{E}_x}{dy}$$

$$\Rightarrow \hat{y} \frac{d\vec{E}_x}{dz} = -j\omega \mu H_y \hat{y}$$

$$H_y = \frac{1}{-j\omega\mu} \frac{\partial}{\partial z} (E_{x_0}^+ e^{-jkz}) = \frac{1}{j\omega\mu} E_{x_0}^+ e^{-jkz} (-jk)$$

$$= \frac{1}{\omega\mu} E_{x_0}^+ e^{-jkz} k$$

$$H_y = \frac{k}{\omega\mu} E_{x_0}^+ e^{-jkz} = H_{y_0}^+$$

$$\eta \text{ (intrinsic impedance)} = \frac{\omega\mu}{k} = \frac{\omega\mu}{\omega\sqrt{\mu\epsilon}} = \sqrt{\frac{\mu}{\epsilon}} \quad \eta: \text{eata}$$

$$\vec{E}(z) = x^{\wedge} E_{x_0}^+ e^{-jkz}$$

$$\vec{H}(z) = \frac{y^{\wedge}}{\eta} E_{x_0}^+ e^{-jkz} = y^{\wedge} H_{y_0}^+ e^{-jkz}$$

In general $E_{x_0}^+ = |E_{x_0}^+| e^{j\phi}$

$$\eta = \frac{\omega\mu}{k} = \sqrt{\frac{\mu}{\epsilon}}$$

Instantaneous (time domain) \vec{E} and \vec{H}

$$\vec{E}(z,t) = k_e [\vec{E}(z) e^{j\omega t}]$$

$$= \text{Re} [x^{\wedge} |E_{x_0}^+| e^{j\phi} e^{-jkz} e^{j\omega t}]$$

$$= x^{\wedge} [|E_{x_0}^+| \cos(\omega t - kz + \phi)] \quad \text{V/m}$$

$$\vec{H}(z,t) = \text{Re} [y^{\wedge} \vec{H}(z) e^{j\omega t}] = y^{\wedge} \frac{|E_{x_0}^+|}{\eta} \cos(\omega t - kz + \phi) \quad \text{A/m}$$

Note :- \vec{E} and \vec{H} are in-phase for loss less medium (since η is real)

$$\omega t - kz = \text{constant} \Rightarrow z = \frac{\omega t - \text{const}}{k} \Rightarrow \frac{\partial z}{\partial t} = \frac{\omega}{k} = \frac{\omega}{\omega\sqrt{\mu\epsilon}} = \frac{1}{\sqrt{\mu\epsilon}}$$

$$v_p = \frac{\omega}{k} = \frac{1}{\sqrt{\mu\epsilon_0}} = \frac{2\pi f}{k} = \frac{2\pi f}{k}$$

$$\lambda = \frac{2\pi}{k} = \frac{v_p}{f} \Rightarrow k = \frac{2\pi}{\lambda} \quad v_p = \frac{\omega}{k} \quad \lambda = \frac{v_p}{f}$$

In free space $\mu = \mu_0 = 4\pi \times 10^{-7}$; $\epsilon = \epsilon_0 = \frac{10^{-9}}{36\pi}$

$$c = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \approx 3 \times 10^8 \text{ m/s}$$

$$\eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = 377 \Omega = 120 \pi$$

Example:- plane wave traveling in the (+z) direction, the electric field points along the peak value of $E = 1.2 \pi \text{ mV/m}$; E has a maximum at $(t=0, z=50\text{m})$; find the time domain expression

$$\vec{E}(z,t), \vec{H}(z,t), \text{ let } E_x = X^{\wedge} |E_x^{\wedge}| e^{j\phi} e^{-jkz}$$

(simple) \Rightarrow

$$c = f \lambda \Rightarrow 3 \times 10^8 = 10^6 \lambda \Rightarrow \lambda = 300 \text{ m}$$

$$k = \frac{2\pi}{\lambda} = \frac{2\pi}{300} \quad \text{Re} [X^{\wedge} 1.2\pi e^{j\phi} e^{-jkz} e^{j\omega t}] = X^{\wedge} 1.2 \cos$$

$$\vec{E}(z,t) = X^{\wedge} 1.2\pi \cos \left(2\pi \times 10^6 t - \frac{2\pi}{300} z + \phi \right)$$

Peak at $t=0, z=50\text{m} \Rightarrow \frac{-2\pi}{300} * 50 + \phi = 0 \Rightarrow \phi = \frac{\pi}{3}$

$$\vec{E}(z,t) = X^{\wedge} 1.2\pi \times 10^{-3} \cos \left(2\pi \times 10^6 t - \frac{2\pi}{300} z + \frac{\pi}{3} \right)$$

$$\vec{H}(z,t) = \hat{y} \cdot 2\pi \cdot 10^{-3} \cos\left(2\pi \cdot 10^6 t - \frac{2\pi}{300} z + \frac{\pi}{3}\right)$$

~~Example~~ General relation between \vec{E} and \vec{H}

$$\vec{H} = \frac{1}{\eta} \hat{k} \times \vec{E} \quad \text{uniform plane wave in a direction of unit vector } \hat{k}$$

$$\vec{E} = -\eta \hat{k} \times \vec{H}$$

$$\text{If } \vec{E} = \hat{x} E_x(z) = \hat{x} E_{x0}^+ e^{-jkz} \quad \hat{k} = \hat{z}$$

$$\vec{H} = \frac{1}{\eta} \hat{z} \times \hat{x} E_{x0}^+ e^{-jkz} = \hat{y} \frac{E_{x0}^+}{\eta} e^{-jkz}$$

If the wave is travelling in the -ve (z) direction $\vec{E} = \hat{x} E_{x0}^+ e^{jkz}$

$$\hat{k} = -\hat{z}$$

$$\vec{H} = \frac{1}{\eta} -\hat{k} \times \vec{E} = \frac{1}{\eta} -\hat{k} \times \hat{x} E_{x0}^+ e^{jkz}$$

$$\vec{H} = -\hat{y} \frac{1}{\eta} E_{x0}^+ e^{jkz}$$

In general :-

$$\vec{E} = \hat{x} E_x(z) + \hat{y} E_y(z) \quad \hat{k} = \hat{z}$$

$$\vec{H} = \frac{1}{\eta} \hat{z} \times [\hat{x} E_x(z) + \hat{y} E_y(z)] = \frac{1}{\eta} \hat{y} E_x(z) - \frac{1}{\eta} \hat{x} E_y(z)$$

$$H_x(z) = -\frac{E_y(z)}{\eta}, \quad H_y(z) = \frac{E_x(z)}{\eta}$$

Example:- 10 MHz, plane wave travelling in anon-magnetic material and $\epsilon_r = 9$

- a) phase velocity b) wave number (k) c) wave length in the medium.
 d) Interinsic impedant.

$$UP = \frac{1}{\sqrt{\mu \epsilon}} = \frac{1}{\sqrt{\mu \epsilon_0 \epsilon_r}} = \frac{c}{\sqrt{\epsilon_r}} \quad \text{where } c = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \quad \text{and } \omega = \frac{2\pi f}{\text{rad/sec}}$$

$$UP = \frac{c}{\sqrt{\epsilon_r}} \Rightarrow \epsilon_r = \left(\frac{c}{UP} \right)^2$$

$$UP = \frac{1}{\sqrt{\mu \epsilon}}$$

$$UP = \frac{1}{\sqrt{\mu \epsilon_0 \epsilon_r}}$$

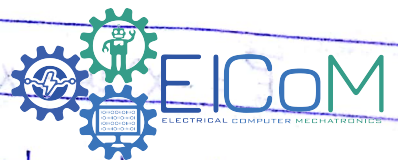
$$UP = \frac{3 \times 10^8}{\sqrt{9}} \Rightarrow UP = 1 \times 10^8 \text{ m/sec}$$

$$UP = \frac{\omega}{k} \Rightarrow k = \frac{\omega}{UP} \Rightarrow k = \frac{2\pi \times 10^7}{10^8} = 0.2\pi$$

$$k = \omega \sqrt{\mu \epsilon} =$$

$$\lambda = \frac{2\pi}{k} = \frac{2\pi}{0.2\pi} = 10 \text{ m}$$

$$\eta = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} = \sqrt{\frac{\mu_0}{\epsilon_0}} \frac{1}{\sqrt{\epsilon_r}} = \frac{120\pi}{\sqrt{9}} = 127.67 \Omega$$



example:- Given $\vec{E} = \hat{z} 10 e^{-j4\pi y}$ (mV/m) wave is travelling in a lossless medium with $n = 188.5$, Find

- (a) magnetic field phasor (b) Instantaneously $\vec{E}(y,t)$ if the medium is non magnetic.

$$\hat{k} = \hat{y}$$

$$\vec{H} = \frac{1}{n} \hat{k} \times \vec{E} = \frac{1}{188.5} \hat{y} \times \hat{z} 10 e^{-j4\pi y} = \hat{x} 53 e^{-j4\pi y}$$

$$k = \omega \sqrt{\mu \epsilon} \Rightarrow \omega_0 = \frac{k}{\sqrt{\mu_0 \epsilon_0 \epsilon_r}} = \frac{kc}{\sqrt{\epsilon_r}}$$

example:- $\vec{H}(z,t) = \hat{z} 30 \cos(10^8 t - 0.5z)$ mA/m ; In a non magnetic direction of the wave propagation.

(b) Phase velocity (c) λ (d) ϵ_r

(e) phasor.

(a) $\hat{k} = \hat{y}$ (b) $UP = \frac{1}{\sqrt{\mu \epsilon}} = \frac{\omega}{k} \Rightarrow k = 0.5$

$$UP = \frac{10^8}{0.5} = 2 \times 10^8 \text{ m/sec}$$

(c) $UP = f\lambda \Rightarrow \lambda = \frac{2 \times 10^8}{0.5} \Rightarrow \lambda = 4\pi$

(d) $\epsilon_r = \left(\frac{c}{UP}\right)^2$

(e) $n = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} = \frac{n_0}{\sqrt{\epsilon_r}} = \frac{120\pi}{\sqrt{2.25}} = 251.3$

$$\vec{E} = -\hat{x} 25.3 \times 30 \cos(10^8 t - 0.5z)$$

example: Non-magnetic ($\mu = \mu_0$); $\vec{E}(z,t) = \hat{y} 3 \sin(\pi \times 10^7 t - 0.2\pi x) + \hat{z} 4 \cos(\pi \times 10^7 t - 0.2\pi x)$ V/m

Find: (a) λ (b) ϵ_r (c) \vec{H} time domain

$$(a) \lambda = \frac{2\pi}{k} = \frac{2\pi}{0.2\pi} = 10 \text{ m} = \lambda$$

$$(b) k = \omega \sqrt{\mu \epsilon_0} \Rightarrow \epsilon_r = \left(\frac{c}{v_p}\right)^2 = \left(\frac{3 \times 10^8}{5 \times 10^7}\right)^2 \Rightarrow \epsilon_r = 36$$

$\mu = \mu_0 \rightarrow$ lossless

$$v_p = \frac{1}{\sqrt{\mu \epsilon_0}} = \frac{\omega}{k} = \frac{\pi \times 10^7}{0.2\pi} = 5 \times 10^7 \text{ m/s}$$

(c) \vec{H} time domain

$$n = n_0 = \frac{120\pi}{\sqrt{\epsilon_r}} = 62.83 \Omega$$

$$\vec{H} = \frac{1}{n} \hat{k} \times \vec{E} \quad ; \hat{k} = \hat{x}$$

$$= \frac{1}{62.83} \hat{x} \times [\hat{y} 3 \sin(\pi \times 10^7 t - 0.2\pi x) + \hat{z} 4 \cos(\pi \times 10^7 t - 0.2\pi x)]$$

$$= \hat{z} 47.7 \sin(\pi \times 10^7 t - 0.2\pi x) - \hat{y} 63.66 \cos(\pi \times 10^7 t - 0.2\pi x)$$

Wave polarization

The locus of the tip of the \vec{E} -vector (in the plane orthogonal to direction of propagation) at a given point in space as a function of time.

* General polarization \rightarrow elliptical

linear \rightarrow circular

$$\vec{E} = \hat{x} E_x(z) + \hat{y} E_y(z)$$

* Polarization depends on :-

① phase of E_{y0} relative to E_{x0} .

② Absolute value of E_{x0} , E_{y0} .

$$E_x(z) = E_{x0} e^{-jkz}$$

$$E_y(z) = E_{y0} e^{-jkz}$$

$$E_{x0} = a_x \leftarrow \text{reference (phase 0)}$$

$$E_{y0} = a_y e^{j\delta} \quad \delta = \phi_y - \phi_x$$

$$E_{x0} = a_x e^{-jkz}$$

$$E_{y0} = a_y e^{-jkz} e^{j\delta}$$

$$E_{y0} = a_y e^{-jkz} e^{j\delta}$$

$$\vec{E}(z) = (\hat{x} a_x + \hat{y} a_y e^{j\delta}) e^{-jkz}$$

* Instantaneous

$$E(z, t) = \text{Re} [\vec{E}(z) e^{j\omega t}]$$

$$= \hat{x} a_x \cos(\omega t - kz) + \hat{y} a_y \cos(\omega t - kz + \delta)$$

Modes of $E(z, t)$

$$\textcircled{1} E(z, t) = [E_x(z, t)] + [E_y(z, t)]$$

$$= a_x \cos(\omega t - kz) + a_y \cos(\omega t - kz + \delta)$$

$$\psi(z, t) = \tan^{-1} \left(\frac{E_y(z, t)}{E_x(z, t)} \right)$$

linear polarization

Close $Z=0$

$S = 0$ (in phase) , $S = -\pi$ (out of phase)

1) $S = 0 \Rightarrow E(0,t) = (x^{\wedge} a_x + y^{\wedge} a_y) \cos(\omega t - kz)$

2) $S = \pi \Rightarrow \vec{E} = (x^{\wedge} a_x - y^{\wedge} a_y) \cos(\omega t - kz)$

$$|\vec{E}(0,t)| = \sqrt{a_x^2 + a_y^2} \cos(\omega t - kz)$$

$$\Psi(t) = \tan^{-1} \left(\frac{E_y(t)}{E_x(t)} \right) =$$

in phase $\beta \Rightarrow \Psi(t) = \tan^{-1} \left(\frac{a_y}{a_x} \right)$

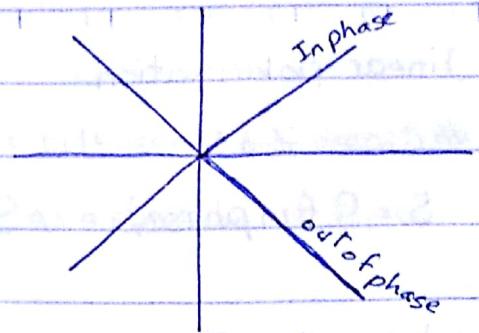
out phase $\Rightarrow \Psi(t) = \tan^{-1} \left(\frac{-a_y}{a_x} \right)$

if $a_y = 0 \Rightarrow \Psi = 0$ or $180^\circ \Rightarrow X$ -polarized

if $a_x = 0 \Rightarrow \Psi = 90^\circ$ or $270^\circ \Rightarrow Y$ -polarized

$$x^{\wedge} a_x + y^{\wedge} a_y e^{j\delta} \quad , \quad \delta = 0, \delta = \pi$$

$$|E| = \sqrt{a_x^2 + a_y^2} \cos \omega t$$



example :- the electric field is $\vec{E}(z,t) = x^{\wedge} 3 \cos(\omega t - kz) + y^{\wedge} 4 \cos(\omega t - kz)$

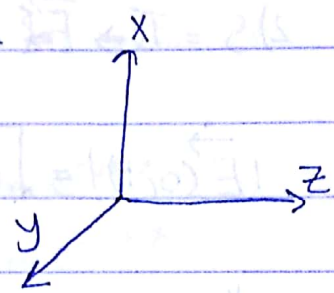
Find :- ① Polarization state. ② Modulus of \vec{E} .

