

تقدم لجنة EiCoM الاكاديمية

دفتر لمادة:

# أنظمة القوى الكهربائية

من شرح:

د. محمود سعادة

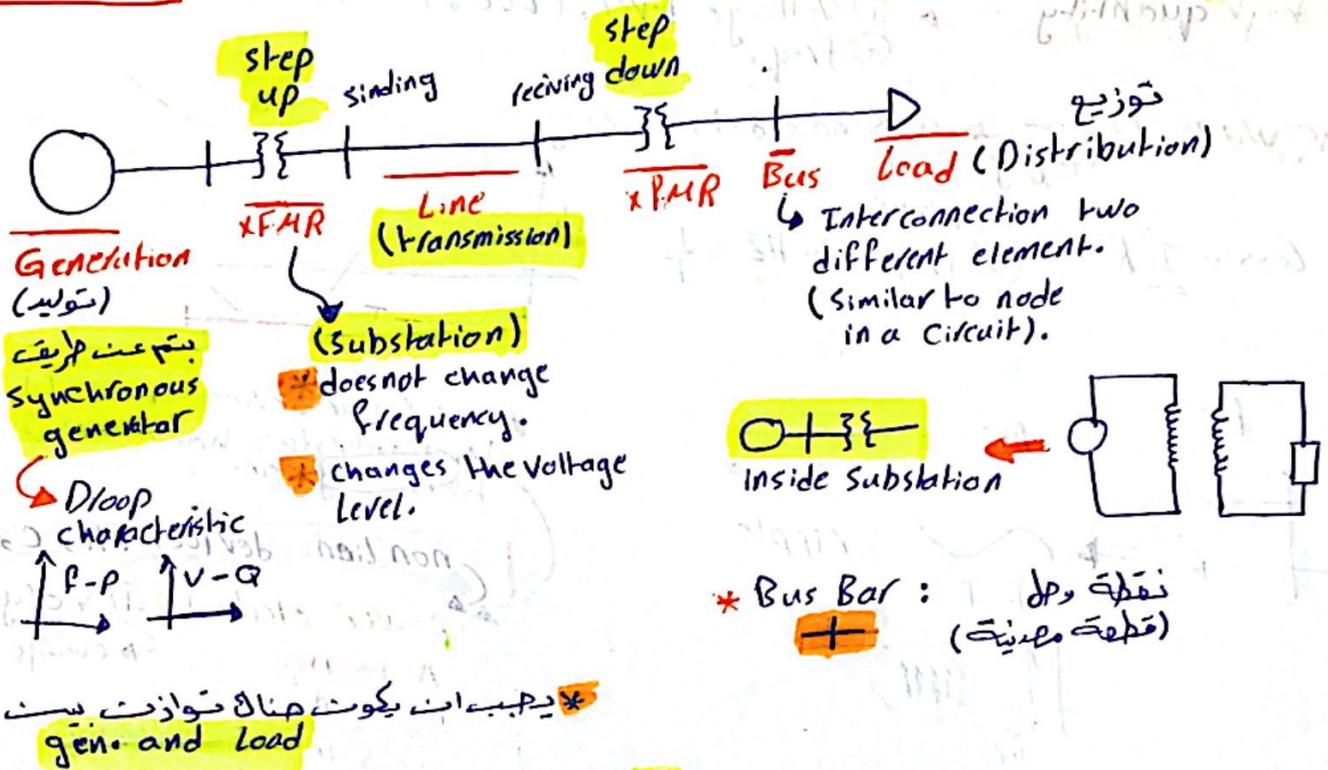
جزيل الشكر للطالب:

محمد جدوع



Generate (1) → Transmit (2) → Consume (Distribution) (3)

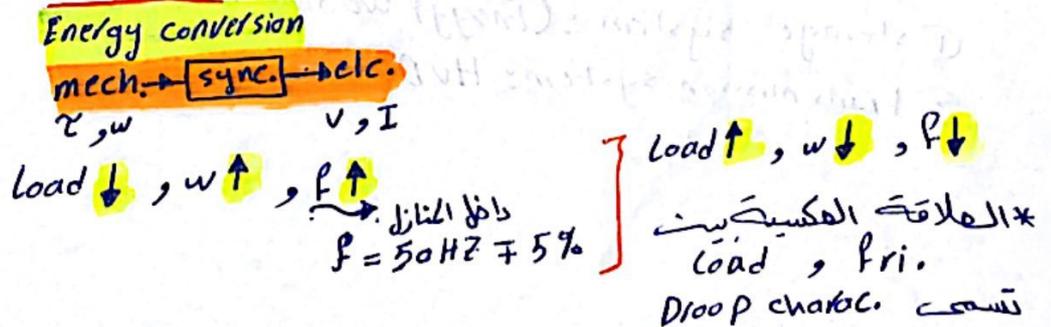
one-line diagram



\* What is the purpose of the power system?  
- to deliver power to the user.

- ① high quality
  - ② low cost. → ① Cost of generation. ② Cost of transmission. ③ Cost of distribution. ④ miss. Cost. (تكاليف اخرى)
  - ③ high reliability → ① Cost of fuel. ② efficiency.
- ① no interaction. (to deliver power without interactions).

\* in any power system Generation has to meet the demand. (Load)



\* Load shedding  $\propto$  اشياء مشتمل من غروب فيض.

\* Protection.  $\propto$

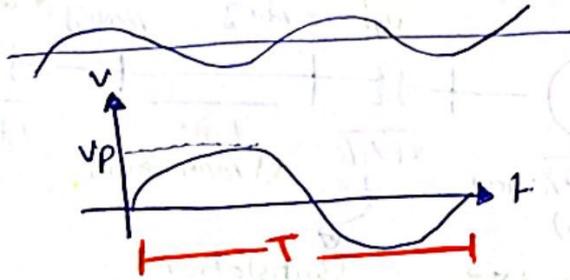
\* selectivity: to minimize affective Area.  $\checkmark$

\* high quality  $\rightarrow$  ① Voltage level. (220V) ③ T.H.D.  
 ② freq.

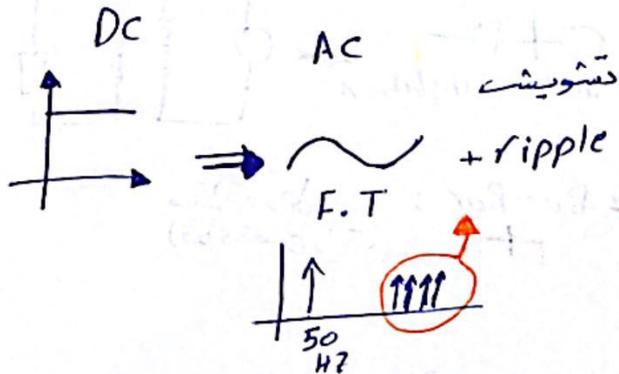
Ac wave: Iron  $\rightarrow$  ① Sinosoidal.

$$Loss = I^2 R$$

- ① Peak
- ② phase
- ③ freq. = 50 Hz =  $\frac{1}{T}$

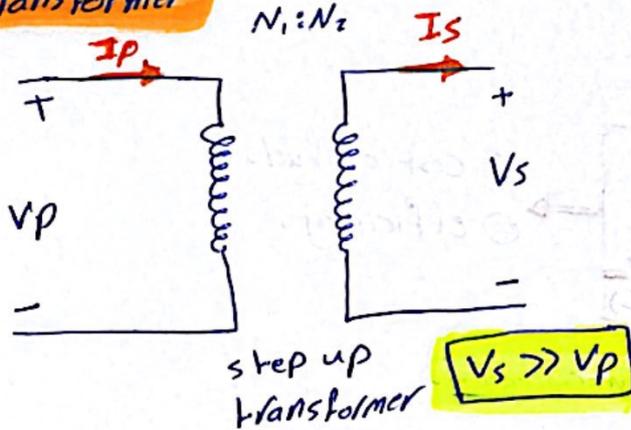


\* Non Linear devices (R, C, L) that introduce harmonics to the system.



- power electronics: ① chargers  $\rightarrow$  charge batteries DC  
 Ac to Dc  
 Dc to Ac  
 ② car. ③ inverter  
 ④ power supply. ⑤ UPS

Transformer



Ideal  $\Rightarrow$  no loss  $\Rightarrow P_{in} = P_{out}$

$$|V_p| |I_p| = |V_s| |I_s|$$

$$\frac{V_p}{V_s} = \frac{N_1}{N_2}$$

حد الفقد من كفاءة  
 هو تقليل التيار عشان  
 يقل Loss. (بقل التيار عن طريق  
 زيادة الفولتية).  
 وبالتالي برسل power لمسافات ابعد.

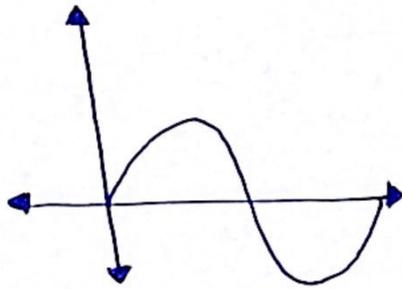
\* transformer  $\Rightarrow$  only in AC

- ① storage system: (Energy) we don't mean high voltage. (batteries)
- ② transmission system: HVDC.

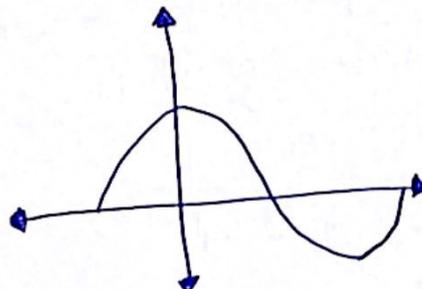
\* كل مازات freq. كل مازات تكلفة النظام وقت حجمه.

\* 10 Hz  $\Rightarrow$  10 cycle / sec.

\* 50 Hz  $\Rightarrow$  50 cycle / sec.  
 (عبرت بالسين (التردد)).  
 (عشرات ميك باستخدم 50 Hz).



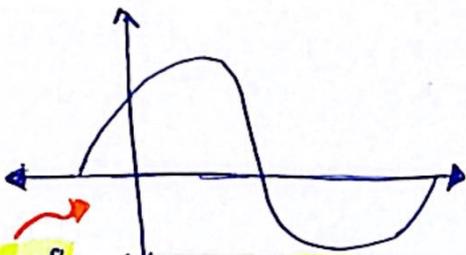
Sin



cos



sinusoidal



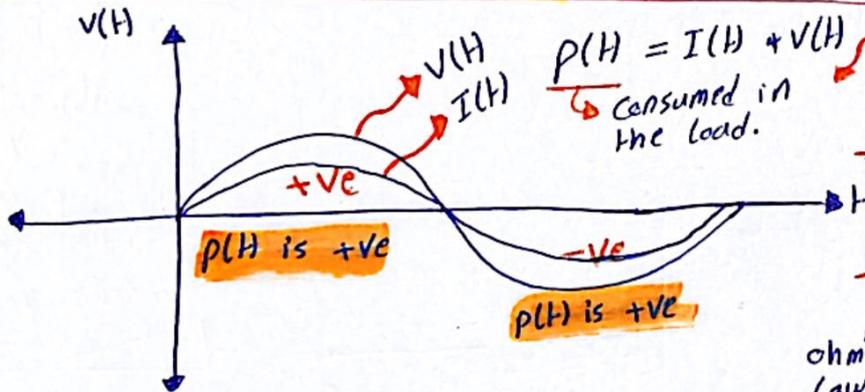
shift = 30°

sin (shift) or cos(shift) [5]

\* Any system has two reference:  
 ① voltage level, frequency (ground)?

$\rightarrow$  ②  $t = 0$ .

\* you always get to choose reference. (ONCE)



$P(t)$  is +ve

$P(t)$  is +ve

$P(t) = i(t) * v(t)$   
 consumed in the load.

$v_{avg} = zero$  و  $i_{avg} = zero$   
 $v(t) = V_p \sin(\omega t + \theta)$

Average  $v(t) = \int_0^T V_p \sin(\omega t + \theta) dt$

$i(t) = \frac{v(t)}{R} = \frac{V_p}{R} * \sin(\omega t + \theta)$   
 ohm's law

\* Average AC voltage is always zero.

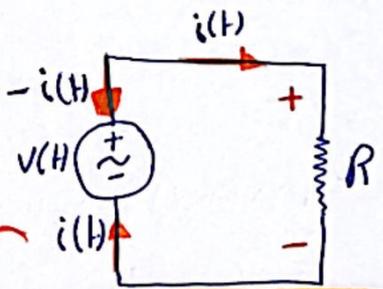
\* Average power =  $\int_0^T \frac{P(t)}{T} dt$

$P(t) = \int_0^T \frac{v(t) i(t)}{T} dt$

\*  $P(t)$ : instantaneous power.

power consumed in gen. =  $-i(t) * v(t) = -P(t)$

power generated in gen. =  $i(t) * v(t) = P(t)$



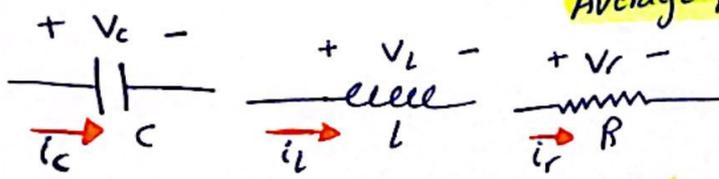
passive sign convention

$V = I * R$   
 $P = V * I$

\* the current entering the +ve terminal as voltage is defined. (consumed power)

[6]

(الطاقة المستهلكة)

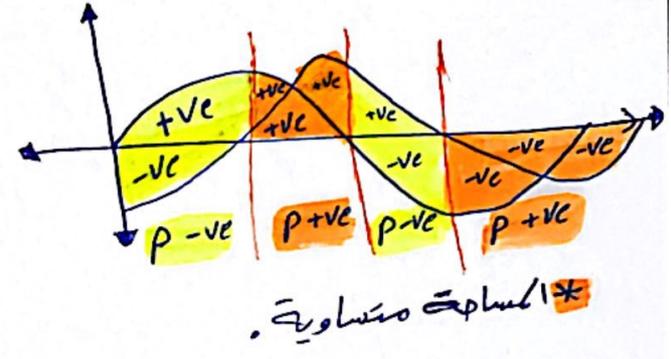


Average power = zero

$$I = C \frac{dV}{dt}$$

$$V = L \frac{di}{dt}$$

$\sin' = \cos \rightarrow \text{shift } 90^\circ$   
 $\cos' = -\sin \rightarrow \text{shift } 90^\circ$



$$i_c = C \frac{dv}{dt}$$

تقاوم القبولية

$$V_L = L \frac{dI}{dt}$$

تقاوم التيار

$$V_R = I \cdot R$$

تقاوم التيار

$$V_c(t) = V_m e^{j(\omega t + \theta)}$$

$$\frac{dV}{dt} = V_m e^{j(\omega t + \theta)} \cdot j\omega$$

$$i_c = C V_m j\omega e^{j(\omega t + \theta)}$$

$$Z_L = \frac{V_L}{I_L}$$

$$Z_L = j\omega L$$

$$|i_c| = j\omega C V_c$$

$$Z_c = \frac{V_c}{I_c} = \frac{1}{j\omega C}$$

[7]

$$* e^{j\theta} = \cos\theta + j \sin\theta$$

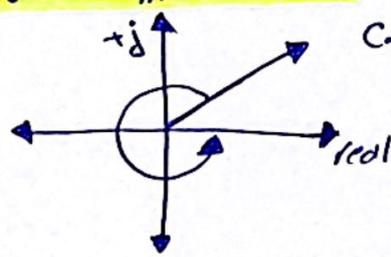
$$V(t) = V_m \cos(\omega t + \theta) + V_m j \sin(\omega t + \theta)$$

$$\rightarrow V(t) = V_m (\cos(\omega t + \theta) + j \sin(\omega t + \theta))$$

$$\rightarrow V(t) = V_m e^{j(\omega t + \theta)}$$

$$\xrightarrow{\omega=0} V(t) = V_m e^{j\theta}$$

$$V(t) = V_m \cos(\omega t + \theta) \leftrightarrow V_m e^{j\theta}$$



C.C.W rotation at  $\omega \rightarrow \omega = 2\pi f$

$$* V(t) = V_m e^{j(\omega t + \theta)}$$

only used to explain Z

\* passive linear circuit

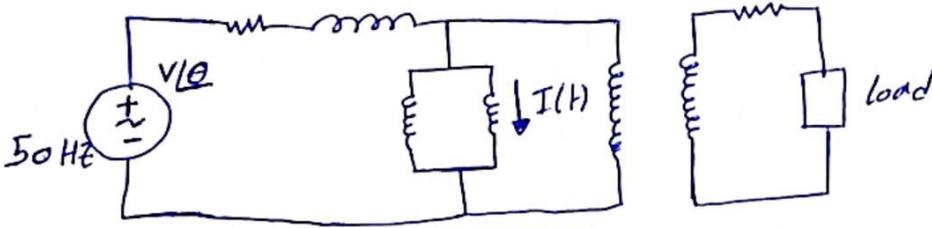
- \* FMR
- Motors
- gener.
- Transm.
- inductors
- capac.

مكونات  
 اعب  
 سيركيت  
 صندوق ما بين  
 على Preq.

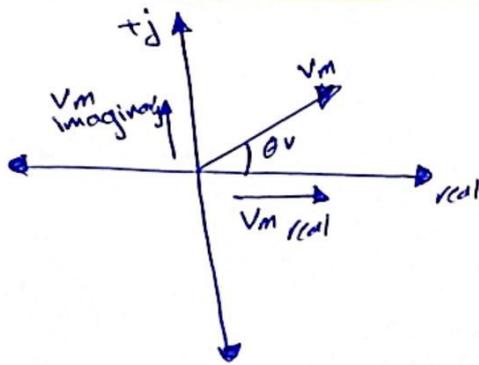
\* عناصر غير على Preq.  
 لازم ادخل على السيركيت  
 non-linear elements [8]

\* you should **never** have " $\sqrt{\phantom{x}} j$ " in the same equation.  
 \* you either solve in the time domain only " $\sqrt{\phantom{x}} No j$ " or in the freq. domain  $No \sqrt{\phantom{x}}$ .

Q) What is freq. of  $I(t)$ ?



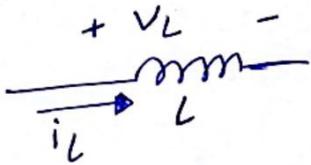
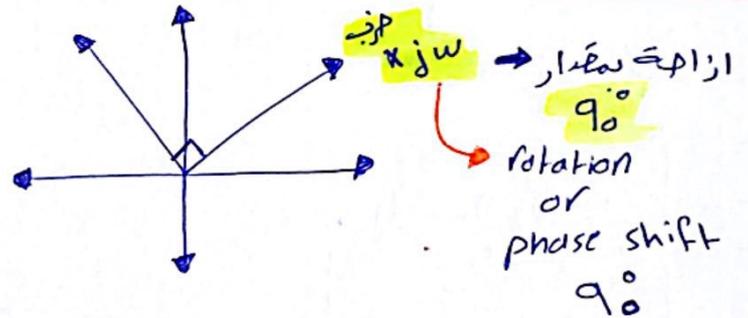
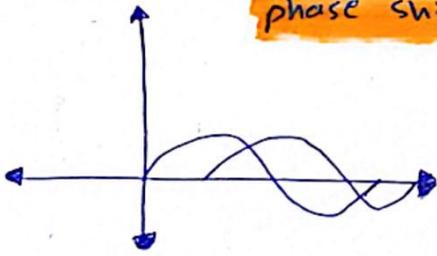
Sol  $\rightarrow$  Freq. of  $I(t) = 50 \text{ Hz}$



polar  $\rightarrow V_m \angle \theta$   
 rect  $\rightarrow V_m \cos \theta + j V_m \sin \theta$