③ Inclination angle.

① Polarization state

$$\delta = \delta_y - \delta_x = 0 - 0 = 0$$

linear polarization (in-phase)

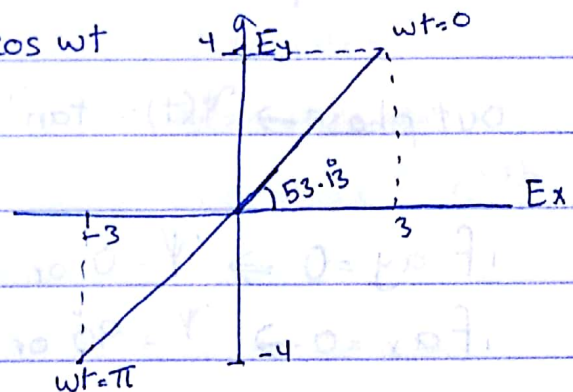


② Modulus of $\vec{E} \cos \omega t$

$$|\vec{E}| = \sqrt{a_x^2 + a_y^2} \cos \omega t = \sqrt{3^2 + 4^2} \cos \omega t = 5 \cos \omega t$$

③ Inclination angle

$$\psi = \tan^{-1}\left(\frac{4}{3}\right) = 53.13^\circ$$



Circular Polarization

1) $a_y = a_x$ 2) $\delta = \pm \pi/2$ $\begin{cases} -\pi/2 \rightarrow \text{right hand circular (RHC)} \\ \pi/2 \rightarrow \text{left hand circular (LHC)} \end{cases}$

LHC polarization ($a_y = a_x, \delta = \pi/2$) = $\delta_y - \delta_x$ (+z \rightarrow)

$$\vec{E}(z) = (x^{\wedge} a_x + y^{\wedge} a_y e^{j\delta}) e^{-jkz}$$

$$\vec{E}(z,t) = R [Ez e^{j\omega t}]$$

$$= x^{\wedge} a_x \cos(\omega t - kz) + y^{\wedge} a_y \cos(\omega t - kz + \frac{\pi}{2})$$

$$\vec{E}(z,t) = \hat{x} a_x \cos(\omega t - kz) + \hat{y} a_y \cos\left(\frac{\pi}{2} - (-\omega t + kz)\right) \Rightarrow \hat{y} a_y \sin(-\omega t + kz)$$

$$\Rightarrow -\hat{y} a_y \sin(\omega t - kz)$$

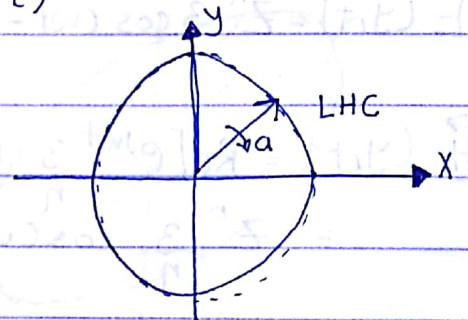
$$\vec{E}(z,t) = \hat{x} a_x \cos(\omega t - kz) - \hat{y} a_y \sin(\omega t - kz)$$

$$|E| = \sqrt{E_x^2 + E_y^2} = \sqrt{a_x^2 \cos^2(\omega t - kz) + a_y^2 \sin^2(\omega t - kz)} \quad \text{let } a_x = a_y = a$$

$$= \sqrt{a^2 \times 1} = a \quad \text{independent on } t \text{ or } z$$

$$\gamma(z,t) = \tan^{-1}\left(\frac{E_y(z,t)}{E_x(z,t)}\right) = \tan^{-1}\left(\frac{-a_y \sin(\omega t - kz)}{a_x \cos(\omega t - kz)}\right) = \tan^{-1}(-\tan(\omega t - kz))$$

$$= -(\omega t - kz) = \boxed{kz - \omega t}$$

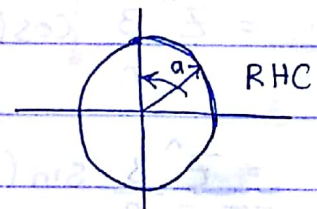


RHC polarization ($a_x = a_y = a$), $\delta = -\pi/2$

$$E(z,t) = \hat{x} a \cos(\omega t - kz) + \hat{y} a \sin(\omega t - kz)$$

$$|\vec{E}| = a$$

$$\gamma(z,t) = \omega t - kz$$



Example :- Find the expression is (RHC, polarization) $E(y,t)$; $H(y,t)$

modulus = 3 mV/m , medium is non magnetic ($\mu = \mu_0$); ($\epsilon = 4\epsilon_0$);

$\sigma = 0$, propagation direction (+y direction)

$$a_x = a_y = 3 \text{ mV/m}, \quad \delta = -\pi/2 = \delta_x - \delta_y \quad \delta_z = 0$$

$$\boxed{\delta = -\pi/2 = \delta_x}$$

$$E(y) = 3 \left(\hat{x} + \hat{z} e^{-j\pi/2} \right) e^{-jk_y y}$$

$$H(y) = \frac{1}{\eta} \hat{k} \times \vec{E} = \frac{1}{\eta} \hat{y} \times 3 \left(\hat{x} + \hat{z} e^{-j\pi/2} \right) e^{-jk_y y} = \frac{3}{\eta} \left(\hat{x} - \hat{z} e^{-j\pi/2} \right) e^{-jk_y y}$$

$$= \frac{3}{\eta} \left(\hat{x} + \hat{z} e^{j\pi/2} \right) e^{-jk_y y}$$

$$k = \omega \sqrt{\mu \epsilon} = \omega \sqrt{\mu_0 \epsilon_0} \sqrt{\epsilon_r} = \frac{\omega}{c} \sqrt{\epsilon_r}$$

$$= \frac{2\pi \times 10^8}{3 \times 10^8} \sqrt{4} = \frac{4\pi}{3}$$

$$\eta = \frac{\eta_0}{\sqrt{\epsilon_r}} = \frac{120\pi}{\sqrt{4}} = 60\pi$$

$$\vec{E}(y, t) = \text{Re}[\vec{E}(y)e^{j\omega t}] = 3 \left[Z^{\wedge} \cos(\omega t - ky) + X^{\wedge} \cos(\omega t - ky - \pi/2) \right] \\ + X^{\wedge} \sin(\omega t - ky)$$

$$\vec{H}(y, t) = \frac{1}{\eta} [E_{y, t}] = \frac{3}{60\pi} [X^{\wedge} \cos(\omega t - ky) - Z^{\wedge} \sin(\omega t - ky)] \text{ mA/m}$$

$$\vec{E}(y, t) = Z^{\wedge} 3 \cos(\omega t - ky) + X^{\wedge} 3 \sin(\omega t - ky) \text{ mV/m}$$

$$\vec{H}(y, t) = \text{Re}[e^{j\omega t} \frac{3}{\eta} (Z^{\wedge} e^{j\pi/2} + X^{\wedge}) e^{-jky}]$$

$$= Z^{\wedge} \frac{3}{\eta} \cos(\omega t - ky + \pi/2) + X^{\wedge} \frac{3}{\eta} \cos(\omega t - ky)$$

$$= Z^{\wedge} \frac{3}{\eta} \cos(\pi/2 - (\omega t - ky)) + X^{\wedge} \frac{3}{\eta} \cos(\omega t - ky)$$

$$= Z^{\wedge} \frac{3}{\eta} \sin(-\omega t + ky) + X^{\wedge} \frac{3}{\eta} \cos(\omega t - ky)$$

$$= -Z^{\wedge} \frac{3}{20\pi} \sin(\omega t - ky) + X^{\wedge} \frac{1}{20\pi} \cos(\omega t - ky)$$

example:- if the electric field phasor is given by $\vec{E}(x) = (y\hat{a}_x - z\hat{a}_y) e^{-jkx}$

find the polarization state.

$$\vec{E} = (x\hat{a}_x + y\hat{a}_y e^{j\delta}) e^{-jkz}$$

$$\vec{E}(x) = (y\hat{a}_x + z\hat{a}_y e^{j\pi/2}) e^{-jkx}$$

$$a_y = a_z = 1$$

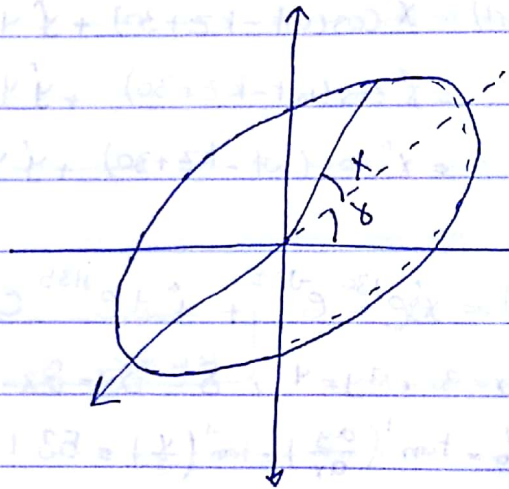
$$\delta = \delta_z - \delta_y = \frac{-\pi}{2} - 0 = -\pi/2 = \delta \Rightarrow \text{RHC polarization}$$

Elliptic polarization

$$a_x \neq 0, a_y \neq 0, \delta \neq 0$$

δ : rotation angle.

χ : ellipticity angle.



$$\vec{E}(z) = (x\hat{a}_x + y\hat{a}_y e^{j\delta}) e^{-jkz}$$

$$\vec{E}(z,t) = x\hat{a}_x \cos(\omega t - kz) + y\hat{a}_y \cos(\omega t - ky)$$

Steps-

$$1) \delta = \delta_y - \delta_x \quad 2) \psi_0 = \tan^{-1} \left(\frac{a_y}{a_x} \right)$$

3) Rotation angle

$$\tan(2\pi) = \rho_{out} = 2\chi \cos(\delta)$$

$$-\pi \leq 2\chi \leq \pi \Rightarrow \frac{-\pi}{2} \leq \delta \leq \frac{\pi}{2}$$

$$\begin{cases} \delta \geq 0 & \text{if } \cos \delta > 0 \\ \delta < 0 & \text{if } \cos \delta < 0 \end{cases}$$

4) elliptically angle (χ)

$$\sin(2\chi) = \sin(2\psi_0) \sin \delta$$

$$\chi > 0 \quad \text{if } \sin \delta > 0$$

$$\chi < 0 \quad \text{if } \sin \delta < 0$$

$$-\frac{\pi}{4} \leq \chi \leq \frac{\pi}{4}$$

$$-\frac{\pi}{2} \leq 2\chi \leq \frac{\pi}{2}$$

example 1- find the polarization state of a plane wave

$$E(z,t) = \hat{x} 3 \cos(\omega t - kz + 30^\circ) - \hat{y} 4 \sin(\omega t - kz + 45^\circ) \text{ mV/m}$$

$$\begin{aligned} \vec{E}(z,t) &= \hat{x} 3 \cos(\omega t - kz + 30^\circ) + \hat{y} 4 \sin(-\omega t + kz - 45^\circ) \\ &= \hat{x} 3 \cos(\omega t - kz + 30^\circ) + \hat{y} 4 \cos(90^\circ + \omega t - kz + 45^\circ) \\ &= \hat{x} 3 \cos(\omega t - kz + 30^\circ) + \hat{y} 4 \cos(\omega t - kz + 135^\circ) \end{aligned} \quad \left. \begin{array}{l} -\hat{y} 4 \cos(90^\circ - \omega t + kz - 45^\circ) \\ -\hat{y} 4 \cos(45^\circ - \omega t + kz) \\ -\hat{y} 4 \cos(\omega t - kz - 45^\circ) \\ -\hat{y} 4 e^{-j45^\circ} e^{-jkz} \\ \hat{y} 4 e^{j135^\circ} e^{-jkz} \end{array} \right\}$$

$$\vec{E}(z) = \hat{x} 3 e^{j30^\circ} e^{-jkz} + \hat{y} 4 e^{j135^\circ} e^{-jkz}$$

$$a_x = 3, a_y = 4, \delta = \delta_y - \delta_x = 135^\circ - 30^\circ = 105^\circ$$

$$\psi_0 = \tan^{-1}\left(\frac{a_y}{a_x}\right) = \tan^{-1}\left(\frac{4}{3}\right) = 53.1^\circ$$

$$-\tan(2\chi) = \tan(2\psi_0) \cos \delta \Rightarrow \tan(2\chi) = 0.89$$

$$\tan^{-1}(0.89) = 41.67^\circ = 2\chi \Rightarrow \chi = 20.83^\circ$$

$$-(180 - 41.67) = -138.33^\circ$$

$$\delta = \begin{cases} 20.83^\circ \\ -69.16^\circ \end{cases}$$

$$\delta > 0 \quad \text{if } \cos \delta > 0$$

$$\cos(\delta) = \cos(105^\circ) = -ve$$

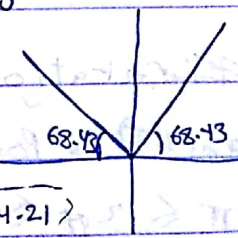
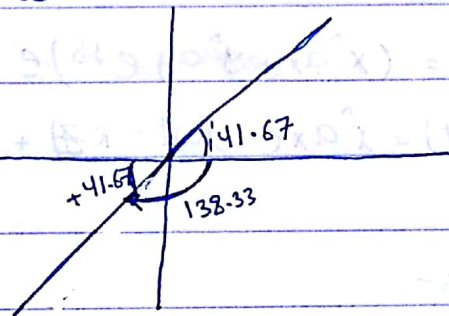
$$\chi = -69.16^\circ$$

$$\sin(2\chi) = \sin(2\psi_0) \sin \delta$$

$$= 0.93$$

$$(0 \rightarrow 90^\circ) (2\chi) = 68.43^\circ$$

$$2\chi \begin{cases} (68.43^\circ) \\ (180 - 68.43^\circ) \end{cases} \Rightarrow \chi \begin{cases} (34.21^\circ) \\ \end{cases}$$



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$$\vec{E}(z,t) = -\hat{x} 10 \sin(\omega t - kz - 60) + \hat{y} 30 \cos(\omega t - kz)$$

$$\begin{aligned} \therefore -\hat{x} 10 \sin(\omega t - kz - 60) &= +\hat{x} 10 \sin(-\omega t + kz + 60) = \hat{x} 10 \cos(\omega t - kz - 60 + 90^\circ) \\ &= \hat{x} 10 \cos(\omega t - kz + 30) \end{aligned}$$

$$\vec{E}(z,t) = \hat{x} 10 \cos(\omega t - kz + 30) + \hat{y} 30 \cos(\omega t - kz)$$

$$\vec{E}(z) = \hat{x} 10 e^{j30^\circ} e^{-jkz} + \hat{y} 30 \cos(\omega t - kz)$$