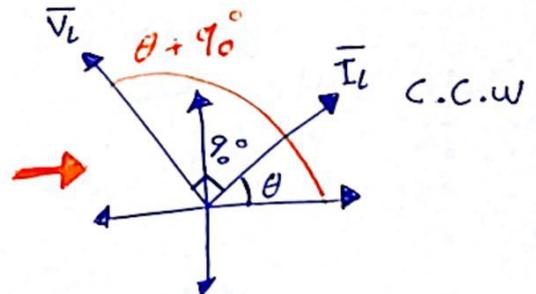
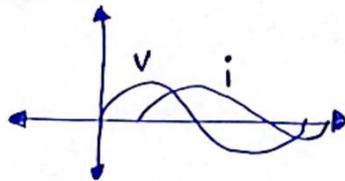
rect.  $\rightarrow$  polar تحويل من  
 ①  $V_m e^{j\theta}$ ,  $V_m \angle \theta = V_m \cos \theta + j V_m \sin \theta$   
 ②  $V_m = \sqrt{(V_m \cos \theta)^2 + (V_m \sin \theta)^2}$   
 ③  $\theta = \tan^{-1} \left( \frac{V_m \sin \theta}{V_m \cos \theta} \right)$

phase shift



$V_L = L \frac{di}{dt}$   
 $\times j\omega = 90^\circ$

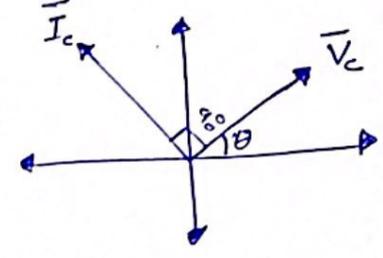
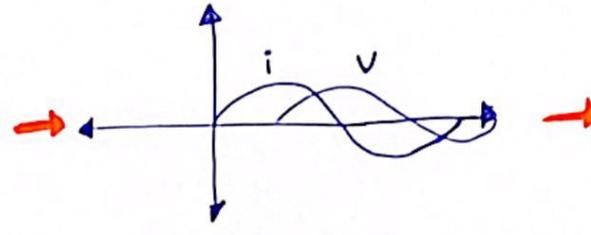
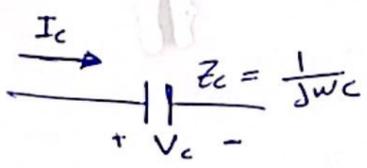
\* +ve area of  $v(t)$  is  $i(t)$  equal -ve area.



$V_L = I_L \omega L$   
 $V_L = j\omega L I_L \rightarrow$  Lagging phase shift  
 $P_{avg} = \int_0^T \frac{P(t)}{T} dt = \text{Zero}$

\* فرق الزاوية بين  $V$  و  $I$   $90^\circ$   
 $I \text{ Lag } V$

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$$V_c = I_c \cdot Z_c$$

$$V_c = I_c \cdot \frac{1}{j\omega C}$$

$$V_c = -j\omega C I_c \rightarrow \text{Leading}$$

**I lead V**

$$\theta_V = 30 \rightarrow \theta_I = -60$$

$$-60 \rightarrow 180 - 60 = 120$$

\* دالة الجهد في القابلية (Lag, lead) في الجهد.

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\* Average voltage and average current are zero.

① We don't care about Average V or I.

$$\text{Average power} = \int_0^T \frac{P(t)}{T} dt = \int_0^T \frac{V(t) \cdot I(t)}{T} dt$$

② we care about average power.

ohm's law  
 $V = I \cdot R$   
Resistive load.

$$= \int_0^T \frac{V^2(t)}{R \cdot T} dt = \int_0^T \frac{I^2(t) \cdot R}{T} dt$$

③ here fore we defined RMS V & I to make Avg calculation easier.

$$= \frac{1}{R} \left[ \frac{\int_0^T V^2(t) dt}{T} \right]^{\frac{1}{2}}$$

Foot of mean of square  
RMS of V(t)

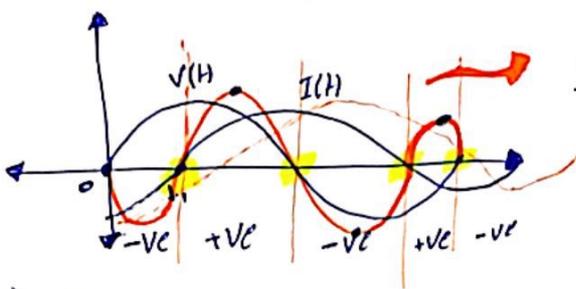
$$= R \left[ \frac{\int_0^T I^2(t) dt}{T} \right]^{\frac{1}{2}}$$

Foot of mean of square  
RMS I  $I_{rms}$ .

$$V(t) = \frac{V_{rms}}{R} = I_{rms} \cdot R$$

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\* if we have an inductive or capacitive load?



inductive  
I lag V

\* عند الفقد  $\pi$  قيمة  
 $P_{avg} = \text{Zero}$ .

\* عند ايمه نقطة غير  
الفقد  $\pi$

$P_{avg} = +ve$

$t = 0$   
 $t = t_1, t_2 > 0$

\*  $i(t - t_1) = i(0) =$   
when  $t = t_1$   
\* time shift.

\* كلما زاد الزمن كلما طار  
تأخير اكثر. (Lagging)

\* عند تقاطع القاطع  
 $P = \text{Zero} \Leftarrow x\text{-axis}$

\* pure inductance  
 $A_{avg} +ve = A_{avg} -ve$   
and phase shift =  $90^\circ$

\* phase angle  $\theta_V - \theta_I$   
 $\theta_V - \theta_I > 0 \rightarrow \text{Lead}$ .

\* freq of V & I = 50Hz

\* freq of P =

\* كل دورة للتيار، يقابلها دورتين  
لليور.

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(Source load)  
(consume) استهلاك الطاقة

\* power will oscillate  
at twice the freq.

\* phase angle zero  
ينطبقوا على بعض النقاط  
تخرج كل المسلمات موجبة  
(Resist.)

\*  $P -ve \rightarrow \text{supply}$ .

Q: Reactive power.

P: Average power.  $\rightarrow$  the actual power being  
consumed by the load.

ex: ① light.

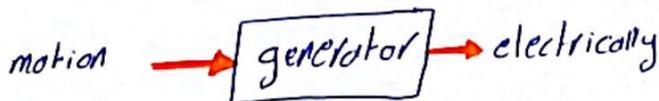
② heat.

③ Washer, fan  
motion

real  $\rightarrow$  rate of energy being consumed

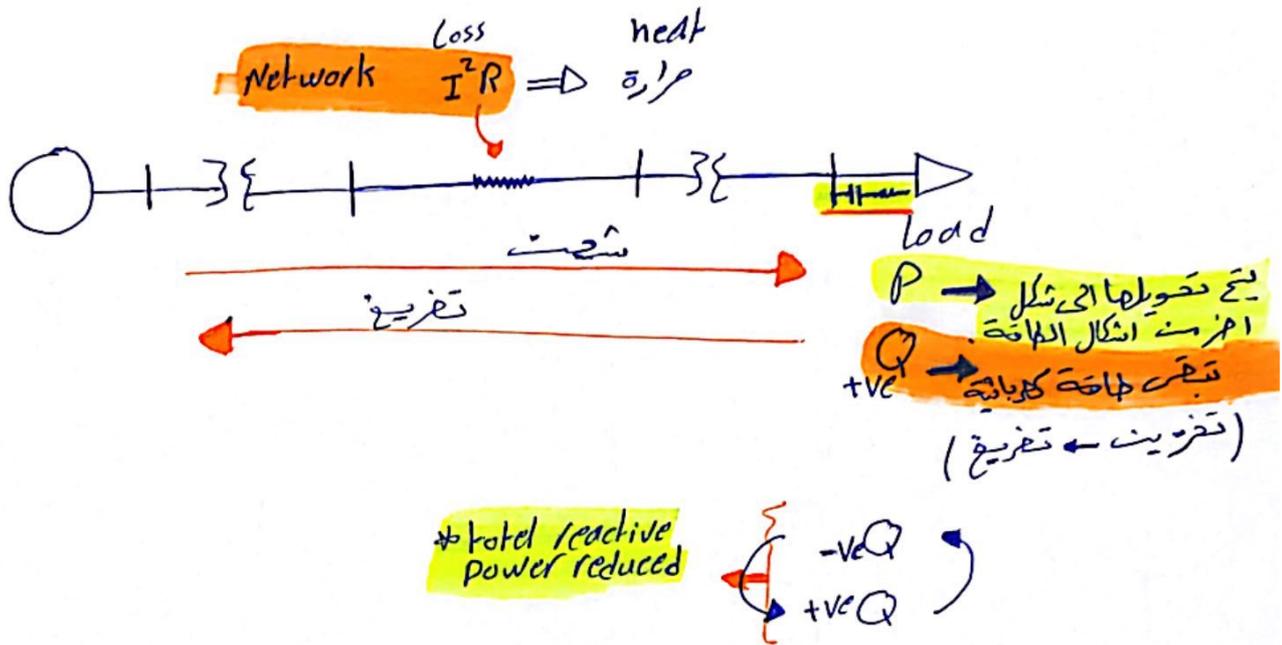
reactive  $\rightarrow$  rate of energy being stored  
and released.

$\rightarrow$  converted from  
elect. to other  
form of energy.



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\* Power factor correction (معدل القدرة)



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$\vec{S}$ : Complex power

$\vec{S} = P + jQ$   $\rightarrow$  real number

$P = |S| \cos \theta$

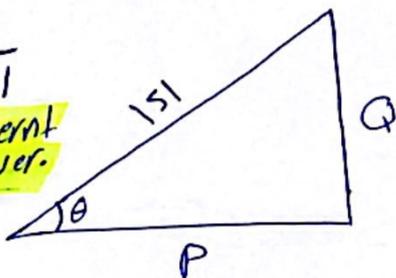
$Q = |S| \sin \theta$

$\theta = \theta_v - \theta_i$

$\theta = \tan^{-1} \left( \frac{Q}{P} \right)$

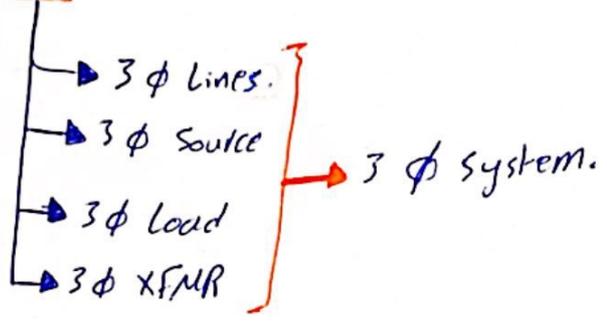
P.f. =  $\frac{P}{|S|}$

$|S|$ : Apperent power.