+ve → LHR

$$\delta = \delta_y - \delta_x = 0 - 30^\circ = -30^\circ, \quad \alpha_x = 10, \quad \alpha_y = 30 \quad \text{-ve} \rightarrow \text{RHR}$$

$$\gamma_0 = \tan^{-1}\left(\frac{30}{10}\right) = 71.56^\circ$$

$$\tan(2\delta) = \tan(2\gamma_0) \tan \delta = -0.648$$

$$(0 \rightarrow 90^\circ) \quad 2\delta = 32.98 \Rightarrow 2\delta = \begin{cases} -32.98 \\ 147.02 \end{cases}$$

$$\cos(\delta) = \cos(-30) = +ve$$

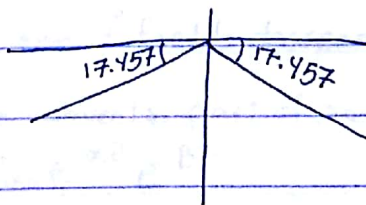
$$2\delta = 147.02 \rightarrow \delta = 73.31$$

-ve → RHR

$$\sin(2\chi) = \sin(2\gamma_0) \sin \delta = -0.3$$

$$(0 \rightarrow 90^\circ) \quad 2\chi = \sin^{-1}(-0.3) = -17.457$$

$$2\chi = \begin{cases} -17.457 & \checkmark \\ -162.54 & \times \end{cases} \quad 2\chi = -17.457 \Rightarrow \chi = -8.728$$



plane wave propagation in Lossy Medium

$$\nabla^2 \vec{E} - \gamma^2 \vec{E} = 0, \quad \gamma^2 = -\omega^2 \mu \epsilon_r = \omega^2 \mu (\epsilon' - j\epsilon'')$$

$$\gamma = \alpha + j\beta$$

$$(\alpha + j\beta)^2 = \alpha^2 - \beta^2 + j2\alpha\beta = -\omega^2 \mu \epsilon' - j\omega^2 \mu \epsilon''$$

$$\alpha^2 - \beta^2 = -\omega^2 \mu \epsilon', \quad 2\alpha\beta = \omega^2 \mu \epsilon''$$

$$\alpha = \omega \left[\frac{\mu \epsilon'}{2} \left[\sqrt{1 + \left(\frac{\epsilon''}{\epsilon'}\right)^2} - 1 \right] \right]^{1/2}$$

$$\beta = \omega \left[\frac{\mu \epsilon'}{2} \left[\sqrt{1 + \left(\frac{\epsilon''}{\epsilon'}\right)^2} + 1 \right] \right]^{1/2}$$

$$e^{-\alpha z} e^{-j\beta z} \Rightarrow Z \propto B = 2\omega^2 \left[\left(\frac{\mu \epsilon'}{2}\right)^2 \left(\frac{\epsilon''}{\epsilon'}\right)^2 \right]^{1/2} = 2\omega \left[\frac{\mu^2 \epsilon''^2}{4} \right]^{1/2}$$

$$\vec{E} = E_{x_0} e^{-\delta z} = E_{x_0} e^{-\alpha z} e^{-j\beta z}$$

$$\vec{E} = \hat{x} E_{x_0} e^{-\alpha z} e^{-j\beta z}$$

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

\hat{x}	\hat{y}	\hat{z}	
d/dx	d/dy	d/dz	$= \hat{x}(0) + \hat{y} \left(\frac{d}{dz} E_{x_0} e^{-\delta z} \right) + \hat{z}(0)$
$E_{x_0} e^{-\delta z}$	0	0	$= -j\omega \mu [\hat{x} H_x + \hat{y} H_y + \hat{z} H_z]$

$$\hat{y} E_{x_0} (-\delta) e^{-\delta z} = \hat{y} (-j\omega \mu H_y)$$

$$H_y = \frac{\delta}{j\omega \mu} E_{x_0} e^{-\delta z} = \frac{E_{x_0}}{n_c} e^{-\alpha z} e^{-j\beta z}$$

~~$$\frac{H_c}{j\omega \mu} = \frac{\delta}{j\omega \mu} \quad n_c = \frac{j\omega \mu}{\delta} = \frac{j\omega \mu}{\sqrt{-\omega^2 \mu \epsilon_c}} = \sqrt{\frac{\mu}{\epsilon_c}}$$~~

$$n_c = \sqrt{\frac{\mu}{\epsilon_c}} = \sqrt{\frac{\mu}{\epsilon' - j\epsilon''}} = \sqrt{\frac{\mu}{\epsilon'} \left(1 - \frac{j\epsilon''}{\epsilon'} \right)^{-1/2}}$$

$$\epsilon' = \epsilon \quad \epsilon'' = \frac{\sigma}{\omega}$$

Both \vec{E} and \vec{H} have no longer equal phase.

Both \vec{E} and \vec{H} fields decrease exponentially with (z) .

Medium converts part of power into heat.

$$\vec{E} = \hat{x} E_{x_0} e^{-\alpha z} e^{-j\beta z}$$

\therefore Skin depth S_s

$S_s = \frac{1}{\alpha}$ at a distance of $z = S_s$ magnitude delves by e^{-1}

S_s characterizes how well an EM wave can penetrate a medium.

1) Perfect dielectric ($\sigma = 0$)

$$S_s = \infty \quad ; \quad \epsilon'' = \sigma / \omega$$

$$\alpha = \omega \left[\frac{\mu \epsilon'}{2} \left(\sqrt{1 + \left(\frac{\epsilon''}{\epsilon'} \right)^2} - 1 \right) \right]^{1/2}$$

2) Perfect conductor ($\sigma = \infty$)

$$\sigma = \infty \Rightarrow \alpha = \infty \Rightarrow S_s = \frac{1}{\alpha} = 0$$

* The factor $\frac{\epsilon''}{\epsilon'} = \frac{\sigma}{\omega \epsilon'}$ plays an important role in determine how lossy the medium.

1) If $\frac{\epsilon''}{\epsilon'} < 10^{-2} \Rightarrow$ The medium is a low-lossy medium.

2) If $\frac{\epsilon''}{\epsilon'} > 100 \Rightarrow$ The medium is a good conductor.

3) If $10^{-2} < \frac{\epsilon''}{\epsilon'} < 10^2 \rightarrow$ The medium have quasi-conductor.

Low lossy direction

$$\gamma^2 = -\omega^2 \mu \epsilon_c \Rightarrow \gamma = j\omega \sqrt{\mu \epsilon_c} \Rightarrow \epsilon_c = \epsilon_r \epsilon'$$

$$\gamma = j\omega \sqrt{\mu \epsilon'} (1 - j \frac{\epsilon''}{\epsilon'})^{1/2}$$

$$\frac{\gamma}{\omega \sqrt{\epsilon'}} = \frac{\sigma}{2\omega \epsilon'}$$

$$(1 - x)^{1/2} \approx 1 - \frac{x}{2}$$

$$\gamma = j\omega \sqrt{\mu \epsilon'} \frac{E''}{2E'} + j\omega \sqrt{\mu \epsilon'}$$

$$= \frac{\omega \epsilon''}{2} \sqrt{\frac{\mu}{\epsilon'}} + j\omega \sqrt{\mu \epsilon'} = \alpha + j\beta$$

$$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon'}}$$

$$\beta = \omega \sqrt{\mu \epsilon'}$$

$$\eta_c = \sqrt{\frac{\mu}{\epsilon'} \left(\frac{1 - j\epsilon''}{\epsilon'} \right)}$$

$$\eta_c = \sqrt{\frac{\mu}{\epsilon_0}} \quad \text{at low-loss medium}$$

* Good conductor $\left(\frac{\epsilon''}{\epsilon'} > 100 \right) = \frac{\sigma}{\omega \epsilon'} > 100$

$$\alpha = \beta = \sqrt{\pi f \mu \sigma}$$

الاثبات $\Rightarrow \alpha = \omega \sqrt{\frac{\mu \epsilon''}{2}} = \omega \sqrt{\frac{\mu \sigma}{2\omega}} = \sqrt{\frac{\mu \sigma \omega}{2}}$

$$= \sqrt{\frac{\mu \sigma \omega}{2}} = \sqrt{\pi f \mu \sigma} = \beta$$

$$\eta_c = \sqrt{\frac{j\mu}{\epsilon''}} = (1+j) \sqrt{\frac{\pi f \mu}{\sigma}} = (1+j) \frac{\alpha}{\sigma}$$

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example :- plane wave in the +ve Z direction, Sea surface $\Rightarrow Z=0$

For sea water $\Rightarrow \epsilon_r = 80$; $\mu_r = 1$; $\sigma = 4$. If the magnetic field at $Z=0$, is given by :- $\vec{H}(0,t) = \hat{y} 100 \cos(2\pi \times 10^3 t + 15^\circ)$ mA/m

[a] Find $\vec{E}(z,t)$, $\vec{H}(z,t)$.

[b] Find the depth at which amplitude of \vec{E} is 1% of the value at $Z=0$.

$$\vec{E}(z) = -\eta \hat{k} \times \vec{H}$$

$$= \hat{x} E_{x0} e^{-\alpha z} e^{-j\beta z}$$

$$0.01 < \frac{E''}{E} < 100$$

$$\vec{H}(z) = \frac{1}{\eta_c} \hat{z} \times \hat{x} E_{x0} e^{-\alpha z} e^{-j\beta z} = \hat{y} \frac{E_{x0}}{\eta_c} e^{-\alpha z} e^{-j\beta z}$$

$$\frac{\sigma}{\omega \epsilon} = \frac{4}{2\pi \times 10^3 \times \epsilon_0 \times 80} = 9 \times 10^5 \gg 100$$

Good conductor

$$\alpha = \sqrt{\pi f \mu \sigma} = \beta \Rightarrow \sqrt{\pi \times 10^3 \times 4\pi \times 10^{-7} \times 4}$$

$$\alpha = 0.126 = \beta$$

$$\eta_c = \frac{(1+j)\alpha}{\sigma} = \frac{(1+j)0.126}{4} = \frac{\sqrt{2}}{4} e^{+j\frac{\pi}{4}} \times 0.126$$

$$\eta_c = 0.044 e^{+j\frac{\pi}{4}}$$

$$E_{x0} = |E_{x0}| e^{j\phi_0}$$

$$\vec{E}(z,t) = \text{Re} [e^{j\omega t} \hat{x} |E_{x0}| e^{j\phi_0} e^{-\alpha z} e^{-j\beta z}]$$

$$= \hat{x} |E_{x0}| e^{-\alpha z} \cos(\omega t - \beta z + \phi_0)$$

$$= \hat{x} |E_{x0}| e^{-0.126z} \cos(2\pi \times 10^3 t - 0.126z + \phi_0)$$

$$\vec{H}(z) = \hat{y} \frac{|E_{x0}|}{\eta_c} e^{-\alpha z} e^{-j\beta z} e^{j\phi_0}$$

$$= \hat{y} |E_{x0}| 22.5 e^{-0.126z} \cos(2\pi \times 10^3 t - 0.126z - 45 + \phi_0)$$

$$\vec{H}(0, t) = \hat{y} 100 \cos(2\pi \times 10^3 t + 15)$$

$$\vec{H}(0, t) = \hat{y} 22.5 |E_{x0}| \cos(2\pi \times 10^3 t - 45 + \phi_0)$$

$$15^\circ = -45 + \phi_0 \Rightarrow \boxed{\phi_0 = 60^\circ}$$

$$E_{x0} = \frac{100 \times 10^{-3}}{22.5} = 4.44 \times 10^{-3}$$

$$\vec{E}(z, t) = \hat{x} 4.44 e^{-0.126z} \cos(2\pi \times 10^3 t - 0.126z + 60^\circ) \text{ mV/m}$$

$$\vec{H}(z, t) = \hat{y} 100 e^{-0.126z} \cos(2\pi \times 10^3 t - 0.126z + 15^\circ) \text{ mV/m}$$

$$\boxed{b} \quad 0.01 = \frac{e^{-\alpha z}}{e^{-\alpha y}}, \quad e^{-\alpha z} = 0.01$$

$$-\alpha z = \ln 0.1 \rightarrow z = \frac{\ln 0.1}{-\alpha} = \frac{\ln 0.1}{-0.126} \Rightarrow \boxed{z = 36 \text{ m}}$$

Example:- Copper parameters, $\mu = \mu_0 = 4\pi \times 10^{-7}$, $\epsilon = \epsilon_0 = 10^9$
 $\sigma = 5.8 \times 10^7$, over what frequency range copper is a good conductor ^{36π}

$$\frac{E''}{E''} = \frac{\sigma}{\omega \epsilon} \gg 100 \Rightarrow \frac{\omega \epsilon}{\sigma} < 0.01 \Rightarrow f < \frac{0.01 \sigma}{2\pi \epsilon}$$

$$f < \frac{0.01 \times 5.8 \times 10^7}{2\pi \times \frac{10^9}{36\pi}} = f < \frac{5.8 \times 10^7}{\frac{200 \times 10^9}{36}} = 1.044 \times 10^{16} \text{ Hz}$$

Example:- wave is travelling in a medium with skin depth (δ_s). find $\frac{E[3\delta_s]}{E[0]}$

$$\frac{E[3\delta_s]}{E[0]} = \frac{E_{x_0} e^{-\alpha 3\delta_s}}{E_{x_0} e^{-0}} = e^{-\alpha 3 \frac{1}{\alpha}} = e^{-3} = 5\%$$

example:- In a medium with $\epsilon_r = 9$, $\mu_r = 1$, $\sigma = 0.1$. Find the phase angle by which \vec{H} leads \vec{E} at 100 MHz

$$\vec{H} = \frac{1}{n_c} \hat{k} \times \vec{E}$$

$$\frac{\sigma}{\omega \epsilon} = \frac{0.1}{2\pi \times 10^8 \times 9 \epsilon_r} = 2$$

$$n_c = \sqrt{\frac{\mu}{\epsilon}} (1 - j \frac{\sigma}{\omega \epsilon})^{-1/2} \Rightarrow \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} \left(1 - j \frac{\sigma}{\omega \epsilon}\right)^{-1/2}$$

$$= \frac{120\pi}{\sqrt{9}} (1 - j 2)^{-1/2} = 84.04 / 31.72$$

$\theta_{n_c} = 31.72$; \vec{H} leads \vec{E} by $(-\theta_{n_c})$; by (-31.72)

Example:- Based on measurement at 1 MHz $\delta_s = 2m$, $n_c = 28.1 / 45^\circ$

$$28.1 / 45^\circ = (1+j) \frac{\alpha \epsilon}{\omega}$$

(a) σ (b) wave length in the medium (c) Phase velocity in the medium

45° of $\eta \rightarrow$ have good conductor

$$\text{(a) } \sigma \rightarrow n_c = (1+j) \frac{\alpha \epsilon}{\omega} \Rightarrow 28.1 / 45^\circ = (1+j) \frac{\alpha \epsilon}{\omega}$$

$$28.1 \cos 45 + j 28.1 \sin 45 = \frac{\alpha \epsilon}{\omega} + \frac{j \alpha \epsilon}{\omega}$$

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$$\alpha = \frac{1}{S_s} = \frac{1}{2} = 0.5 = \alpha$$

$$\sigma = \frac{\alpha}{28.1/\sqrt{2}} \Rightarrow \sigma = \frac{0.5}{28.1/\sqrt{2}} = \sigma = 2.52 \times 10^{-2}$$

$$B = \alpha = 0.5$$

$$\textcircled{b} \lambda = \frac{2\pi}{B} = \frac{2\pi}{0.5} = 4\pi = 12.57 \text{ m} = \lambda$$

$$\textcircled{c} v_p = \frac{\omega}{B} = \frac{2\pi \times 10^6}{0.5} = 4\pi \times 10^6 = 1.256 \times 10^7 \text{ m/sec}$$

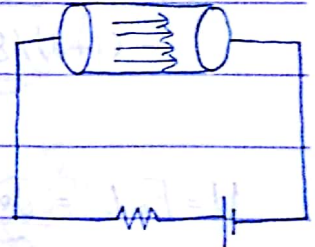
$$v_p = f\lambda$$

First

Current flow in a good conductor

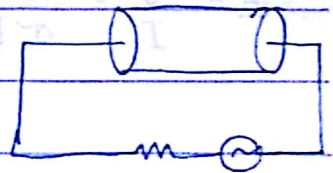
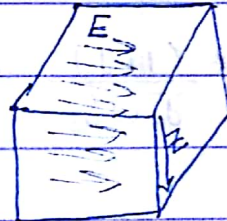
→ For DC, current has uniform density over the wire cross section.