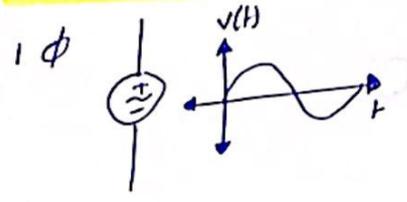


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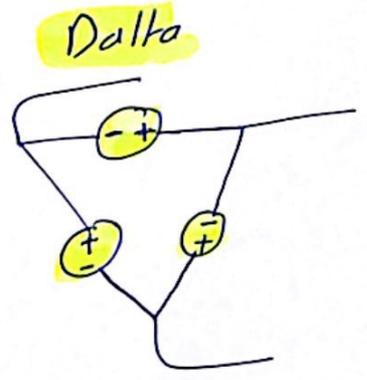
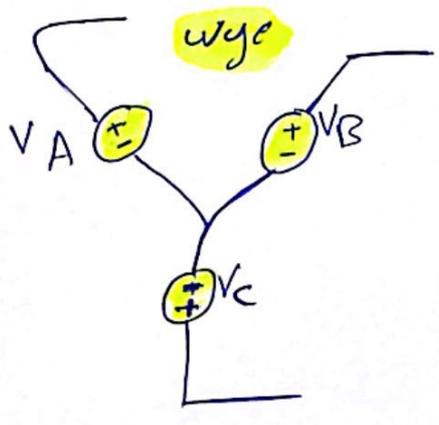
3φ → 3 single phase



3 φ source



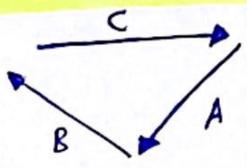
3 φ source



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- ① Balanced 3φ source
- ② unbalanced 3φ source.

\*  $\bar{V}_A + \bar{V}_B + \bar{V}_C = \text{zero}$

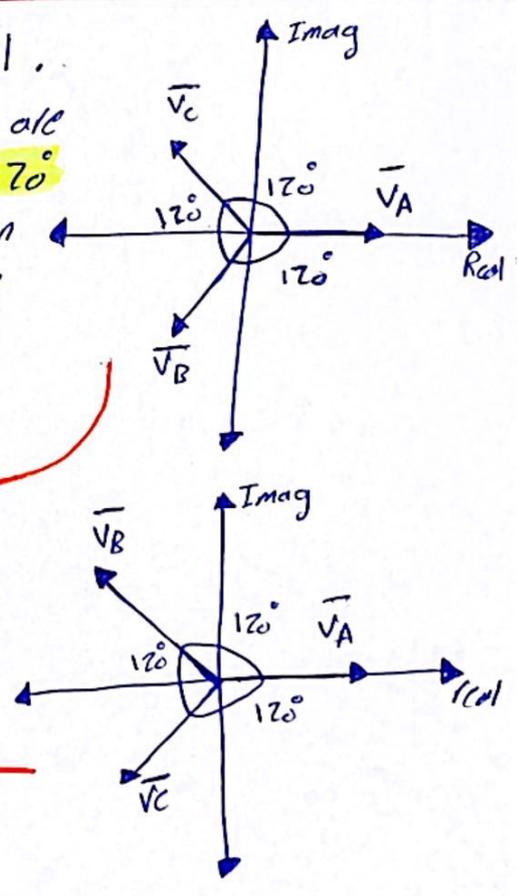


- ①  $|\bar{V}_A| = |\bar{V}_B| = |\bar{V}_C|$
- ② Angle  $\theta_A \rightarrow \theta_B \rightarrow \theta_C$  are equally spaced →  $120^\circ$  phase shift between each two consecutive phases.

$\frac{360}{3} = 120^\circ$

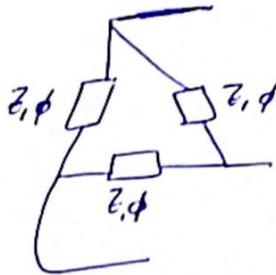
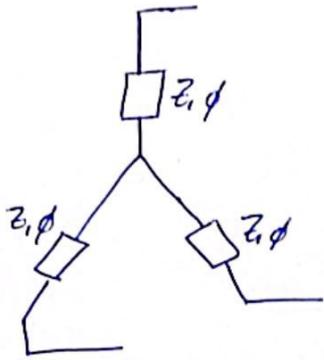
$\bar{V}_A \rightarrow \bar{V}_C \rightarrow \bar{V}_B$   
ACB sequence.  
(-ve sequence)

$\bar{V}_A \rightarrow \bar{V}_B \rightarrow \bar{V}_C$   
ABC sequence  
(+ve sequence)



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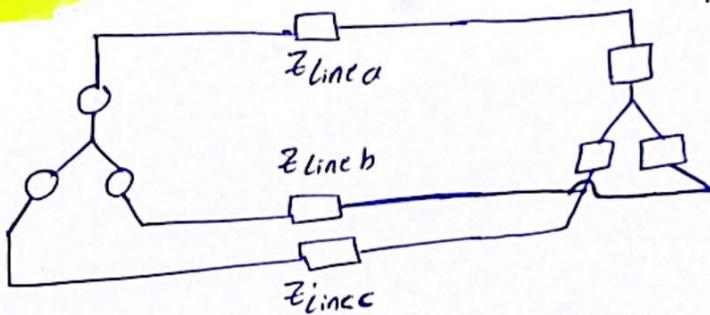
**3  $\phi$  load:**



→ For 3  $\phi$  loads to be balanced  
 $Z_a = Z_b = Z_c = Z_{\phi}$

\*  $P_t(H) = P_a(H) + P_b(H) + P_c(H)$  → always true

**3  $\phi$  system:** → if every components in the system is balanced, the system is balanced:



- ① source .
- ② load .
- ③ line .

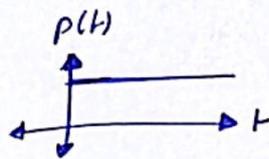
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\* in balanced three phase system:

- ① the current passing in the system will also be balanced.
- ② each phase will consume the same amount of real and reactive power.

$P_A = P_B = P_C$

③  $P_{3\phi}(H) = \text{constant}$



\* power in 3  $\phi$  system:

$P_{3\phi} = 3 \cdot P_{1\phi}$

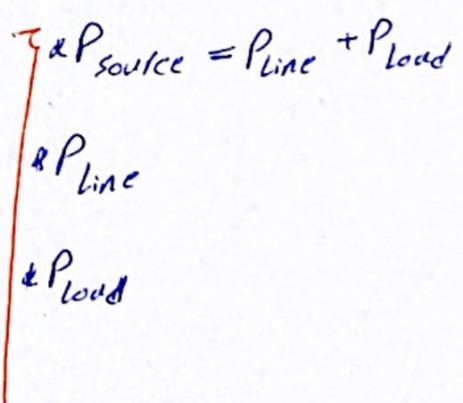
$Q_{3\phi} = 3 \cdot Q_{1\phi}$

$S_{3\phi} = 3 \cdot S_{1\phi}$

$S_{1\phi} = \bar{V}_{1\phi} \cdot \bar{I}_{1\phi}^* = |V_{1\phi}| |I_{1\phi}| \angle(\theta_v - \theta_i)$

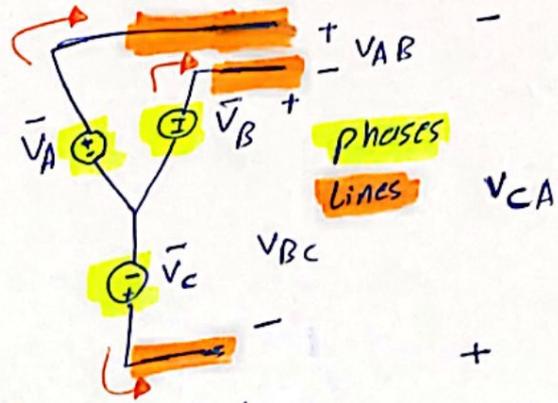
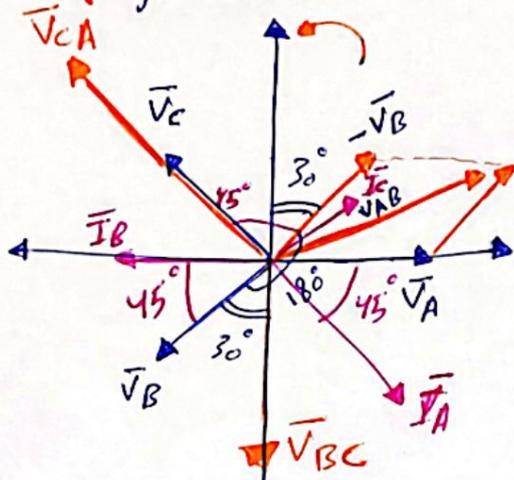
$S_{3\phi} = 3 \cdot \bar{V}_{1\phi} \cdot \bar{I}_{1\phi}^* = 3 |V_{1\phi}| |I_{1\phi}| \angle(\theta_v - \theta_i)$

wye connection  $\frac{1}{\sqrt{3}} |V_L| \cdot I_L \angle(\theta_{V_L} - \theta_{I_L}) = \sqrt{3} |V_L| |I_L| \angle(\theta_{V_L} - \theta_{I_L})$  **20**



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\*  $V$  &  $I$  in 3 phases:  
(wye-connected 3  $\phi$  source)



\*  $V_A, V_B$  and  $V_C$  are phase voltages.  
\* فرق الزاوية بين  $V_A$  و  $V_{AB}$   $75^\circ$

\*  $V_{AB}$  &  $V_{BC}$  &  $V_{CA}$  are line voltages.