→ For time varying current density decreases exponentially towards the axis of wire.



$$\vec{E}(0) = \hat{x} E_0$$

$$E(z) = \hat{x} E_0 e^{-\alpha z} e^{-j\beta z}$$



$$\vec{H} = \frac{1}{\eta} \hat{z} \times \hat{x} E_0 e^{-\alpha z} e^{-j\beta z} = \frac{y}{\eta_c} E_0 e^{-\alpha z} e^{-j\beta z}$$

$$\vec{J}_x(z) = \sigma E_0 e^{-\alpha z} e^{-j\beta z} \quad J_0 = \sigma E_0$$

$$\vec{J}_x(z) = J_0 e^{-\alpha z} e^{-j\beta z} \quad (\text{A/m}^2)$$

For good conductor $(\alpha = \beta = \frac{1}{\delta_s})$

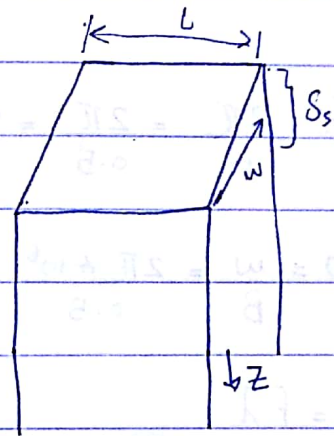
$$\vec{J}_x(z) = J_0 e^{-(\alpha+j\beta)z} = J_0 e^{-(1+j)z/\delta_s}$$

Current flow in a good conductor

$$I = \int_{z=0}^{\infty} \vec{J}_x(z) \frac{w dz}{\text{Area}}$$

$$= w \int_0^{\infty} J_0 e^{-(1+j)z/\delta_s} dz$$

$$= \frac{J_0 w e^{-(1+j)z/\delta_s}}{-(1+j)/\delta_s} \Big|_0^{\infty} = \frac{J_0 w \delta_s}{(1+j)}$$



$$V = E_0 L = \frac{J_0 L}{\sigma}$$

$$Z = \frac{V}{I} = \frac{\frac{J_0 L}{\sigma}}{\frac{J_0 w \delta_s}{(1+j)}} = \frac{(1+j)}{\sigma \delta_s} \frac{L}{w}$$

Z_s (Surface impedance)

Impedance for unit length ($l=1m$) and width ($w=1m$)

$$Z_s = \frac{1+j}{\sigma \delta_s} = R_s + j\omega L_s$$

↑ surface resistant

$$R_s = \frac{1}{\sigma \delta_s} \Rightarrow R_s = \sqrt{\frac{\pi f \mu}{\sigma}} = \sqrt{\frac{\pi f \mu \sigma}{\sigma^2}}$$

$$\omega L_s = \frac{1}{\sigma \delta_s} \Rightarrow L_s = \frac{1}{2\pi f \leftarrow w \sigma \delta_s} = \frac{1}{2} \sqrt{\frac{\mu}{\pi f \sigma}}$$

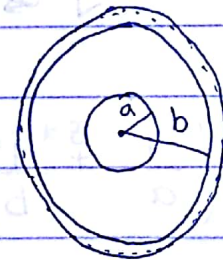
R_s

$$R = R_s \frac{L}{\text{width}} = \frac{1}{\sigma \delta_s} \frac{L}{\text{width}}$$

Coaxial cable reactance.

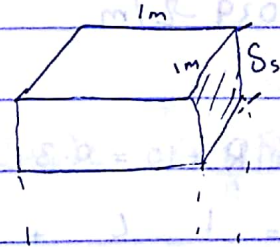
For copper $\sigma = 5.8 \times 10^{27}$, Assume $f = 1 \text{ MHz}$

$$\delta_s = \frac{1}{\sqrt{\pi f \mu \sigma}} = 0.066 \text{ mm}$$

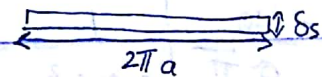


$$R' = \frac{R}{L}$$

$$R = \frac{R_s L}{\text{width}}$$

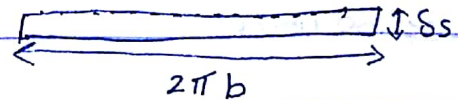


$$R' = \frac{R_s}{\text{width}}$$



$$R' = R_{in} + R_{out}, \quad R_{in} = \frac{R_s}{2\pi a}, \quad R_{out} = \frac{R_s}{2\pi b}$$

$$R'_{total} = \frac{R_s}{2\pi} \left(\frac{1}{a} + \frac{1}{b} \right) \Omega / m$$



example :- Copper coaxial cable $\sigma = 5.8 \times 10^7$, $\epsilon_r = 1$, $\mu_r = 1$,

outer thickies = 0.5mm, $a = 0.5$ cm, $b = 1$ cm, Find:-

① Surface resistance (R_s) ② Ac resistance at 10 MHz ③ R_{AC}/R_{DC}

$\frac{\sigma}{\omega \epsilon} \gg 1$ good conductor

$$\delta_s = \frac{1}{\alpha} = \frac{1}{\sqrt{\pi f \mu \sigma}} = \frac{1}{\sqrt{\pi \times 10^7 \times 4\pi \times 10^{-7} \times 5.8 \times 10^7}} = 0.021 \text{ mm}$$

$$\frac{d}{\delta_s} = \frac{0.5 \text{ mm}}{0.021} \approx 25 \text{ (very thick)}$$

$$R_s = \frac{1}{\sigma \delta_s} = \sqrt{\frac{\pi f \mu}{\sigma}} = \frac{1}{5.8 \times 10^7 \times 2.1 \times 10^{-5}} = 8.4 \times 10^{-4} \Omega$$

$$R' = \frac{R_s}{2\pi} \left(\frac{1}{a} + \frac{1}{b} \right) = \frac{8.4 \times 10^{-4}}{2\pi} \left(\frac{1}{0.5} + 1 \right) \times 10^3$$
$$= 0.039 \Omega/\text{m}$$

$$R_{AC}(10\text{m}) = R' \times 10 = 0.39 \Omega$$

$$R_{DC}(10\text{m}) = \frac{1}{\sigma} \frac{L}{A_{in}} + \frac{1}{\sigma} \frac{L}{A_{out}}$$

$$= \frac{10}{\sigma \pi} \left[\frac{1}{0.005^2} + \frac{1}{0.01^2} \right] = 7.624 \times 10^{-3}$$

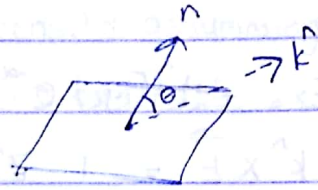
$$\frac{R_{AC}}{R_{DC}} \approx 50 \text{ times}$$



Power Density

Poynting vector \vec{S}

$$\vec{S} = \vec{E} \times \vec{H} \text{ (time domain) (W/m}^2\text{)}$$



$$P = \int_A \vec{S} \cdot \vec{n} \, dA = SA \cos \theta$$

\uparrow
surface normal

Time-average power density

$$\vec{S}_{av} = \frac{1}{2} \text{Re} [\vec{E} \times \vec{H}^*] \text{ (W/m}^2\text{)}$$

For a lossless medium

$$\vec{E}(z) = (x^{\wedge} E_{x0} + y^{\wedge} E_{y0}) e^{-jkz}$$

$$\vec{H}(z) = \frac{1}{\eta} k^{\wedge} \times \vec{E} \Rightarrow \frac{1}{\eta} z^{\wedge} \times (x^{\wedge} E_{x0} + y^{\wedge} E_{y0}) e^{-jkz}$$

$$= \frac{1}{\eta} (y^{\wedge} E_{x0} - x^{\wedge} E_{y0}) e^{-jkz}$$

$$S = \frac{1}{2} \text{Re} [\vec{E} \times \vec{H}] = \frac{1}{2} \text{Re} [(x^{\wedge} E_{x0} + y^{\wedge} E_{y0}) e^{-jkz} \times \frac{1}{\eta} (y^{\wedge} E_{x0} - x^{\wedge} E_{y0}) e^{-jkz}]$$

$$= \frac{1}{2} \text{Re} \left[\frac{z^{\wedge} E_{x0}^2}{\eta} + \frac{z^{\wedge} E_{y0}^2}{\eta} \right] \quad [E^2 = E_{x0}^2 + E_{y0}^2]$$

$$= \frac{1}{2} \text{Re} [z^{\wedge}] = z^{\wedge} \frac{E}{\eta}$$

lossy medium

$$\vec{E}(z) = (x^{\wedge} E_{x_0} + y^{\wedge} E_{y_0}) e^{-\alpha z} e^{-j\beta z}$$

$$\vec{H} = \frac{1}{\eta_c} \hat{k} \times \vec{E} = \frac{1}{\eta_c} \hat{z} \times (x^{\wedge} E_{x_0} + y^{\wedge} E_{y_0}) e^{-\alpha z} e^{-j\beta z}$$

$$= \frac{1}{\eta_c} [y^{\wedge} E_{x_0} - x^{\wedge} E_{y_0}] e^{-\alpha z} e^{-j\beta z}$$

$$\vec{S}_{av} = \frac{1}{2} \text{Re} [(x^{\wedge} E_{x_0} + y^{\wedge} E_{y_0}) e^{-\alpha z} e^{-j\beta z}] \times \frac{1}{\eta_c} [y^{\wedge} E_{x_0} - x^{\wedge} E_{y_0}] e^{-\alpha z} e^{j\beta z}$$

$$= \frac{1}{2} \text{Re} [(z^{\wedge} E_{x_0}^2 + x^{\wedge} E_{y_0}^2) e^{-2\alpha z}] \quad \eta_c = \eta_c e^{j\theta}$$

$$\vec{S}_n = \frac{z^{\wedge} |E_0|^2}{2|\eta_c|} e^{-2\alpha z} \cos(\theta_n)$$

problems:- 1, 3, 4, 6, 11, 14, 15, 16, 19, 21, 23, 28

$$\vec{S} = \vec{E} \times \vec{H}$$

$$\vec{S}_{av} = \frac{1}{2} \text{Re} [\vec{E} \times \vec{H}] = \frac{z^{\wedge} |E|^2}{2\eta} \quad (\text{W/m}^2) \quad \text{loss less}$$

$$= \frac{z^{\wedge} |E_0|^2}{2|\eta_c|} e^{-2\alpha z} \cos(\theta_n)$$

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Example :- A submarine at depth 200m uses a wire antenna to receive at 1kHz. Find the average power density assuming

$$|E_0| = 4.44 \text{ mV/m} ; \alpha = 0.126 ; \eta_c = 0.044 / 45^\circ$$

$$\vec{S}_{\text{av}}(z) = \hat{z} \frac{|E_0|^2}{2\eta_c} e^{-2\alpha z} \cos(\theta_n)$$

$$= \hat{z} \frac{(4.44 \times 10^{-3})^2}{2 \times 0.044} e^{-0.126z} \cos(\theta_n) = 2.1 \times 10^{-26} \text{ W/m}^2$$

Example :- A wave travelling in a non-magnetic medium with $\epsilon_r = 9$ has \vec{E} -

$$\text{Field } \vec{E}(z, t) = \hat{y} 3 \cos(\pi \times 10^7 t + kx) - \hat{z} 2 \cos(\pi \times 10^7 t + kx)$$

Find :-

① direction of prop $\rightarrow -\hat{x} (-x)$.

② average power carried by wave.

$$\eta = \sqrt{\frac{\mu}{\epsilon}} = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} = \frac{\eta_0}{\sqrt{\epsilon_r}} = \frac{120\pi}{3} = 40\pi$$

$$\vec{S}_{\text{av}} = \hat{k} \frac{|E_0|^2}{2\eta} = -\hat{x} \frac{(3^2 + 2^2)}{2 \times 40\pi} = -\hat{x} 0.05 \left(\frac{\text{W}}{\text{m}^2} \right)$$

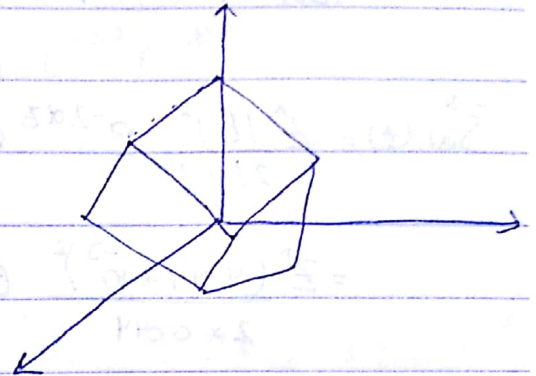
Example:- plane wave $\vec{E} = \hat{x} 100 e^{-20y} \cos(2\pi \times 10t - 40y)$

$$\vec{H} = -\hat{z} 0.64 e^{-20y} \cos(2\pi \times 10t - 40y - 36.85^\circ)$$

$$\vec{H} = \frac{1}{\eta_c} k \vec{E}$$

(a) Find the power density of the wave.

(b) Find the average power entering the box.



$$\vec{S} = \vec{E} \times \vec{H}$$

$$= \hat{x} (100 e^{-20y} \cos(\omega t - 40y)) \times (-\hat{z} 0.64 e^{-20y} \cos(\omega t - 40y - 36.85^\circ))$$

$$= \hat{y} 64 e^{-40y} \frac{1}{2} [\cos(2\omega t - 80y - 36.85^\circ) + \cos(36.85^\circ)]$$

$$\text{Power entry} = ac \left[S_{av} \Big|_{y=0} - S_{av} \Big|_{y=b} \right]$$

$$S_{av} = \frac{y \hat{E}_0^2}{2\eta_c} e^{-2\alpha y} \cos(\theta_n) = ac \times \frac{E_0^2}{2\eta_c} (1 - e^{-2\alpha b}) \cos(\theta_n)$$

$$\vec{H} = \frac{1}{\eta_c} k \vec{E} \Rightarrow -\hat{z} 0.64 e^{-20y} \frac{e^{-j(40y + 36.85^\circ)}}{\cos 2\pi t} = \frac{1}{\eta_c} \hat{y} \times \hat{x} 100 e^{-20y} e^{-j40y}$$

$$\eta_c = \frac{100}{0.64} e^{+j36.85}$$

$$ac \frac{100^2}{2 \frac{100}{0.64}} \cos(36.85^\circ) (1 - e^{-40b}) \text{ watt}$$

$$S_{av} = \frac{1}{2} \text{Re} \left[\hat{x} (100 e^{-20y} e^{-j40y}) \times \hat{z} (0.64 e^{-j36.85} e^{-20y} e^{-j40y}) \right]$$

$$= \frac{1}{2} \hat{y} 64 \cos(36.85^\circ) e^{-40y}$$

5

CH. 8 :- Wave Reflection and Transmission

* Normal incidence (lossless medium)

$\sigma = 0$

The boundary surface is ($Z=0$)

medium - 1 : $Z < 0$

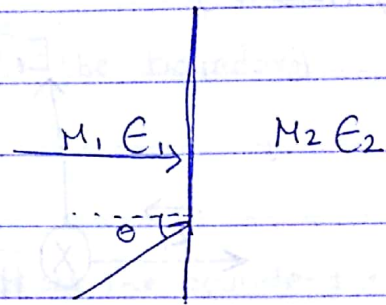
medium - 2 : $Z > 0$

In medium

we have x-polarized wave.

(\vec{E}_i, \vec{H}_i) i: incident

k_i : direction of incident wave = \hat{z}



incident wave

transmitted wave

reflected wave

medium 1

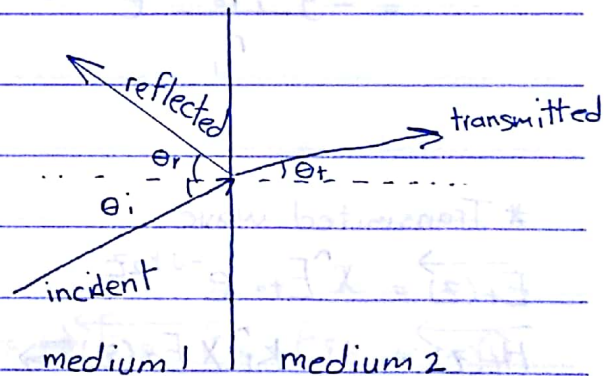
medium 2

Reflected wave

(\vec{E}_r, \vec{H}_r) $k_r = -\hat{z}$

Transmitted wave

(\vec{E}_t, \vec{H}_t) $k_t = \hat{z}$



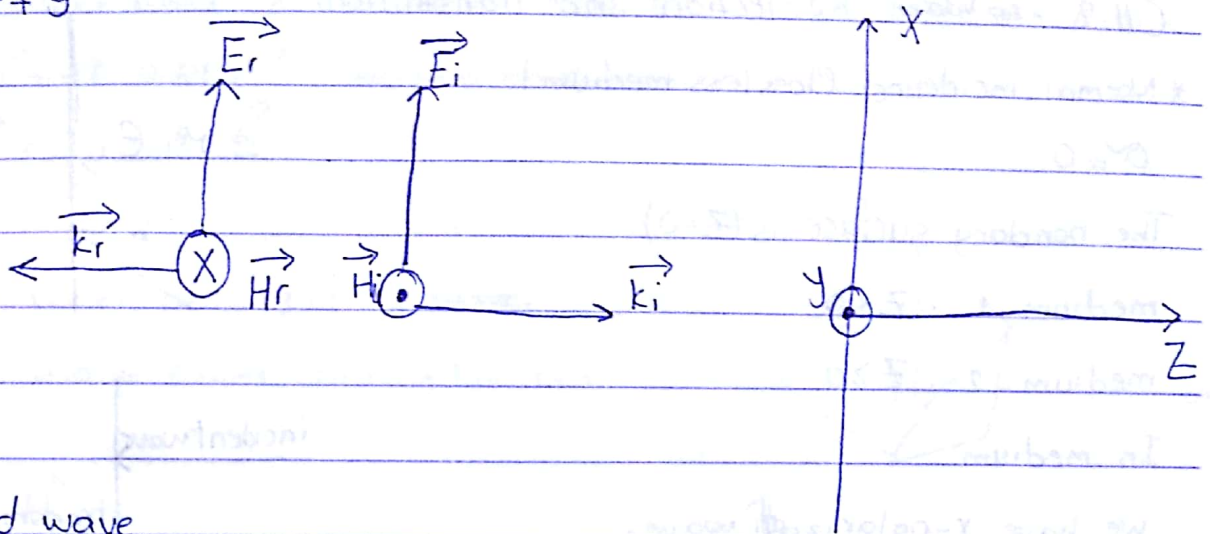
Incident wave

$\vec{E}_i(z) = \hat{x} E_{i0} e^{-jk_1 z}$

$H_i(z) = \frac{1}{\eta_1} k_i \hat{x} \times \vec{E}_i = \frac{1}{\eta} \hat{z} \times \hat{x} E_{i0} e^{-jk_1 z}$

$= \frac{1}{\eta} \hat{y} E_{i0} e^{-jk_1 z}$

⊙ H \rightarrow \hat{y}



* Reflected wave

$$\vec{E}_r(z) = \hat{x} E_{r0} e^{jk_1 z}$$

$$\vec{H}_r(z) = \frac{1}{\eta_1} \hat{k}_r \times \vec{E}_r(z) = \frac{1}{\eta_1} (-\hat{z}) \times \hat{x} E_{r0} e^{jk_1 z}$$

$$= -\hat{y} \frac{E_{r0}}{\eta_1} e^{jk_1 z}$$

μ_1	μ_2
ϵ_1	ϵ_2
$k_1 = \omega \sqrt{\mu_1 \epsilon_1}$	$k_2 = \omega \sqrt{\mu_2 \epsilon_2}$
$\eta_1 = \sqrt{\frac{\mu_1}{\epsilon_1}}$	$\eta_2 = \sqrt{\frac{\mu_2}{\epsilon_2}}$

* Transmitted wave

$$\vec{E}_t(z) = \hat{x} E_{t0} e^{-jk_2 z}$$

$$\vec{H}_t(z) = \frac{1}{\eta_2} \hat{k}_t \times \vec{E}_t(z) \Rightarrow \frac{1}{\eta_2} \hat{z} \times \hat{x} E_{t0} e^{-jk_2 z}$$

$$\Rightarrow \frac{1}{\eta_2} \hat{y} E_{t0} e^{-jk_2 z}$$

Relating E_{r0} and E_{t0} to E_{i0}

All wave have E H fields tangential to the boundary.

Boundary condition :-

Tangential \vec{E} and \vec{H} fields are continuous across the boundary.

Medium - 1

$$\vec{E}_1(z) = \vec{E}_i(z) + \vec{E}_r(z) = x^{\wedge} (E_{i0} e^{-jk_1 z} + E_{r0} e^{jk_1 z})$$

$$\vec{H}_1(z) = y^{\wedge} \left(\frac{E_{i0}}{\eta_1} e^{-jk_1 z} - \frac{E_{r0}}{\eta_1} e^{jk_1 z} \right)$$

Medium - 2

$$\vec{E}_2(z) = x^{\wedge} E_{t0} e^{-jk_2 z}$$

$$\vec{H}_2(z) = y^{\wedge} \frac{E_{t0}}{\eta_2} e^{-jk_2 z}$$

* Boundary condition $x^{\wedge} (E_{i0} - E_{r0})$

$$\text{[A]} \vec{E}_1(z=0) = \vec{E}_2(z=0) \Rightarrow x^{\wedge} (E_{i0} - E_{r0}) = E_{t0} \rightarrow E_{i0} - E_{r0} = E_{t0} \dots \text{①}$$

$$\frac{E_{i0}}{\eta_1} - \frac{E_{r0}}{\eta_1} = \frac{E_{t0}}{\eta_2}$$

$$\text{[B]} \vec{H}_1(0) = \vec{H}_2(0) \Rightarrow y^{\wedge} \left(\frac{E_{i0}}{\eta_1} - \frac{E_{r0}}{\eta_1} \right) = y^{\wedge} \frac{E_{t0}}{\eta_2} \dots \text{②}$$

$$\frac{E_{i0}}{\eta_1} - \frac{E_{r0}}{\eta_1} = \frac{E_{i0}}{\eta_2} - \frac{E_{r0}}{\eta_2} \Rightarrow \frac{E_{i0}}{\eta_1} - \frac{E_{i0}}{\eta_2} = \frac{E_{r0}}{\eta_1} + \frac{E_{r0}}{\eta_2}$$

$$E_{i0} (n_2 - n_1) = E_{r0} (n_2 + n_1)$$

$$\frac{E_{r0}}{E_{i0}} = \frac{n_2 - n_1}{n_2 + n_1} = \Gamma \text{ (reflection coefficient)}$$

$$\frac{E_{i0}}{\eta_1} - \left(\frac{E_{T0} - E_{i0}}{\eta_1} \right) = \frac{E_{T0}}{\eta_2}$$

$$E_{i0} \left(\frac{2}{\eta_1} \right) = E_{T0} \left(\frac{1}{\eta_2} + \frac{1}{\eta_1} \right)$$

$$E_{i0} \left(\frac{2}{\eta_1} \right) = E_{T0} \left(\frac{\eta_1 + \eta_2}{\eta_1 \eta_2} \right) \Rightarrow \frac{E_{T0}}{E_{i0}} = \frac{2\eta_2}{\eta_1 + \eta_2}$$

$$\Gamma \text{ (transmission coefficient)} = \frac{2\eta_2}{\eta_1 + \eta_2}$$

$$\tau = 1 + \Gamma$$

* For non magnetic media $\mu_1 = \mu_2 = \mu_0$

$$\eta = \sqrt{\frac{\mu}{\epsilon}} \quad ; \quad \eta_1 = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}} \quad , \quad \eta_2 = \sqrt{\frac{\mu_0}{\epsilon_0 \epsilon_r}}$$

$$\Gamma = \frac{\sqrt{\epsilon_r} - \sqrt{\epsilon_r}}{\sqrt{\epsilon_r} + \sqrt{\epsilon_r}} \quad \text{non-magnetic}$$

Power Flow in Lossless medium

$$\vec{S}_{av} = \frac{1}{2} \operatorname{Re} [\vec{E}_1 \times \vec{H}_1^*]$$

$$\vec{E}_1 = \hat{x} \left[E_{i0} e^{-jk_1 z} + \sqrt{M} E_{i0} e^{+jk_1 z} \right]$$

$$\vec{H}_1 = \hat{y} \left[\frac{E_{i0}}{\eta_1} e^{-jk_1 z} + \frac{M E_{i0}}{\eta_1} e^{+jk_1 z} \right]$$

$$\vec{H}_1^* = \hat{y} \left[\frac{E_{i0}^*}{\eta_1} e^{+jk_1 z} + \frac{M^* E_{i0}^*}{\eta_1} e^{-jk_1 z} \right]$$

$$\vec{S}_{av} = \frac{1}{2} \left[\hat{z} \frac{|E_{i0}|^2}{\eta_1} \left(1 - |M|^2 - M^* e^{-j2k_1 z} + M e^{j2k_1 z} \right) \right]$$

$$\Rightarrow \frac{1}{2} |E_{i0}| e^{j\theta_r} e^{j2k_1 z} - |M| e^{-j\theta_r} e^{-j2k_1 z} \quad * \frac{2j}{2j}$$

$$= \frac{1}{2} \operatorname{Re} \left[Z^{\wedge} \left[\frac{|E_{i0}|^2}{\eta_1} (1 - |M|^2 + 2j \sin(2k_1 z + \theta_r)) \right] \right]$$

$$\vec{S}_{av1} = \frac{Z^{\wedge} |E_{i0}|^2 (1 - |M|^2)}{2\eta_1}$$

$$\vec{S}_{avi} = \frac{Z^{\wedge} |E_{i0}|^2}{2\eta_1}, \quad \vec{S}_{avr} = -\frac{Z^{\wedge} |M|^2 |E_{i0}|^2}{2\eta_1} = -|M|^2 \vec{S}_{avi}$$

$$\vec{S}_{avr} = -|M|^2 \vec{S}_{avi}$$

$$\vec{S}_{av2} = \frac{1}{2} \operatorname{Re} [\vec{E}_+ \times \vec{H}_+^*] = \frac{1}{2} \operatorname{Re} \left[X^{\wedge} \tau E_{i0} e^{-jk_2 z} \times Y^{\wedge} \tau^* E_{i0}^* e^{jk_2 z} \right]$$

$$= \frac{1}{2} \operatorname{Re} \left[Z^{\wedge} |\tau|^2 |E_{i0}|^2 \right]$$

$$\vec{S}_{av2} = \frac{Z^{\wedge} |\tau|^2 |E_{i0}|^2}{2\eta_2}$$

$$\vec{S}_{av1} = \vec{S}_{av2} \quad \rightarrow \frac{|\tau|^2}{\eta_2} = \frac{1 - |M|^2}{\eta_1} \quad \text{Loss less}$$

example :- $\epsilon_{r1} = 9$, normal incidence, $\epsilon_{r2} = 4$, both media are non magnetic given $H_i(y,t) = \hat{z} 2 \cos(2\pi \times 10^9 t - ky)$ (A/m)

Find :-

a) time domain expression for E, H in each of the two media.

$$\vec{E}(y,t) = -\eta_1 \vec{k}_i \times \vec{H}_i$$

$$\eta_1 = \frac{\eta_0}{\sqrt{\epsilon_{r1}}} = \frac{120\pi}{3} = 40\pi \Omega$$

$$k_1 = \frac{\omega}{c} \sqrt{\epsilon_{r1}} = \frac{2\pi \times 10^9}{3 \times 10^8} \sqrt{9} = 20\pi$$

$$\vec{E}(y,t) = -\hat{x} 251.34 \cos(2\pi \times 10^9 t - 20\pi y)$$

$$k_2 = \frac{\omega}{c} \sqrt{\epsilon_{r2}} = \frac{40\pi}{3}$$

$$\eta_2 = \frac{\eta_0}{\sqrt{\epsilon_{r2}}} = \frac{120\pi}{2} = 60\pi \Omega$$