\*  $V_{AB} = \bar{V}_A - \bar{V}_B = \bar{V}_A + (-\bar{V}_B)$

\*  $V_{AB} = |V_A| \angle 0^\circ + |V_A| \angle 60^\circ = |V_A| \angle 0^\circ + |V_A| \cos(60^\circ) + j|V_A| \sin(60^\circ)$

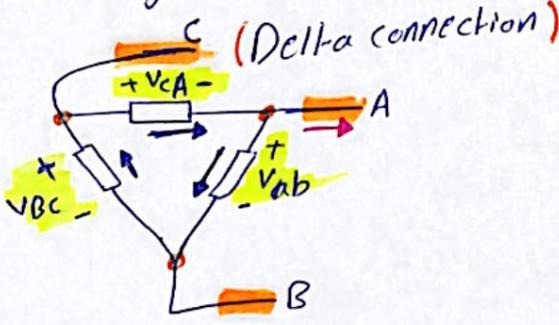
$$V_{AB} = \sqrt{[|V_A| + |V_A| \cos(60^\circ)]^2 + [|V_A| \sin(60^\circ)]^2}$$

$$\tan^{-1} \frac{|V_A| \sin(60^\circ)}{|V_A| + |V_A| \cos(60^\circ)}$$

$V_{AB} = \sqrt{3} |V_A| \angle 30^\circ$   
 $\angle V_A + 30^\circ$

\*  $|V_{AB}| = |V_{BC}| = |V_{CA}|$

\* In a wye connection line current & phase currents are equal.



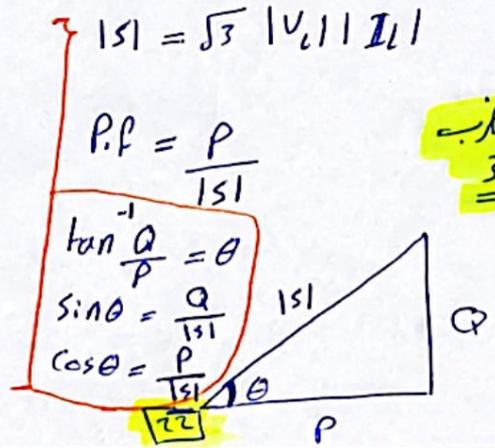
\* تكامله معارلات متساوية 20

$\theta_{V, \phi} - \theta_{I, \phi} = \theta$

$\bar{S}_{3, \phi} = P_{3, \phi} + j Q_{3, \phi}$

$P = \sqrt{3} |V_L| |I_L| \cos \theta$

$Q = \sqrt{3} |V_L| |I_L| \sin \theta$



\* زاوية  $S_{3, \phi}$  متساوية زاوية  $S_{1, \phi}$   
\*  $P.F._{3\phi} = P.F._{1\phi}$

load    old

$P_{old}$

$Q_{old} \rightarrow$  inductive

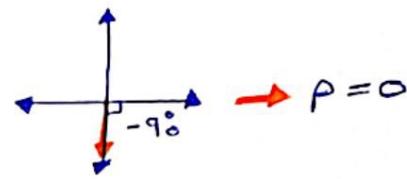
$S_{old} = P + jQ \rightarrow +ve$

new

Load + capacitor

$P_{old}$   
 $Q_{old}$

$Q, P=0$   
 $-ve$   
 $Z = \frac{1}{j\omega C} = \frac{-j}{\omega C} = \frac{1}{\omega C} \angle -90^\circ$



$|S| = \sqrt{Q^2 + P^2}$

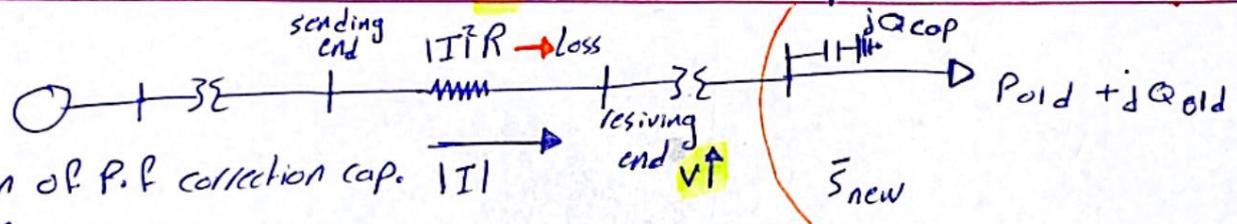
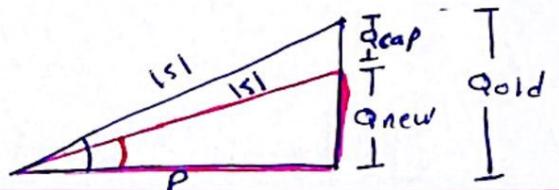
$\theta = \tan^{-1}(\frac{Q}{P})$

$V = 0.95 \rightarrow 1.05$   
 $220V$

\*  $P_{new} = P_{old}$

$Q_{new} = Q_{old} + Q_{cap.}$   
 $+ve \quad -ve$

$Q_{new} < Q_{old}$



\* Addition of P.f correction cap.

\* power factor improved.

\*  $|I| \downarrow$ .

\*  $|I|^2 \times R \downarrow$  loss. ①

\* efficiency  $\uparrow$ .

\* P.f = 0.7  $\rightarrow$  0.95

$\square \times$

\*  $Q_{cap} = Q_{old}$  \*

كابتكبر لاضو بتكبر قبيته

\*  $Q_{new} = -ve$

\* تكاليف اكثر \*

\* voltage drop  $\downarrow$  ②

(( 0.95  $\rightarrow$  1.05 ))  $\square \rightarrow 220$

$\square$

## \* per-unit %

- \* power system quantities such as voltage, current, power and impedance are often expressed in per-unit or percent of specified base values.
- \* For example if a base voltage of 20 kV is specified then the voltage 18 kV is  $(18/20) = 0.9$  per unit or 90%.
- \* calculations can then be made with per-unit quantities rather than with the actual quantities.
- \* one advantage of the per-unit system is that by properly specifying base quantities the transformer equivalent circuit can be simplified.
- \* The ideal transformer winding can be eliminated such that voltages, currents, and external impedance and admittances expressed in per-unit do not change when they are referred from one side of a transformer to the other.

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- \* This is a significant advantage when analyzing large systems with hundreds of transformers.



- \* The other advantage is that per-unit impedances of electrical systems components of similar type usually lie within a narrow numerical range when the equipment ratings are used as base values.   
 ↳ gross errors are easily distracted.

- \* usually manufacturers specify impedances in per-unit.

$$\text{perunit quantity} = \frac{\text{Actual quantity}}{\text{base value of quantity}}$$

unitless

has a unit

has a unit

- \* base values are real number.
- \* Therefore, the angle of the perunit quantity is the same as the angle of the actual quantity.

$$\begin{aligned} \bar{S} &= P + jQ \\ \bar{Z} &= R + jX \\ \bar{Y} &= G + jB \end{aligned}$$

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\*  $\frac{\text{قيمة حقيقية}}{\text{النسبة المئوية}}$   
 % per unit

- power.
- voltage.
- Impedance.
- current.

- time domain
- inductance (time domain)
  - capacitor.
  - real and reactive power.

\* Two independent base values can be arbitrarily selected at one point in a power system.

\* You need to have the following base values

على نحو استبدادي

\* other bases has to follow electrical laws.

- ① base voltage. ✓
- ② base current. = base Power / base Voltage
- ③ base impedance. = base voltage / base current.
- ④ base power. ✓

\* one point in a power system. → is divided into zones for the purpose of p.u. calculations.  
 → the zones divided by XFMR.

①  $P_{base \ 1\phi} = Q_{base \ 1\phi} = S_{base \ 1\phi}$

②  $I_{base} = \frac{S_{base \ 1\phi}}{V_{baseLN}}$

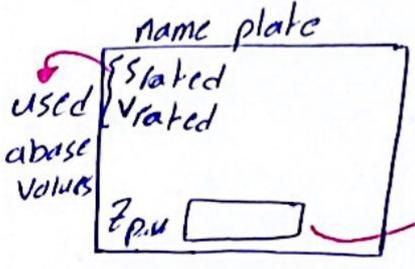
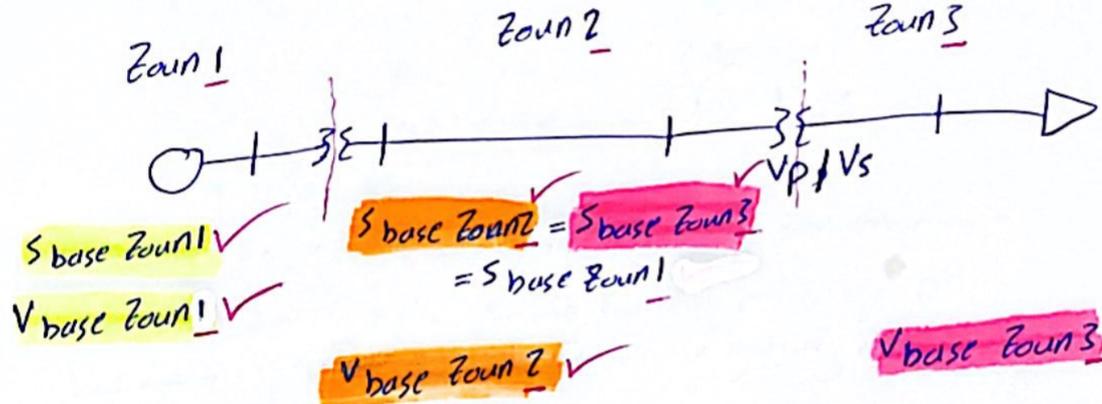
③  $Z_{base} = R_{base} = X_{base} = \frac{V_{baseLN}}{I_{base}} = \frac{V_{baseLN}^2}{S_{base \ 1\phi}}$

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④  $Y_{base} = G_{base} = B_{base} = \frac{1}{Z_{base}}$

\* Two important Rules:

- ① the value of  $S_{base \ 1\phi}$  is the same for the entire power system of concern.
- ② the ratio of the voltage bases on either side of a transformer is selected to be the same as the ratio of the transformer voltage ratings.



$Z_{p.u} = \frac{Z_{actual}}{Z_{base}} \rightarrow$  قوتة باقية  
 $= \frac{V_{base}^2}{S_{base}} \rightarrow$  ?  
 selected as machine rated value.

28

القيمة العنصرية JS \*  
 rated power  
 . القيمة

**Example 3.3 (109):** A single-phase two-winding transformer is rated 20 kVA, 480/170 volts, 60 Hz. The equivalent leakage impedance of the transformer referred to the 170 volt winding, denoted winding 2, is  $Z_{eq2} = 0.0525 \angle 78.13^\circ \Omega$ . Using the transformer ratings as base values, determine the per unit leakage impedance referred to winding 2 and referred to winding 1.