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{20\pi}{100\pi} = \frac{1}{5}$$

$$\tau = 1 + \Gamma \Rightarrow \tau = 1.2$$

$$\vec{E}_r(t,y) = \Gamma E_i(y,t) = -\hat{x} (50.27 \cos(2\pi \times 10^9 t + 20\pi y))$$

$$E_t(y,t) = E_i(y,t) + E_r(y,t)$$

$$= -\hat{x} (251.34 \cos(2\pi \times 10^9 t - 20\pi y) + 50.27 \cos(2\pi \times 10^9 t + 20\pi y))$$

$$\eta_2 = 60\pi$$

$$E_t(y,t) = -\hat{x} (\tau 40\pi) e^{-jk_2 y}$$

$$k_2 = 41.917$$

$$\vec{E}_t(y,t) = -\hat{x} (96\pi \cos(2\pi \times 10^9 t - k_2 y))$$

$$= -\hat{x} [301.593 \cos(2\pi \times 10^9 t - 41.92 y)]$$

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$$\begin{aligned}\vec{H}_r(y,t) &= \frac{1}{\eta_1} \hat{k}_r \times \hat{E}_r = \frac{1}{40\pi} (-\hat{y} \times -\hat{x}^{16\pi} \cos(\omega t + 20\pi y)) \\ &= \frac{16}{40\pi} (-\hat{z} \cos(\omega t + 20\pi y)) = -\hat{z} 0.4 \cos(\omega t + 20\pi y) \text{ (A/m)}\end{aligned}$$

$$\begin{aligned}\vec{H}_t(y,t) &= \frac{1}{\eta_2} \hat{k}_t \times \hat{E}_t = \frac{1}{60\pi} (\hat{y} \times -\hat{x}^{96\pi} \cos(\omega t - \frac{40\pi}{3} y)) \\ &= \hat{z} 1.6 \cos(\omega t - \frac{40\pi}{3} y) \text{ (A/m)}\end{aligned}$$

$$\boxed{2} S_{\text{av}i} = \frac{\hat{y} |E_{i0}|^2}{2\eta_1} = \frac{\hat{y} (80\pi)^2}{2 \times 40\pi} = \hat{y} 80\pi \text{ W/m}^2$$

$$\vec{S}_{\text{avr}} = -\hat{y} \Gamma^2 S_{\text{av}i} = -\hat{y} (0.04) 80\pi = -\hat{y} 10.05 \text{ W/m}^2$$

$$\vec{S}_{\text{av}t} = \frac{\hat{y} \tau^2 E_i^2}{2\eta_2} = \hat{y} 1.44 \frac{(80\pi)^2}{2 \times 60\pi} = \hat{y} 24.127 \text{ W/m}^2$$

Example:- A beam of yellow light with wave length $0.6 \mu\text{m}$ is normally incident in air upon a glass surface. Assume the glass is sufficiently thick as to ignore its back surface. If the surface is situated in the plane $z=0$ and the relative permittivity of glass is 2.25 , determine:-

- ① the location of the electric field maxima in medium 1
- ② the standing wave ratio
- ③ the fraction of incident power transmitted into the glass medium.

$$\textcircled{1} \eta_1 = \frac{\eta_0}{\sqrt{\epsilon_r}} = 120\pi \text{ } \Omega; \eta_2 = \frac{120\pi}{\sqrt{2.25}} = 251.327 = 80\pi$$

$$\Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} = \frac{-40\pi}{200\pi} = -0.2$$

$$\tau = 1 + \Gamma = 0.8$$

$$S = \frac{1 + |\Gamma|}{1 - |\Gamma|} = \frac{1.2}{0.8} = 1.5$$

$$\textcircled{3} \frac{S_{\text{avg}}}{S_{\text{avg}}} = \tau^2 \frac{E_{i0}^2}{\eta_2} = \frac{\eta_1}{\eta_2} \tau^2 \Rightarrow \frac{120\pi}{80\pi} 0.64 = 0.96 = 96\%$$

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ex:- A 1 GHz x^{\wedge} polarization plane wave in Air travelling in the (+z) direction is incident on the (x-y) plane (z=0) with $E_{i0} = 12 \text{ mV}$ on a material surface with ($\mu_r = 1, \epsilon = 1, \sigma = 5.8 \times 10^7 \text{ S/m}$). obtain time domain expression for \vec{E}, \vec{H} in the medium

$$\vec{E}_i(z) = x^{\wedge} E_{i0} e^{-jk_1 z}$$

$$c = \frac{\omega}{k} = v p^2$$

$$k_1 = \frac{\omega}{c}$$

$$k_1 = \frac{2\pi}{\lambda}$$

$$n_1 = n_0 = 120\pi = 377$$

$$\frac{\sigma}{\omega \epsilon} = \frac{5.8 \times 10^7}{2\pi \times 10^9 \times \frac{1}{36\pi} \times 10^{-9}} = 10^4 \gg 100$$

good conductor

$$n_{c2} = (1+j) \sqrt{\frac{\pi f \mu}{\sigma}} = (1+j) 8.25 \times 10^{-3} \Omega$$

$$\Gamma = \frac{n_{c2} - n_{c1}}{n_{c2} + n_{c1}} = \frac{(1+j) 8.25 \times 10^{-3} - 377}{(1+j) 8.25 \times 10^{-3} + 377} = -1$$

$$\tau = 1 + \Gamma = 0$$

$$\vec{E}_i(z) = x^{\wedge} E_{i0} \frac{(e^{-jk_1 z} - e^{jk_1 z})}{2j} = -x^{\wedge} E_{i0} (2j) \sin k_1 z$$

$$\begin{aligned} \vec{E}_i(z,t) &= -x^{\wedge} 24 \cos\left(\omega t - \frac{\pi}{2}\right) \sin k_1 z \\ &= x^{\wedge} 24 \sin\left(\frac{20\pi}{3} z\right) \sin(2\pi \times 10^9 t) \text{ mV/m} \end{aligned}$$

$$\begin{aligned}
 \vec{H}_i(z,t) &= \frac{1}{\eta} \hat{k} \times \vec{E} \\
 &= \frac{1}{\eta} \hat{z} \times x \hat{E}_{i0} e^{-jk_1 z} - \frac{1}{\eta} (-z) \times x \hat{E}_{i0} e^{jk_1 z} \\
 &= \frac{2}{\eta} y \hat{E}_{i0} e^{-jk_1 z} + \frac{y \hat{E}_{i0}}{\eta} e^{jk_1 z} \\
 &= y \hat{64} \cos\left(\frac{20\pi z}{3}\right) \cos(2\pi \times 10^4 t) \text{ (MA/m)}
 \end{aligned}$$

* if $\Gamma = -1$ the boundary between dielectric and perfect conductor.

Snell's Law

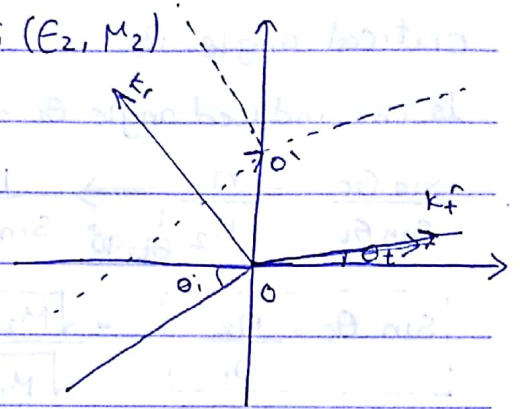
Boundary between two dielectric media (ϵ_1, μ_1) ; (ϵ_2, μ_2)

$$\frac{A_i \cdot \hat{O}}{u_{p1}} = \frac{A_{ro}}{u_{p1}} = \frac{A_t}{u_{p2}}$$

$$A_i \cdot \hat{O} = O O' \sin \theta_i$$

$$A_{ro} = O O' \sin \theta_r$$

$$A_t = O O' \sin \theta_t$$

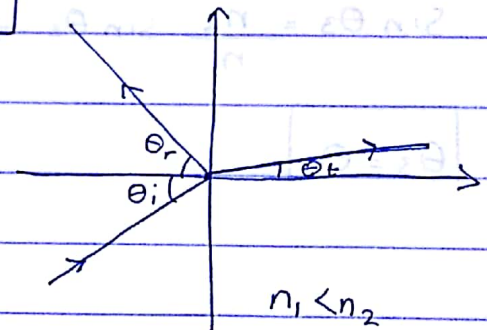


$$\frac{O O' \sin \theta_i}{u_{p1}} = \frac{O O' \sin \theta_r}{u_{p1}} \Rightarrow \sin \theta_i = \sin \theta_r$$

$$\frac{O O' \sin \theta_i}{u_{p1}} = \frac{O O' \sin \theta_t}{u_{p2}} \Rightarrow \frac{\theta_t}{\theta_i} = \frac{u_{p2}}{u_{p1}} \quad \text{Snell's Law for refraction}$$

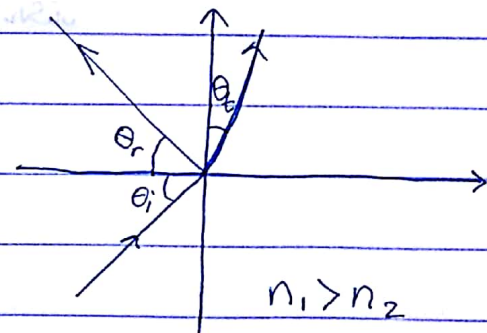
$$= \frac{\sqrt{\mu_1 \epsilon_1}}{\sqrt{\mu_2 \epsilon_2}}$$

$$\frac{\sin \theta_t}{\sin \theta_i} = \frac{n_1}{n_2}$$



Index of Refraction

$$n = \frac{c}{u_p} = \sqrt{\frac{\mu_r \mu_0 \epsilon_0 \epsilon_r}{\mu_0 \epsilon_0}} = \sqrt{\mu_r \epsilon_r}$$



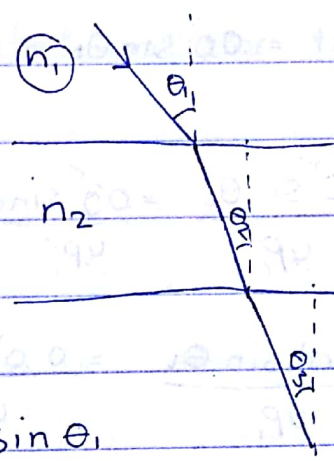
critical angle θ_c

Is the induced angle θ_i at which $\theta_t = \frac{\pi}{2}$

$$\frac{\sin \theta_t}{\sin \theta_i} = \frac{n_1}{n_2} \xrightarrow{\theta_t = 90^\circ} \frac{1}{\sin \theta_c} = \frac{n_1}{n_2}$$

$$\sin \theta_c = \frac{n_2}{n_1} = \frac{\sqrt{\mu_{2\beta} \epsilon_{2\beta}}}{\sqrt{\mu_{1\beta} \epsilon_{1\beta}}} \quad (\mu = \mu_0)$$

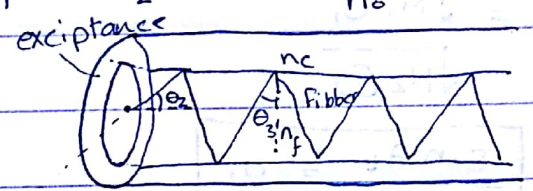
$$= \frac{\sqrt{\epsilon_{2\beta}}}{\sqrt{\epsilon_{1\beta}}} = \sqrt{\frac{\epsilon_{2\beta}}{\epsilon_{1\beta}}} \quad (\text{non-magnetic})$$



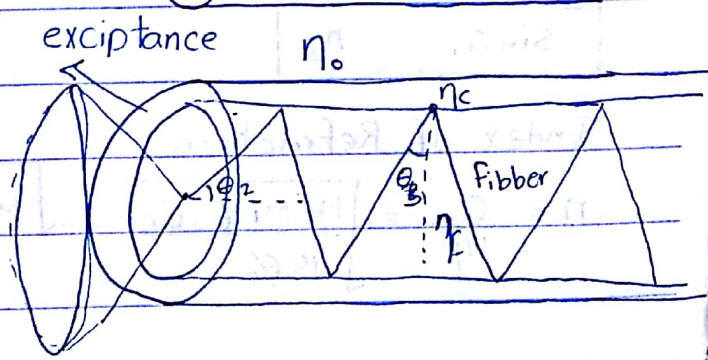
$$\frac{\sin \theta_2}{\sin \theta_1} = \frac{n_1}{n_2}, \quad \frac{\sin \theta_3}{\sin \theta_2} = \frac{n_2}{n_1}$$

$$\sin \theta_3 = \frac{n_2}{n_1} \sin \theta_2 \Rightarrow \sin \theta_3 = \frac{n_2}{n_1} * \frac{n_1}{n_2} \sin \theta_1$$

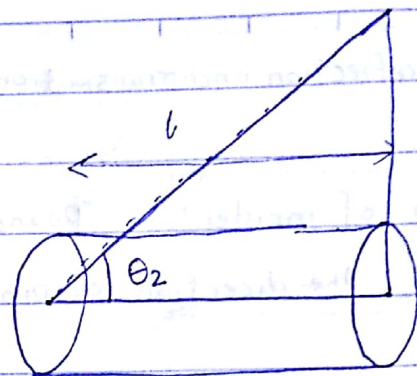
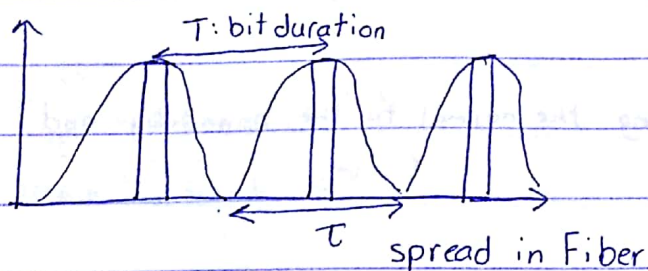
$$\theta_3 = \theta_1$$



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Max bit rate



$$T \geq 2\tau, \quad f = \frac{1}{T}$$

$$t_{\min} = \frac{L}{v_p} = \frac{L}{c/n_f} = \frac{L n_f}{c} \quad (c = n_f v_p)$$

$$L_{\max} = \frac{L}{\cos \theta_2} = L \frac{n_f}{n_c} = \frac{L n_f}{c} \left(\frac{n_t}{n_c} - 1 \right)$$

$$t_{\max} = \frac{L_{\max}}{v_p} = \frac{L n_f / n_c}{c/n_f} = \frac{L n_f^2}{c n_c}$$

$$f_{\max} = \frac{1}{2T} = \frac{c n_c}{2 L n_f (n_f - n_c)}$$

example:- 1 km opt. fiber $n_f = 1.52$, $n_c = 1.49$

$$\theta_a = \sin^{-1} \left(\frac{1}{n_o} \sqrt{n_f^2 - n_c^2} \right) = \sin^{-1} \left(\frac{1}{120\pi} \sqrt{1.52^2 - 1.49^2} \right)$$

$$= \sin^{-1} \left(\frac{1}{120\pi} * 0.3 \right)$$

$$f_p = \frac{c n_c}{2 L n_f (n_f - n_c)} = 4.9013 * 10^6 \text{ Hz b/sec}$$

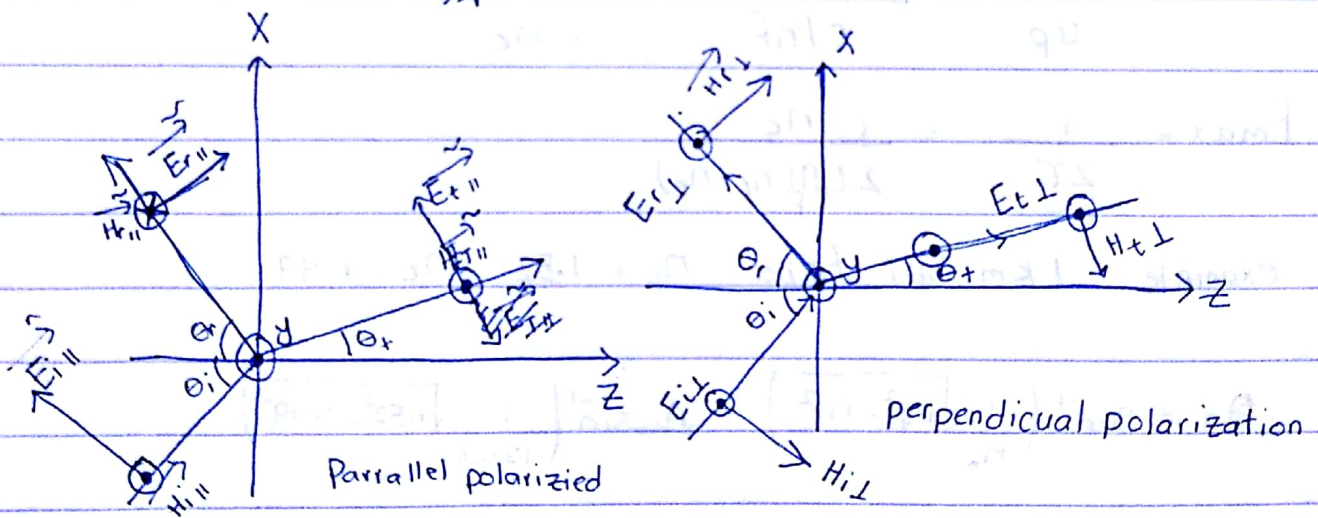
Wave reflection and transmission of oblique incident

*plane of incident :- Plane containing the normal to the boundary and the direction of propagation

Best way to solve reflection/transmission problem.