→  $S_{base} = 20 \text{ kVA}$

$V_{base1} = 480 \text{ V}$

$V_{base2} = 170 \text{ V}$

$Z_{base2} = \frac{V_{base2}^2}{S_{base}} = \frac{170^2}{20 \text{ k}}$

$Z_{base2} = 0.72 \Omega$  → base values are always real number.

$Z_{2 \text{ p.u.}} = \frac{Z_{\text{Actual}}}{Z_{base}} = \frac{0.0525 \angle 78.13^\circ}{0.72 \angle 0^\circ}$

$Z_{2 \text{ p.u.}} = 0.0729 \angle 78.13^\circ \text{ per unit}$

$Z_{base1} = \frac{480^2}{20 \text{ k}} = 11.52 \Omega$

$Z_{eq1}(\text{Actual}) = a^2 Z_{eq2} = \left(\frac{N_1}{N_2}\right)^2 Z_{eq2}$   
 $= \left(\frac{480}{170}\right)^2 \cdot 0.0525 \angle 78.13^\circ$

$Z_{eq1} = 0.84 \angle 78.13^\circ \Omega$

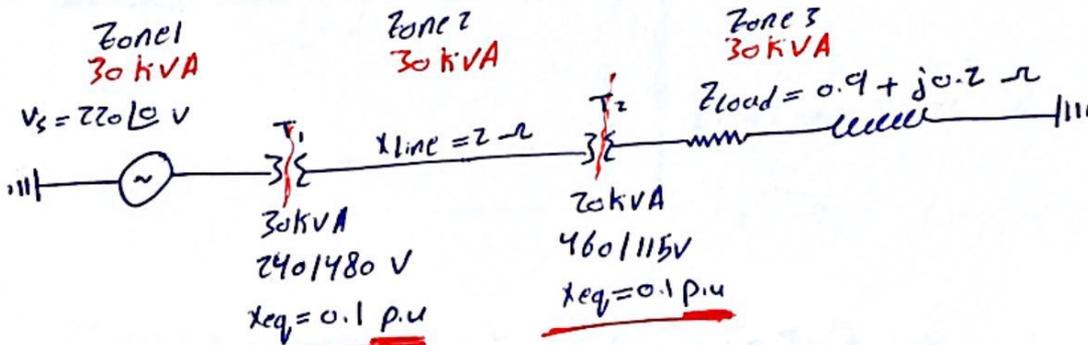
$Z_{1 \text{ p.u.}} = \frac{Z_{eq}(\text{Actual})}{Z_{base}} = \frac{0.84 \angle 78.13^\circ}{11.52}$

$Z_{1 \text{ p.u.}} = 0.0729 \angle 78.13^\circ \text{ per unit}$

$\frac{V_{base1}}{V_{base2}} = \frac{V_{rated1}}{V_{rated2}} = \frac{480}{170}$

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**Example 3.4 (112):**  $S_{base} = 30 \text{ kVA}$ ,  $V_{base1} = 240 \text{ V}$



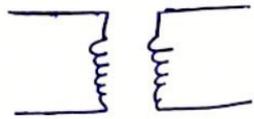
\*Two important Rules:

- ① Arbitrary choose two values and calculate the other two.
- ② The base power is the same for the entire system.
- ③ The ratio of the transformer voltage ratings.

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one line Diagram system.  
↳ is not circuit.



Circuit model

Zone 1

$$30 \text{ kVA} = S$$

$$240 \text{ V} = V_{\text{base}}$$

$$Z_{\text{base1}} = \frac{V_{\text{base}}^2}{S_{\text{base}}} = \frac{240^2}{30\text{k}} = 1.92 \Omega$$

$$I_{\text{base}} = \frac{V_{\text{base}}}{Z_{\text{base}}} = \frac{S_{\text{base}}}{V_{\text{base}}} = \frac{30\text{k}}{240} = 125 \text{ A}$$

Zone 2

$$S = 30 \text{ kVA}$$

$$V_{\text{base}} = \frac{480}{240} \times 240 = 480 \text{ V}$$

$$Z_{\text{base2}} = \frac{480^2}{30\text{k}} = 7.68 \Omega$$

$$I_{\text{base}} = \frac{30\text{k}}{480} = 62.5 \text{ A}$$

Zone 3

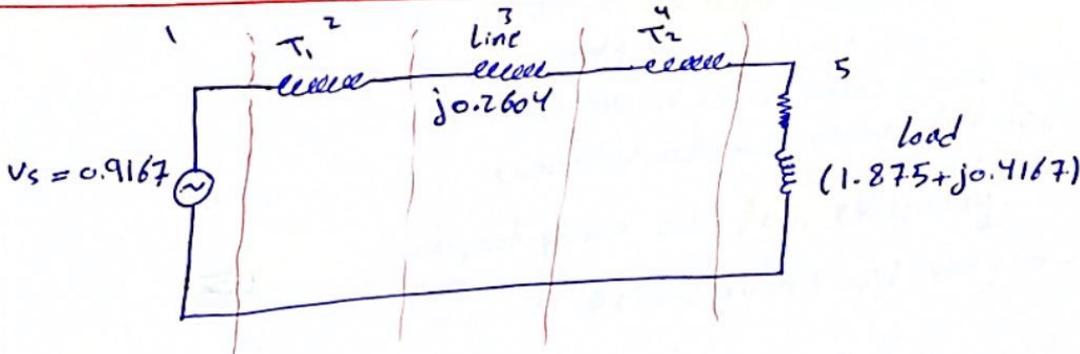
$$S = 30 \text{ kVA}$$

$$V_{\text{base}} = \frac{115}{460} \times 480 = 120 \text{ V}$$

$$Z_{\text{base3}} = \frac{120^2}{30\text{k}} = 0.48 \Omega$$

$$I_{\text{base}} = \frac{30\text{k}}{120} = 250 \text{ A}$$

31



$$Z_{\text{load p.u}} = \frac{Z_{\text{Actual}}}{Z_{\text{Load base}}} = \frac{0.9 + j0.2}{0.48} = 1.875 + j0.4167 \text{ p.u.}$$

$$V_s \text{ p.u.} = \frac{220}{240} = 0.9167 \text{ p.u.}$$

$$Z_{\text{Line}} = R + jX = j2 \rightarrow Z_{\text{Line p.u}} = \frac{j2}{7.68} = j0.2604 \text{ p.u.}$$

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$$X_{T_2}(\text{Actual}) = X_{T_2} \times Z_{\text{base old}}$$

$$X_{T_2} \text{ pu new} = \frac{X_{T_2}(\text{Actual}) Z_{\text{oun 2}}}{Z_{\text{base new } Z_{\text{oun 2}}}}$$

$$X_{T_2} \text{ pu } Z_{\text{oun 2}} = 0.1 \times \left( \frac{460}{20k} \right)^2 \Big/ \frac{480^2}{30k} \rightarrow Z_{\text{base new}} \text{ (we used these base value for the system)}$$

$$= 0.1378 \text{ p.u.}$$

$$X_{\text{p.u. new}} = X_{\text{p.u. old}} \times \left( \frac{V_{\text{old}}}{V_{\text{new}}} \right)^2 \times \frac{S_{\text{new}}}{S_{\text{old}}} = 0.1378 \text{ p.u.}$$

(primary)

$$X_{T_2} \text{ p.u. (secondary)} = 0.1 \times \left( \frac{115}{120} \right)^2 \times \frac{30k}{20k} = 0.1378 \text{ p.u.}$$

\* p.u value is the same refer to either side of the transformer.

\* Actual is different when refer to primary or secondary

proper base values  
 ① same base power for entire system.

② same trans. ratio.

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**Ex 3** A three phase bank of three 75 kVA single phase transformers steps down the three phase primary feed voltage from 12.47 kV to 208/120 V and is connected as shown below. The three phase bank is supplying a three phase wye-connected induction motor load operating at 0.75 power factor, 90 kW output power and 80% efficiency.

a) Draw two phasor diagrams for the primary and secondary voltages.

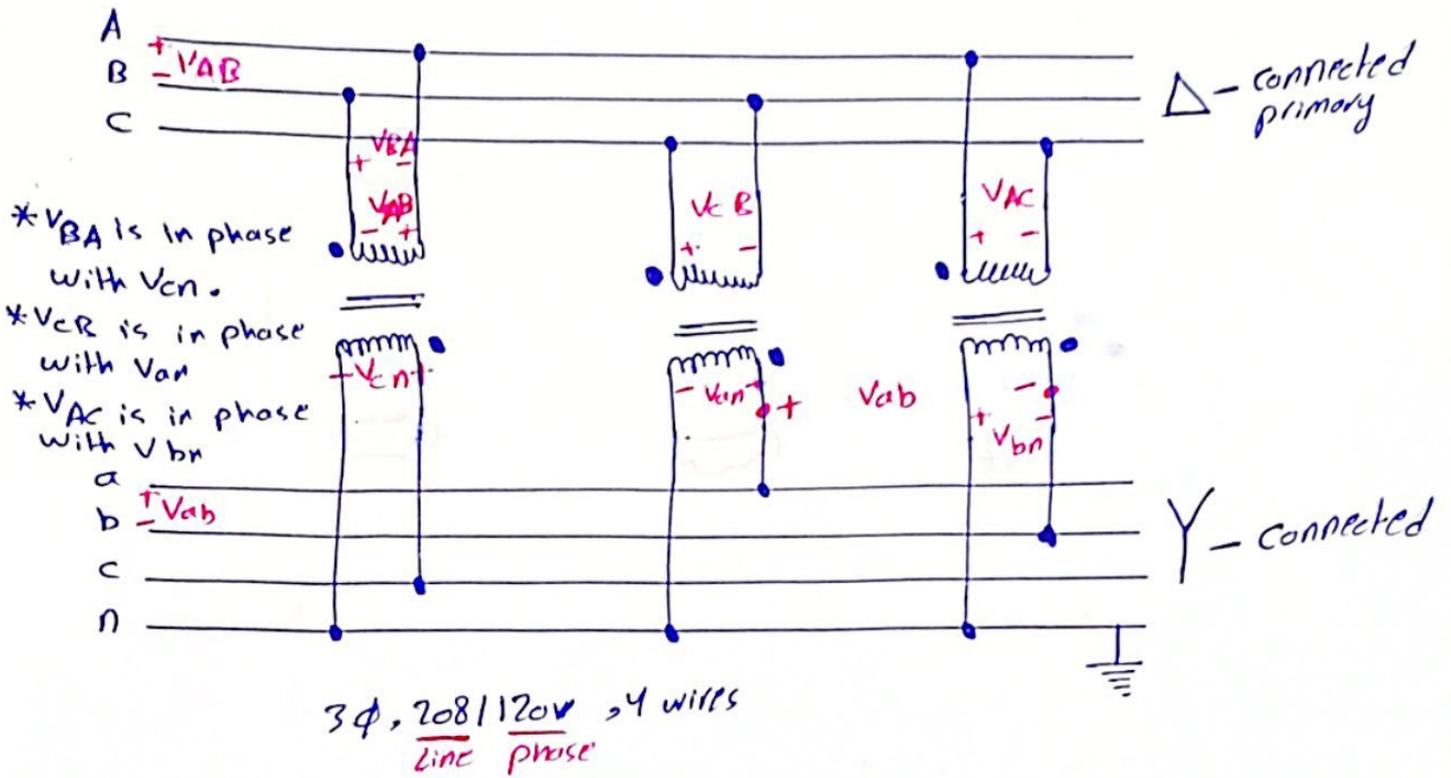
b) Determine the phase difference between  $\bar{V}_{AB}$  &  $\bar{V}_{ab}$ .