- 1) Dec the incident wave (\vec{E}_i, \vec{H}_i) into
 - a) perpendicular polarized component $(\vec{E}_{i\perp}, \vec{H}_{i\perp})$
 - b) parallel polarized component $(\vec{E}_{i\parallel}, \vec{H}_{i\parallel})$

- 2) Reflected wave $(\vec{E}_{r\perp}, \vec{H}_{r\perp})$
 ~~$(\vec{E}_{r\parallel}, \vec{H}_{r\parallel})$~~ $(\vec{E}_{r\parallel}, \vec{H}_{r\parallel})$

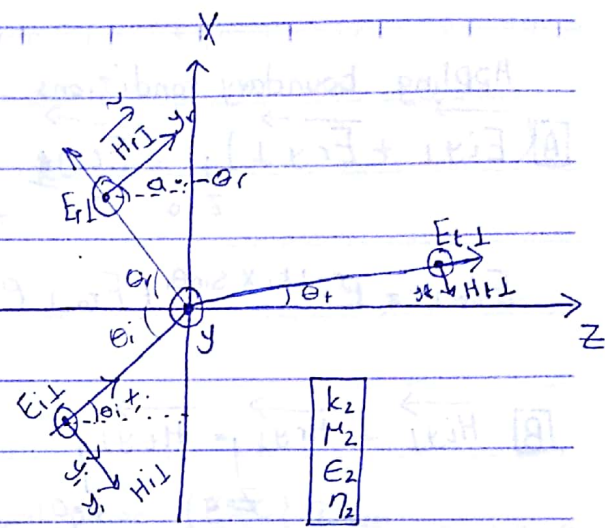


perpendicular polarization

$$\vec{E}_{i\perp} = \hat{y} E_{i0\perp} e^{-jk_1 x_i}$$

$$\vec{H}_{i\perp} = \frac{\hat{y} E_{i0\perp}}{\eta_1} e^{-jk_1 x_i}$$

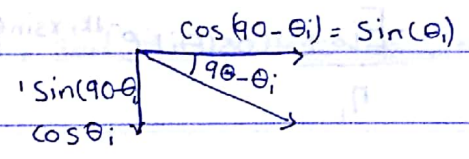
k_1
M_1
E_1
η_1



$$\Delta Z \rightarrow \Delta Z \cos \theta_i$$

$$\Delta X \rightarrow \Delta X \sin \theta_i$$

$$x_i = X \sin \theta_i + Z \cos \theta_i$$



$$\vec{E}_{i\perp} = \hat{y} E_{i0\perp} e^{-jk_1 (X \sin \theta_i + Z \cos \theta_i)}$$

$$\vec{H}_{i\perp} = (-\hat{x} \cos \theta_i + \hat{z} \sin \theta_i) \frac{E_{i0\perp}}{\eta_1} e^{-jk_1 (X \sin \theta_i + Z \cos \theta_i)}$$

$$x_i = -\hat{x} \cos \theta_i + \hat{z} \sin \theta_i$$

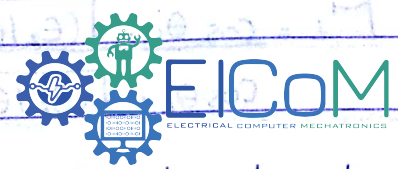
Reflected wave :- ^{الموجة المنعكسة على الواجهة}

$$\vec{E}_{r\perp} = +\hat{y} E_{r0\perp} e^{-jk_1 (X \sin \theta_r - Z \cos \theta_r)}$$

$$\vec{H}_{r\perp} = (\hat{x} \cos \theta_r + \hat{z} \sin \theta_r) \frac{E_{r0\perp}}{\eta_1} e^{-jk_1 (X \sin \theta_r - Z \cos \theta_r)}$$

$$\vec{E}_{t\perp} = \hat{y} E_{t0\perp} e^{-jk_2 (Z \cos \theta_t + X \sin \theta_t)}$$

$$\vec{H}_{t\perp} = (-\hat{x} \cos \theta_t + \hat{z} \sin \theta_t) \frac{E_{t0\perp}}{\eta_2} e^{-jk_2 (Z \cos \theta_t + X \sin \theta_t)}$$



Applying boundary conditions

$$\boxed{A} \left(\vec{E}_{iy\perp} + \vec{E}_{ro\perp} \right) \Big|_{z=0} = \vec{E}_{to\perp} \Big|_{z=0}$$

$$E_{io\perp} = E^{-jk_1 x \sin\theta_i} + E_{ro\perp} e^{-jk_1 x \sin\theta_r} = E_{to\perp} e^{-jk_2 x \sin\theta_t}$$

$$\boxed{B} \left(\vec{H}_{iy\perp} + \vec{H}_{ry\perp} \right) \Big|_{z=0} = \vec{H}_{ty\perp} \Big|_{z=0}$$

$$\frac{-E_{io\perp}}{\eta_1} \cos(\theta_i) e^{-jk_1 x \sin\theta_i} + \frac{E_{ro\perp}}{\eta_1} \cos(\theta_r) e^{-jk_1 x \sin\theta_r} = \frac{-E_{to\perp}}{\eta_2} \cos(\theta_t) e^{-jk_2 x \sin\theta_t}$$

$$1) E_{io\perp} + E_{ro\perp} = E_{to\perp}$$

$$2) \frac{\cos(\theta_i)}{\eta_1} [-E_{io\perp} + E_{ro\perp}] = -\frac{\cos(\theta_t)}{\eta_2} E_{to\perp}$$

$$\Gamma_{\perp} = \frac{E_{ro\perp}}{E_{io\perp}} = \frac{\eta_2 \cos\theta_i - \eta_1 \cos\theta_t}{\eta_2 \cos\theta_i + \eta_1 \cos\theta_t}$$

$$\boxed{\mathcal{T}_{\perp} = 1 + \Gamma_{\perp}}$$

$$\mathcal{T}_{\perp} = \frac{E_{to\perp}}{E_{io\perp}} = \frac{2\eta_2 \cos\theta_i}{\eta_2 \cos\theta_i + \eta_1 \cos\theta_t}$$

\therefore if medium 2 is perfect conductor $\sigma = \infty$

$$\rightarrow \eta_2 = 0$$

$$\Gamma_{\perp} = -1, \quad \mathcal{T}_{\perp} = 0$$

\therefore For non-magnetic conductor ($\mu_1 = \mu_2 = \mu_0$)

$$\Gamma_{\perp} = \frac{\cos\theta_i - \sqrt{(\epsilon_2/\epsilon_1) - \sin^2\theta_i}}{\cos\theta_i + \sqrt{(\epsilon_2/\epsilon_1) - \sin^2\theta_i}}$$

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$$\Gamma_{\perp} = \frac{\cos \theta_i - \sqrt{(E_2/E_1) - \sin^2 \theta_i}}{\cos \theta_i + \sqrt{(E_2/E_1) - \sin^2 \theta_i}}$$

we have used $\frac{E_2}{E_1} = \left(\frac{\eta_2}{\eta_1}\right)^2$ "SNLS Low"

Example:- plane wave air is incident on soil plane ($Z=0$) Given:-

$E_{i\perp} = \hat{y} 100 \cos(\omega t - \pi x - 1.73\pi z)$ V/m, soil is loss less, with $(\epsilon_r = 4)$, non magnetic. Find:-

- a) Find k_1, k_2, θ_i b) Find all fields (E, H) in both medium

X-Z plane is the incident plane prop $(+\hat{z})(+\hat{x})$ directions,

$$\vec{E}_{i\perp} = \hat{y} 100 e^{-j\pi x - j1.73\pi z} = \hat{y} 100 e$$

$$k_1 (z \cos \theta_i + x \sin \theta_i) = k_1 z \cos \theta_i + k_1 x \sin \theta_i$$

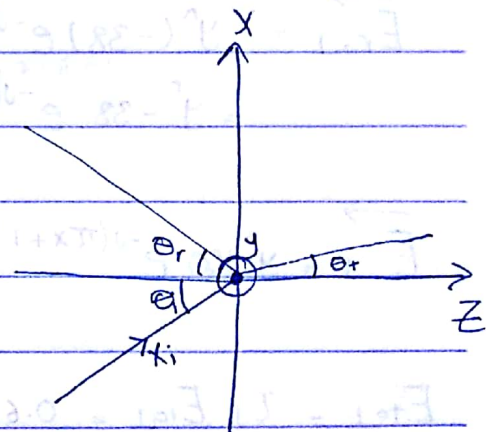
$$k_1 z \cos \theta_i + k_1 x \sin \theta_i = \pi x + 1.73\pi z$$

$$k_1 \sin \theta_i = \pi, \quad k_1 \cos \theta_i = 1.73\pi$$

$$k_1^2 \sin^2 \theta_i + k_1^2 \cos^2 \theta_i = \pi^2 + 1.73^2 \pi^2$$

$$k_1^2 (\sin^2 \theta_i + \cos^2 \theta_i) = \pi^2 (1 + 1.73^2)$$

k_1



$$\frac{k_1 \sin \theta_i}{k_1 \cos \theta_i} = \frac{\pi}{1.73\pi} \Rightarrow \tan \theta_i = \frac{1}{1.73} \Rightarrow \boxed{\theta_i = 30^\circ}$$

$$\frac{\sin \theta_t}{\sin \theta_i} = \frac{k_1}{k_2} \Rightarrow \sin \theta_t = \sin \theta_i \frac{2\pi}{4\pi} = 0.25$$

$$\theta_t = \sin^{-1}(0.25) = 14.5^\circ$$

$$\eta_2 = \frac{\eta_0}{\sqrt{\epsilon_r}} = \frac{120\pi}{\sqrt{4}} = 60\pi \quad \eta_1 = 120\pi = 37.7$$

$$E_{r0L} = \sqrt{\epsilon_r} E_{i0L} = -0.38 * 100 = -38$$

$$\begin{aligned} \vec{E}_{r0L} &= \hat{y}(-38) e^{-jx r k_1} = \hat{y}(-38) e^{-jk_1(-z \cos \theta_r + x \sin \theta_r)} \\ &= \hat{y}(-38) e^{-j(\pi x - 1.73\pi z)} \end{aligned}$$

$$\vec{E}_L = \hat{y} 100 e^{-j(\pi x + 1.73\pi z)} - \hat{y} 38 e^{-j(\pi x + 1.73\pi z)}$$

$$E_{t0L} = T_L E_{i0L} = 0.62 * 100 = 62$$

$$\begin{aligned} \vec{E}_{iL} = \vec{E}_{tL} &= \hat{y} 62 e^{-jk_2 x t} \\ &= \hat{y} 62 e^{-jz\pi(z \cos 14.5 + x \sin 14.5)} = \hat{y} 62 e^{-j(3.87\pi z + \pi x)} \end{aligned}$$

$$E_{iL} = \hat{y} [100 \cos(\omega t - \pi x - 1.73\pi z) - 38 \cos(\omega t - \pi x + 1.73\pi z)]$$

$$E_{2L} = \hat{y} 62 \cos(\omega t - \pi x - 3.87\pi z)$$

$$H_{i\perp} = \frac{1}{\eta_1} \cos \theta_i k_i \times \vec{E}_{i\perp}$$

$$\hat{k}_i = \cos \theta_i \hat{z} + \sin \theta_i \hat{x} = \hat{z} 0.866 + \hat{x} 0.5$$

$$\vec{H}_{i\perp} = \frac{1}{\eta_1} (\hat{z} 0.866 + \hat{x} 0.5) \times (\hat{y} 100 e^{-j(\pi x + 1.73\pi z)})$$

$$= \frac{1}{377} (-\hat{x} 86.6 + \hat{z} 50) e^{-j(\pi x + 1.73\pi z)}$$

$$= (-\hat{x} 0.2297 + \hat{z} 0.1326) \cos(\omega t - \pi x - 1.73\pi z)$$

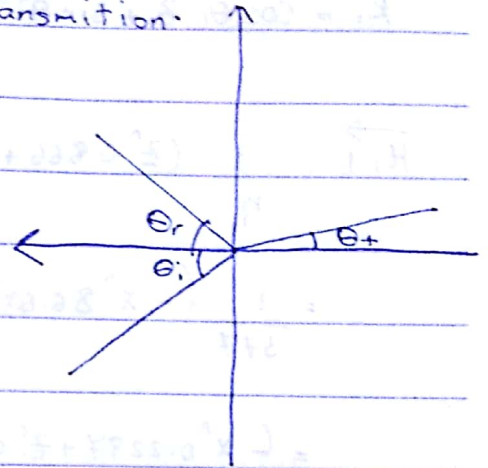
$$\hat{k}_r = \hat{z} (-\cos \theta_r) + \hat{x} (\sin \theta_r) = -\hat{z} 0.866 + \hat{x} 0.5$$

$$\vec{H}_{r\perp} = \frac{1}{377} (-\hat{z} 0.866 + \hat{x} 0.5) \times (-\hat{y} 38 e^{-j(\pi x - 1.73\pi z)})$$

Example:- Uniform plane wave in air having $E_i = -y\hat{y} \cos(\omega t - 4x - 3z)$ is incident on a dielectric slab $z \geq 0$ with $(\mu_r = 1, \epsilon_r = 2.5, \sigma = 0)$

Find :-

- ① angle of incident. ② angle of transmission.
- ③ reflected of transmitted \vec{E} -fields.
- ④ " of transmitted \vec{H} -fields.



Sol
 $\vec{E}_i = y\hat{y} E_{i0} e^{-jk_1(x \sin \theta_i + z \cos \theta_i)}$

$E_{i0} = -8$

$k_1 \sin \theta_i = 4 \Rightarrow \tan \theta_i = \frac{4}{3} \Rightarrow \theta_i = 53.13$

$k_1 \cos \theta_i = 3 \Rightarrow k_1 = 5$

$\frac{k_1}{k_2} = \frac{\sqrt{\mu_0 \epsilon_0}}{\sqrt{\mu_0 \epsilon_0 \epsilon_r}} = \frac{1}{\sqrt{2.5}} \Rightarrow k_2 = 5\sqrt{2.5}$

$\eta_2 = 75.895\pi$

$\eta_1 = 120\pi$

$k_2 = 7.9$

$\frac{\sin \theta_t}{\sin \theta_i} = \frac{5}{7.9} \Rightarrow \sin \theta_t = \frac{0.51}{7.9} \Rightarrow \theta_t = 30.395^\circ$

$\Gamma = \frac{\eta_2 \cos \theta_i - \eta_1 \cos \theta_t}{\eta_2 \cos \theta_i + \eta_1 \cos \theta_t} = \frac{45.537\pi - 103.51\pi}{45.537\pi + 103.51\pi}$

$= -0.38896$

$\tau = 0.611$

$$E_{o1} = (-0.389)(-8) = 3.112$$

$$\begin{aligned} \vec{E}_{r1} &= \hat{y} 3.112 e^{-jk_1(x \sin \theta_r - z \cos \theta_r)} = \hat{y} 3.112 e^{-j\sqrt{2}(-z + 4x)} \\ &= \hat{y} 3.112 \cos(\omega t + 3z - 4x) \end{aligned}$$

$$\vec{E}_{t1} = \hat{y} -4.888 e^{-jk_2(+x \cos \theta_t + z \sin \theta_t)}$$

$$\theta_i = 53.13^\circ = \theta_r$$

$$= \hat{y} -4.888 e^{-j(+6.82x + 4z)}$$

$$= \hat{y} -4.888 \cos(\omega t - 4z - 6.82x)$$

$$\theta_t = 30.39^\circ$$

$$\hat{k}_i = \hat{z} \sin \theta_i - \hat{x} \cos \theta_i = 0.8\hat{z} - 0.6\hat{x}$$

$$\hat{k}_r = \hat{z} \sin \theta_r + \hat{x} \cos \theta_r = 0.8\hat{z} + 0.6\hat{x}$$

$$\hat{k}_t = -\hat{x} \cos \theta_t + \hat{z} \sin \theta_t = 0.51\hat{z} - 0.863\hat{x}$$

$$\vec{H}_{i1} = \frac{1}{\eta_1} \hat{k}_i \times \vec{E}_{i1} = \frac{1}{120\pi} (0.8\hat{z} - 0.6\hat{x}) \times \hat{y} 3.112 \cos(\omega t - 4x - 3z)$$

$$= (0.8\hat{x} + 0.6\hat{z}) \frac{3.112}{120\pi} \cos(\omega t - 4x - 3z)$$

$$\vec{H}_{r1} = \frac{1}{\eta_1} \hat{k}_r \times \vec{E}_{r1} = \frac{1}{120\pi} (0.8\hat{z} + 0.6\hat{x}) \times \hat{y} 3.112 \cos(\omega t + 3z - 4x)$$

$$= (-0.8\hat{x} + 0.6\hat{z}) \frac{3.112}{120\pi} \cos(\omega t + 3z - 4x)$$

$$= (-6.6\hat{x} + 4.953\hat{z}) \cos(\omega t + 3z - 4x) \text{ m}$$

$$\vec{H}_{t1} = \frac{1}{75.895} (0.51\hat{z} - 0.863\hat{x}) \times \hat{y} (-4.888 \cos(\omega t - 4x - 6.82z))$$

$$= \frac{1}{75.895} (0.51\hat{x} + 0.863\hat{z}) 68.4 \cos(\omega t - 4x - 6.82z) \text{ m}$$

$$x_i^{\wedge} = x \sin \theta_i + z \cos \theta_i$$

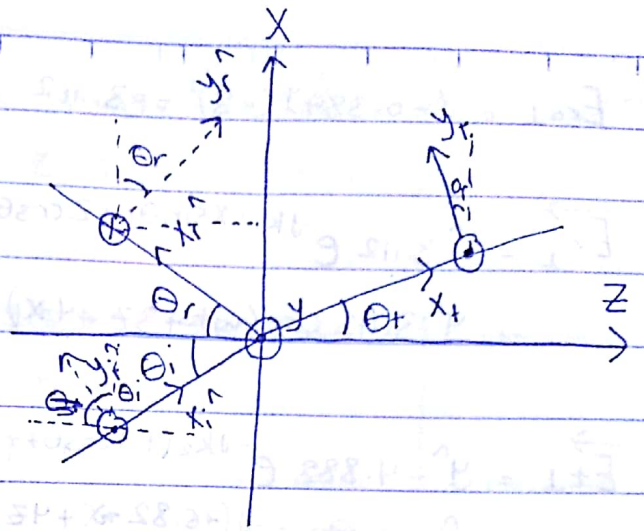
$$y_i^{\wedge} = x^{\wedge} \cos \theta_i - z^{\wedge} \sin \theta_i$$

$$x_r^{\wedge} = x \sin \theta_r - z \cos \theta_r$$

$$y_r^{\wedge} = x^{\wedge} \cos \theta_r - z^{\wedge} \sin \theta_r$$

$$x_t^{\wedge} = x \sin \theta_t + z \cos \theta_t$$

$$y_t^{\wedge} = x^{\wedge} \cos \theta_t - z^{\wedge} \sin \theta_t$$



$$\Gamma_{11} = \frac{E_{r011}}{E_{i011}} = \frac{\eta_2 \cos \theta_t - \eta_1 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i}$$

$$T_{11} = \frac{E_{t011}}{E_{i011}} = \frac{2 \eta_2 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i} \quad T_{11} = (1 + \Gamma_{11}) \frac{\cos \theta_i}{\cos \theta_t}$$

if the second medium is perfect conductor

$$\eta_c = \sqrt{\frac{\mu}{\epsilon (1 + j \frac{\sigma}{\omega \epsilon})}} \quad \sigma = \infty \Rightarrow \eta_c = 0 \Rightarrow \Gamma_{11} = -1 \Rightarrow T_{11} = 0$$

Example: A plane wave in air with $\vec{E}_i = (-10\hat{y} + 5\hat{z})\cos(\omega t - 2y - 4z)$
 ($\sigma = 0, \mu = \mu_0, \epsilon_r = 4$)

- ① θ_i, θ_r and θ_t ③ total \vec{E} and \vec{H} fields
 ② $\Gamma_{||}$ and $\tau_{||}$ ④ Average incident power vector

① $k_1 \sin \theta_i = 2$ $\tan \theta_i = 2 \Rightarrow \theta_i = 26.565^\circ$ $k_1 = 4.472$

$k_1 \cos \theta_i = 4$

$\frac{k_2}{k_1} = \sqrt{\frac{\mu r_2}{\mu r_1}} \Rightarrow k_2 = 2k_1 \Rightarrow k_2 = 8.9443$

$\frac{\sin \theta_t}{\sin \theta_i} = \frac{k_1}{k_2} \Rightarrow \theta_t = 12.921^\circ$ $\theta_r = 12.921^\circ$ $\eta_1 = 120\pi$ $\eta_2 = 60\pi$

$\Gamma_{||} = \frac{58.481 - 107.33}{58.481 + 107.33} = -0.295$ $\tau_{||} = 0.705$

~~Er =~~

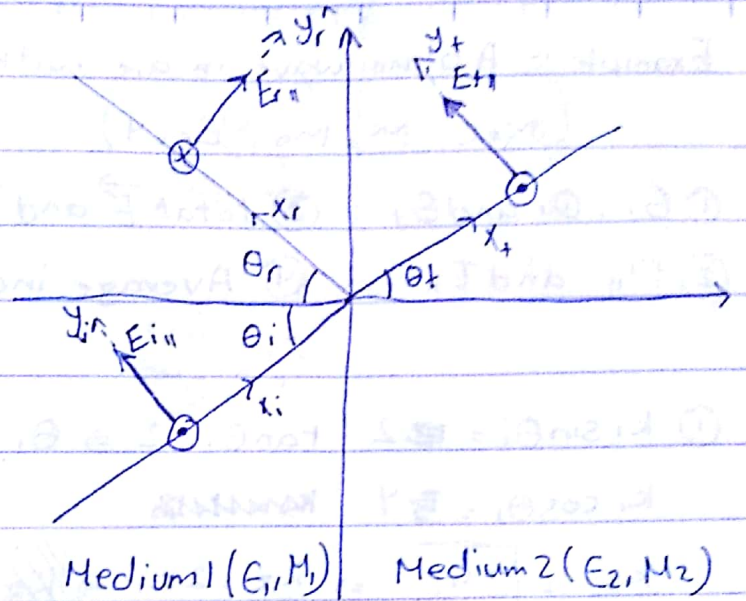
$E_r = (2.95\hat{y} + 1.475\hat{z})e^{-j(2y-4z)}$

$E_t = (-7.05\hat{y} + 3.525\hat{z})e^{-j(2y+4z)}$

$\vec{H}_i = \frac{1}{120\pi} (\hat{y} 0.89 - \hat{z} 0.447) \times (-10\hat{y} + 5\hat{z}) e^{-j(2y+4z)}$

$= \frac{1}{120\pi} (-\hat{x}) e^{-j(2y+4z)}$

Parallel polarization



$$x_i = x \sin \theta_i + z \cos \theta_i$$

$$x_r = x \sin \theta_r - z \cos \theta_r$$

$$x_t = x \sin \theta_t + z \cos \theta_t$$

$$y_i = x \cos \theta_i - z \sin \theta_i$$

$$y_r = x \cos \theta_r + z \sin \theta_r$$

$$y_t = x \cos \theta_t - z \sin \theta_t$$

$$\Gamma_{\parallel} = \frac{E_{r\parallel}}{E_{i\parallel}} = \frac{\eta_2 \cos \theta_t - \eta_1 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i} \Rightarrow \text{if } \eta_2 = 0 \Rightarrow \boxed{\Gamma_{\parallel} = -1}$$

$$T_{\parallel} = \frac{E_{t\parallel}}{E_{i\parallel}} = \frac{2\eta_2 \cos \theta_i}{\eta_2 \cos \theta_t + \eta_1 \cos \theta_i}$$

$$T_{\parallel} = (1 + \Gamma_{\parallel}) \frac{\cos \theta_i}{\cos \theta_t}$$

$$\text{For non magnetic :- } \Gamma_{\parallel} = -\frac{(E_2/E_1) \cos \theta_i + \sqrt{(E_2/E_1) - \sin^2 \theta_i}}{(E_2/E_1) \cos \theta_t + \sqrt{(E_2/E_1) - \sin^2 \theta_i}} \quad (M_1 = M_2)$$

The vector of average incident power density

$$\vec{S}_{\text{avi}} = \frac{|E_i|^2}{2\eta_1} \hat{k}_i = \frac{10^2 + 5^2}{2 \times 377} (\hat{z} \cdot 0.894 + \hat{y} \cdot 0.447)$$

$$= \hat{y} \cdot 0.074 + \hat{z} \cdot 0.118 \text{ (W/m}^2\text{)}$$

or

$$\vec{S}_{\text{avi}} = \frac{1}{2} \text{Re}(\vec{E} \times \vec{H}) = \frac{1}{2} \text{Re}((-10\hat{y} + 5\hat{z})e^{-j(2y+4z)} \times (\hat{x} \cdot 24.638)e^{j(2y+4z)})$$

Brewster angle (θ_B)

Is the incident angle at which ($\Gamma = 0$)

1) perpendicular polarization

$$\Gamma_{\perp} = \frac{\eta_2 \cos \theta_i - \eta_1 \cos \theta_t}{\eta_2 \cos \theta_i + \eta_1 \cos \theta_t} = 0 \Rightarrow \eta_2^2 \cos^2 \theta_i = \eta_1^2 \cos^2 \theta_t$$

$$\eta_2 \cos \theta_i + \eta_1 \cos \theta_t$$

$$\Rightarrow \eta_2^2 (1 - \sin^2 \theta_i) = \eta_1^2 (1 - \sin^2 \theta_t) \Rightarrow \left. \begin{array}{l} \sin \theta_t = \frac{v_2}{v_1} = \sqrt{\frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2}} \\ \frac{\sin \theta_t}{\sin \theta_i} = \frac{v_2}{v_1} \end{array} \right\}$$

$$\sin^2 \theta_t = \frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \sin^2 \theta_i$$

$$\sin^2 \theta_i \left(-\eta_2^2 + \eta_1^2 \frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \right) = \eta_1^2 - \eta_2^2$$

$$\sin \theta_{B1} \left(-\frac{\mu_2}{\epsilon_2} + \frac{\mu_1}{\epsilon_1} \frac{\mu_1 \epsilon_1}{\mu_2 \epsilon_2} \right) = \frac{\mu_1}{\epsilon_1} \frac{\mu_2}{\epsilon_2}$$

$$\sin \theta_{B1} = \frac{\sqrt{1 - \frac{\mu_1 \epsilon_2}{\mu_2 \epsilon_1}}}{\sqrt{1 - \left(\frac{\mu_1}{\mu_2}\right)^2}}$$

if $\mu_1 = \mu_2 \Rightarrow \sin \theta_i = \infty$

θ_{BL} dose not exist for $\mu_1 = \mu_2$ or for non magnetic

2) Parrallel polarization

$$\Gamma_{11} = 0$$

$$\sin \theta_{B11} = \sqrt{\frac{1 - \frac{\epsilon_1 \mu_2}{\epsilon_2 \mu_1}}{1 - (\epsilon_1 / \epsilon_2)^2}}$$

For non magnetic ($\mu_1 = \mu_2 = \mu_0$)

$$\theta_{B11} = \sin^{-1} \left(\sqrt{\frac{1}{1 + (\epsilon_1 / \epsilon_2)^2}} \right) = \tan^{-1} \left(\sqrt{\frac{\epsilon_2}{\epsilon_1}} \right)$$

$\neq \theta_B$ is called pollarizing angle

if a wave with perpendicular, parrallel polarization is incident on non-magnetic material at $\theta_i = \theta_{B11}$ so that the \parallel -component ($\Gamma_{11} = 0$) is totally transmitted only the \perp component is reflected.

Reflected and transmittivities

$$R_{\perp} (\text{reflected}) = \frac{P_{r\perp}}{P_{i\perp}}$$

$$P_{i\perp} = S_{i\perp} A_i = \frac{|E_{i\perp}|^2}{2\eta_1} A \cos \theta_i$$

$$P_{r\perp} = \frac{|E_{r\perp}|^2}{2\eta_1} A \cos \theta_r$$

$$R_{+\perp} = \frac{|E_{+\perp}|^2}{2\eta_2} A \cos \theta_+$$

$$R_{\perp} = \frac{|E_{r\perp}|^2}{|E_{i\perp}|^2} = |\Gamma_{\perp}|^2$$

$$R_{\parallel} = \frac{P_{r\parallel}}{P_{i\parallel}} = |\Gamma_{\parallel}|^2$$

$$T_{(\text{transmittivity})} = \frac{P_t}{P_i}$$

$$T_{\perp} = \frac{(|E_{t\perp}|^2 / 2\eta_2) A \cos \theta_+}{(|E_{i\perp}|^2 / 2\eta_1) A \cos \theta_i} = |\tau_{\perp}|^2 \frac{\eta_1 \cos \theta_+}{\eta_2 \cos \theta_i}$$

$$T_{\parallel} = \frac{(\tau_{\parallel})^2 \eta_1 \cos \theta_+}{\eta_2 \cos \theta_i}$$

$$|\Gamma_{\perp}|^2 + |\tau_{\parallel}|^2 \frac{\eta_1 \cos \theta_+}{\eta_2 \cos \theta_i} = 1$$

$$P_i = P_r + P_t$$

$$|\Gamma_{\perp}|^2 + |\tau_{\perp}|^2 \frac{\eta_1 \cos \theta_+}{\eta_2 \cos \theta_i} = 1$$