c) Determine the line current (mag. & angle) flowing into each secondary-phase conductor.

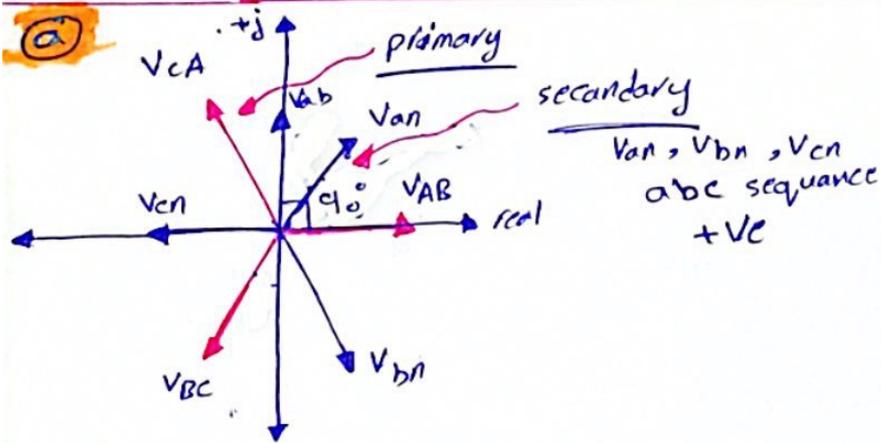
the current (mag. & angle) flowing in each primary winding, and the line current (mag. & angle) flowing in each phase of the primary feed.

→ Feeder = Line

3  $\phi$ , 12.47, 3 wires



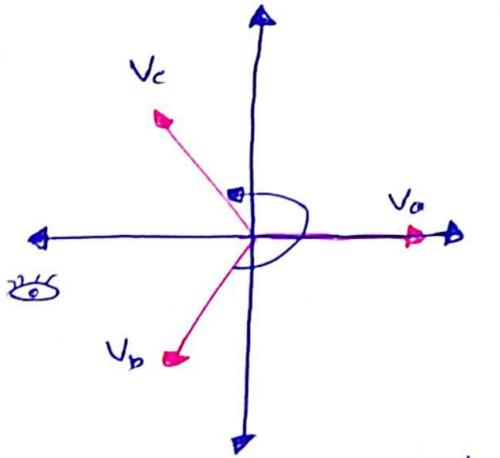
35



- B)  $V_{ab}$  leads  $V_{AB}$  by  $90^\circ$ .  
 or  $V_{AB}$  lags  $V_{ab}$  by  $90^\circ$ .  
 or  $V_{AB}$  leads  $V_{ab}$  by  $-90^\circ$ .  
 or  $V_{ab}$  lags  $V_{AB}$  by  $270^\circ$ .  
 or  $V_{ab}$  lags  $V_{AB}$  by  $270^\circ$ .

36

\*both are balanced.



$V_a, V_b, V_c, V_a, V_b$   
 $\underline{abc}$   
 +ve sequence

$$\theta_a - 120^\circ = \theta_b$$

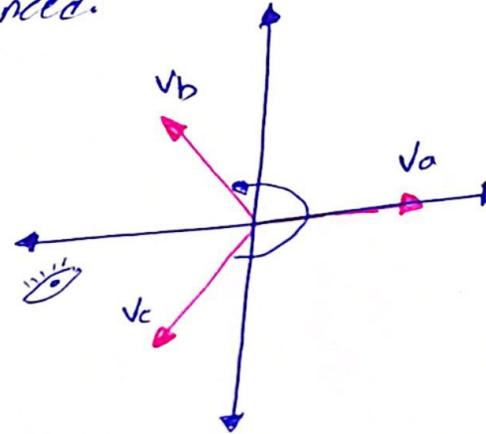
$$\theta_b - 120^\circ = \theta_c$$

$$\theta_c - 120^\circ = \theta_a$$

\* $\theta_a$  lead  $\theta_b$  by  $120^\circ$ .

\* $\theta_b$  lead  $\theta_c$  by  $120^\circ$ .

\* $\theta_c$  lead  $\theta_a$  by  $120^\circ$ .



$V_b, V_a, V_c, V_b, V_a, V_c$   
 $\underline{acb}$   
 -ve sequence

$$\theta_a + 120^\circ = \theta_b$$

$$\theta_b + 120^\circ = \theta_c$$

$$\theta_c + 120^\circ = \theta_a$$

**CH-4**

- ① Copper
  - ② Aluminum
- used in Transmission line.

**- Copper:**

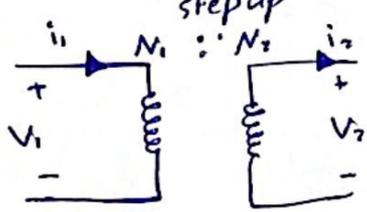
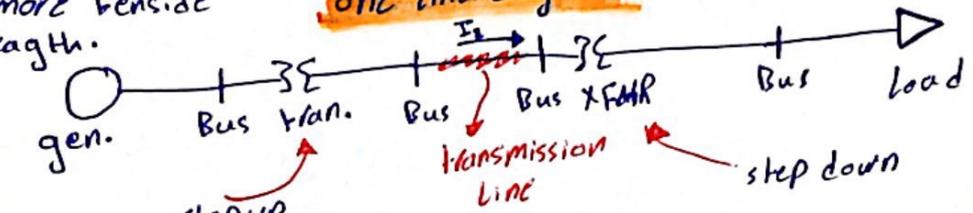
- ① has better conductivity.
- ② more expensive.
- ③ heavier.
- ④ Less flexible.
- ⑤ has more tensile strength.

**- Aluminum:**

- ① cheaper.
- ② loss conductivity.
- ③ Lightweight.
- ④ more flexible.
- ⑤ has less tensile strength.

→ has replaced copper as most used.

**one line diagram**



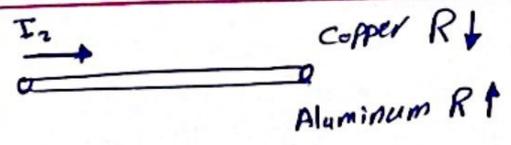
$$\frac{V_1}{V_2} = \frac{N_1}{N_2}$$

$$\frac{I_1}{I_2} = \frac{N_2}{N_1}$$

$$V_1 < V_2 \rightarrow I_2 < I_1$$

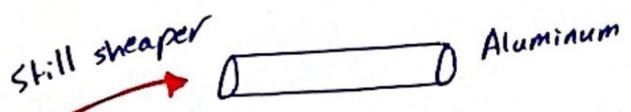
$$\frac{V_1}{V_2} = \frac{I_2}{I_1}$$

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**ACSR** ⇒ Aluminum conductor steel reinforced.

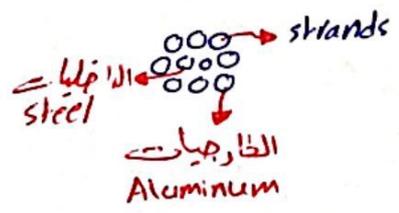
Loss =  $I^2 R$



\* we use Aluminum conductors with large cross section area to overcome its higher resistivity.

- \* Conductors are made from strands for the following reasons:
  - ① flexibility.
  - ② flexibility in manufacturing.
- twisting (stranding)
- to hold strands together.

\* typically overhead transmission line are Bare {no insulator}.



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L, C, R  
Parameters → we will in the transmission line module

① R (Resistance)

$$R_{dc, T} = \frac{\rho_T L}{A} \text{ (}\Omega\text{)}$$

L: length (m)

A: cross-sectional Area (m<sup>2</sup>)

$\rho_T$ : Resistivity ( $\Omega \cdot m$ )  
at temp T

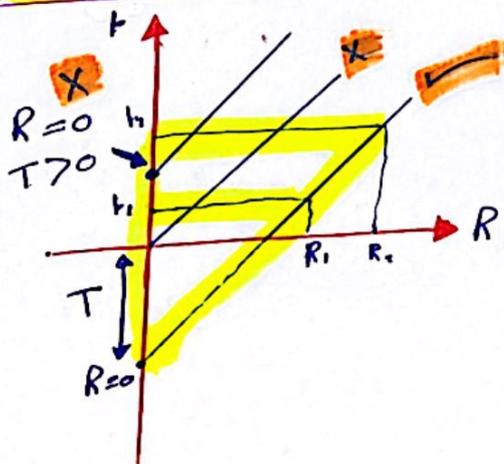
\* Resistance depends on:

- ① material.
- ② length.
- ③ cross-section.
- ④ temperature.
- ⑤ spiraling. (طردية) →  $\rightarrow$   $\rightarrow$  Length  $\rightarrow$  4E

⑥ frequency (skin effect).

\* effective area is less than Actual Area.

\* A<sub>res.</sub> > dc res.



$$\frac{R_2}{R_1} = \frac{T + t_2}{T + t_1}$$

↓  
depends on material

\* The point of steel:  
① increase tensile strength.

\* cmil: unit of measurement of Area.

1 cmil =  Area

1 mil =  $\frac{1}{1000}$  inch

1 inch = 2.54 cm

1 mile = 1.609 km

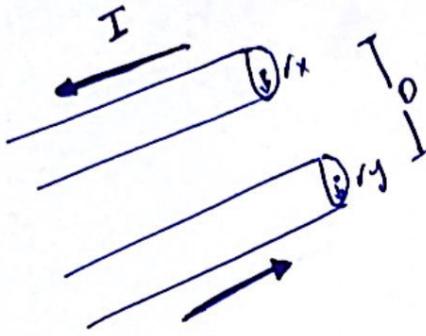
1 foot = 12 inch

⚡

TABLE 4.3 (175)  
4.1

**\* INDUCTANCE:**

- Single phase, two wire line (Solid)



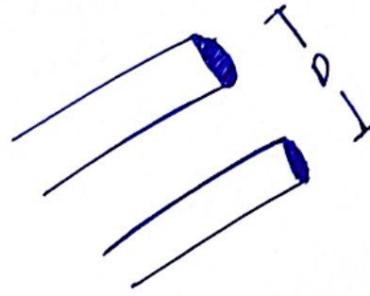
$$L_x = 2 \times 10^{-7} \ln \frac{D}{r_x'} \text{ H/m}$$

$$L_y = 2 \times 10^{-7} \ln \frac{D}{r_y'} \text{ H/m}$$

$$r_x' = e^{-1/4} r_x$$

$$r_y' = e^{-1/4} r_y$$

- Single phase stranded conductor:



$$L = 2 \times 10^{-7} \ln \frac{D}{\text{GMR}} \rightarrow \text{from table}$$

GMR: Geometric mean radius.

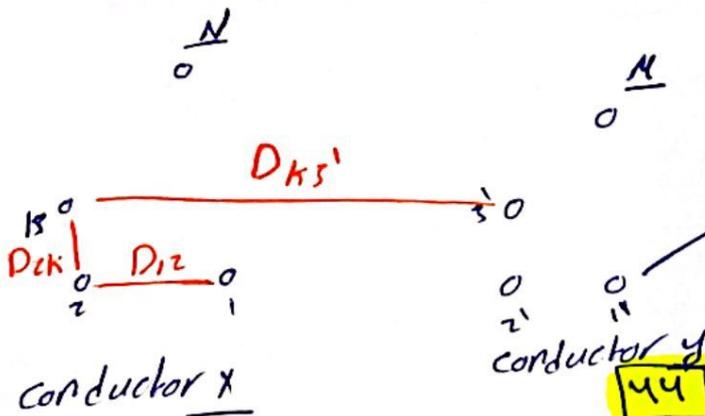
43

- ① Single phase Solid.
- ② Single phase stranded.
- ③ Single phase **Bundle**

(Same conductor sharing the same current).

- **Double circuit**: not electrically connected.

\* Single phase **Bundled** stranded conductor.



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$$L_x = 2 \times 10^{-7} \ln \left( \frac{D_{xy}}{D_{xx}} \right)$$

$$D_{xy} = \left( \prod_{k=1}^N \prod_{n=1}^N D_{kn} \right)^{1/2N}$$

$$\rightarrow \left( \begin{matrix} D_{11} & D_{12} & \dots & D_{1N} \\ D_{21} & D_{22} & \dots & D_{2N} \\ \dots & \dots & \dots & \dots \\ D_{N1} & D_{N2} & \dots & D_{NN} \end{matrix} \right)^{1/2N}$$

$$D_{xx} = \left( \prod_{k=1}^N \prod_{m=1}^N D_{km} \right)^{1/N^2}$$

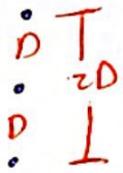
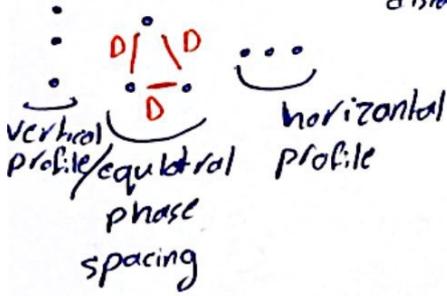
$$\rightarrow \left( \begin{matrix} D_{11} & D_{12} & \dots & D_{1N} \\ D_{21} & D_{22} & \dots & D_{2N} \\ \dots & \dots & \dots & \dots \\ D_{N1} & D_{N2} & \dots & D_{NN} \end{matrix} \right)^{1/N^2}$$

Solid  $\rightarrow r'$   
 stranded  $\rightarrow \text{GMR} \rightarrow \text{from table}$

$L = 2 \times 10^{-7} \ln\left(\frac{D}{r}\right) \Rightarrow$  Solid, single phase, not stranded, not bundled.

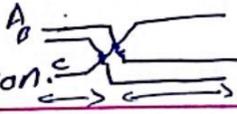
① 3φ (GMD)

$L = 2 \times 10^{-7} \ln\left(\frac{GMD}{r}\right)$  geometric mean distance



$\Rightarrow$  Inductance of 3φ cond. will not be equal.

$\Rightarrow$  Transposition.



$GMD = (D \times D \times 2D)^{\frac{1}{3}}$

② Stranded (GMR)

$L = 2 \times 10^{-7} \ln\left(\frac{D}{GMR}\right)$

Geometric mean radius.

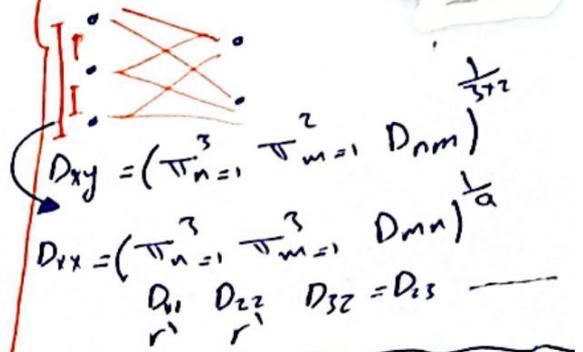
$D_{xx} = \left( \prod_{m=1}^{32} \prod_{n=1}^{32} D_{mn} \right)^{\frac{1}{32}}$

We never calculate GMR of stranded cond. we use table #



③ Bundled (D<sub>xy</sub>)

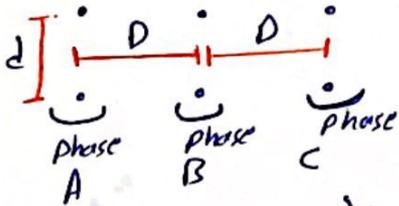
$L = 2 \times 10^{-7} \ln\left(\frac{D_{xy}}{D_{xx}}\right)$



④ 3φ, stranded  
 $L = 2 \times 10^{-7} \ln\left(\frac{GMD}{GMR}\right)$

⑤ 3φ, Bundled, solid  
 $L = 2 \times 10^{-7} \ln\left(\frac{GMD}{D_{xx}}\right)$

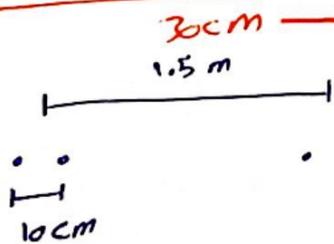
⑤



$GMD = (D \times D \times 2D)^{\frac{1}{3}} \quad D \gg d$

GMD =

⑥ 1φ, Bundled.



$D_{xy} = (1.5 \times 1.6 \times 1.4 \times 1.5)^{\frac{1}{4}} = 1.498 \approx 1.5 \text{ m}$

\* Capacitance of 1φ Solid transmission line:

$C_{xy} = \frac{\pi \epsilon}{\ln(D/r)} \text{ F/m}$  radius not r

$C_{xn} = 2 \times C_{xy}$

\* for 3φ

$C_{an} = \frac{2 \pi \epsilon}{\ln(D/r)} \text{ F/m}$  do not use r

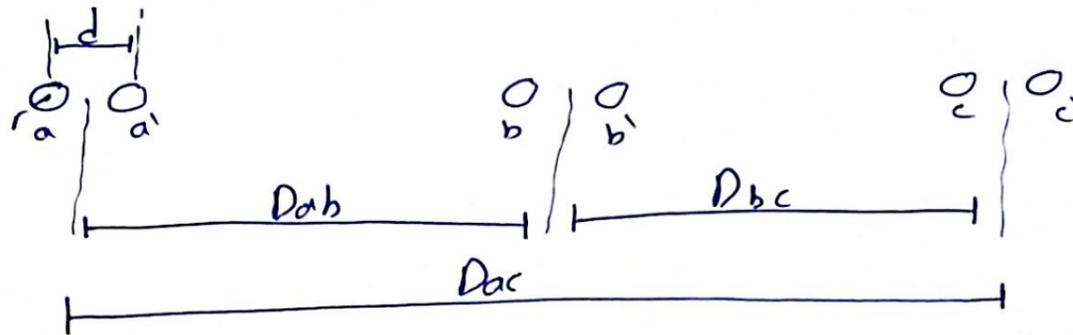
① Effect of stranding - uses GMR instead of r. (for inductance only).

② effect of bundling - uses GMD instead of D.

$D_{xy} = (30 \times 40 \times 20 \times 30)^{\frac{1}{4}} = 29 \text{ cm} \neq 30 \text{ cm}$

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\*  $d = 15 \text{ cm}$   
 &  $D_{ab} = D_{bc} = 3 \text{ m}$



-  $C_{an} = \frac{2\pi\epsilon}{\ln\left(\frac{GMD}{GMR}\right)}$  *use with bundling only.*

-  $GMD = (D_{ab} * D_{bc} * D_{ac})^{\frac{1}{3}}$

-  $GMR = (r * d)^{\frac{1}{2}} = \left( \frac{D_{aa}}{r} * \frac{D_{aa'}}{d} * \frac{D_{a'a}}{r} * \frac{D_{a'a'}}{d} \right)^{\frac{1}{4}}$