

تقدم لجنة EiCoM الاكاديمية

دفتر لمادة:

الات كهربائية (2)

جزيل الشكر للطالب:

عدنان حوراني





Three phase Transformers :-

All major power systems in the world are Three phase.

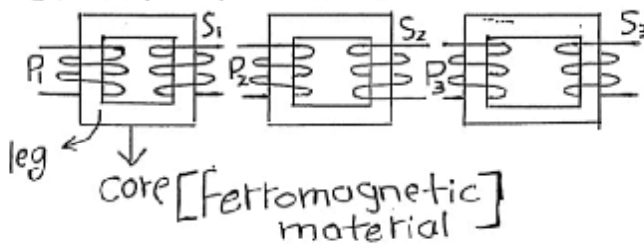
مكتبة خوارزمي
تصوير - قرطاسية - مكياجيات
٠٧٧٢٣٥٠٧٧٩

Three phase Transformers

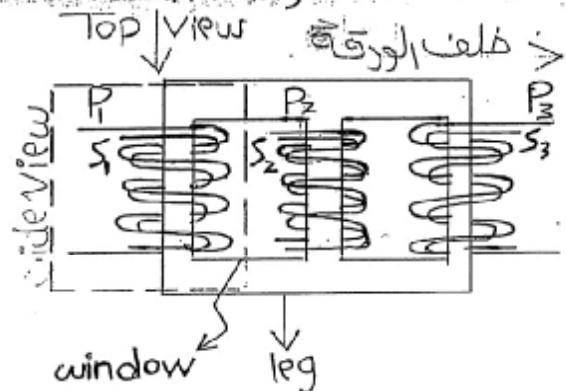
- * Smaller
- * lighter
- * Cheaper
- * Slightly more efficient

Three Single Phase Transformer.

each unit in the bank could be replaced individually in the event of trouble.



Single three phase Transformer.



μ_r (2000-6000).

لازم نعمل عزل بين
كل ϕ والأخرى.

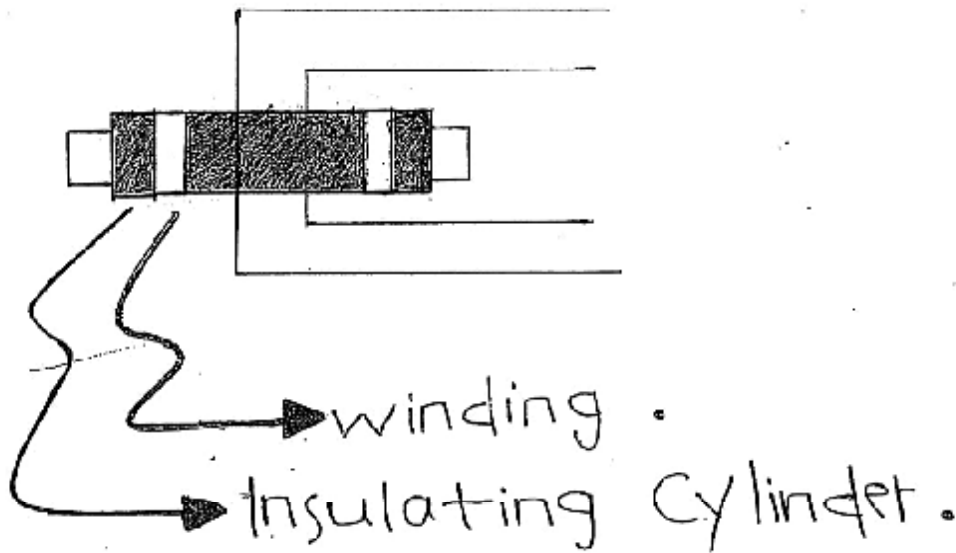
R very small. $[R = \frac{l}{\mu_r \mu_0 A}]$.

Flux flow through the path easily.

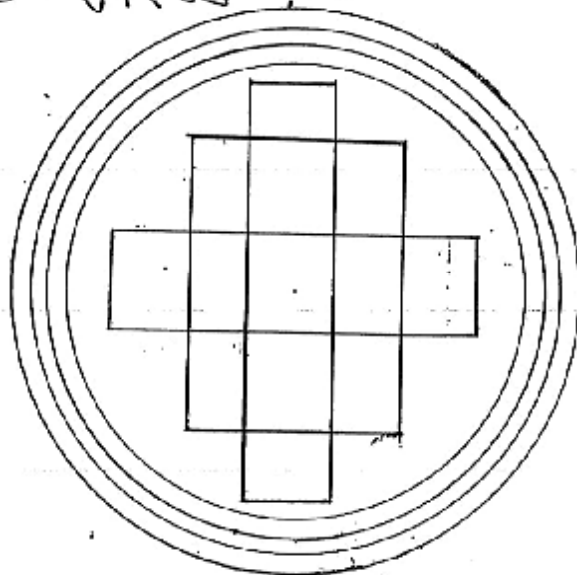
Energy ↓

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→ Side view :-



→ Top view :-



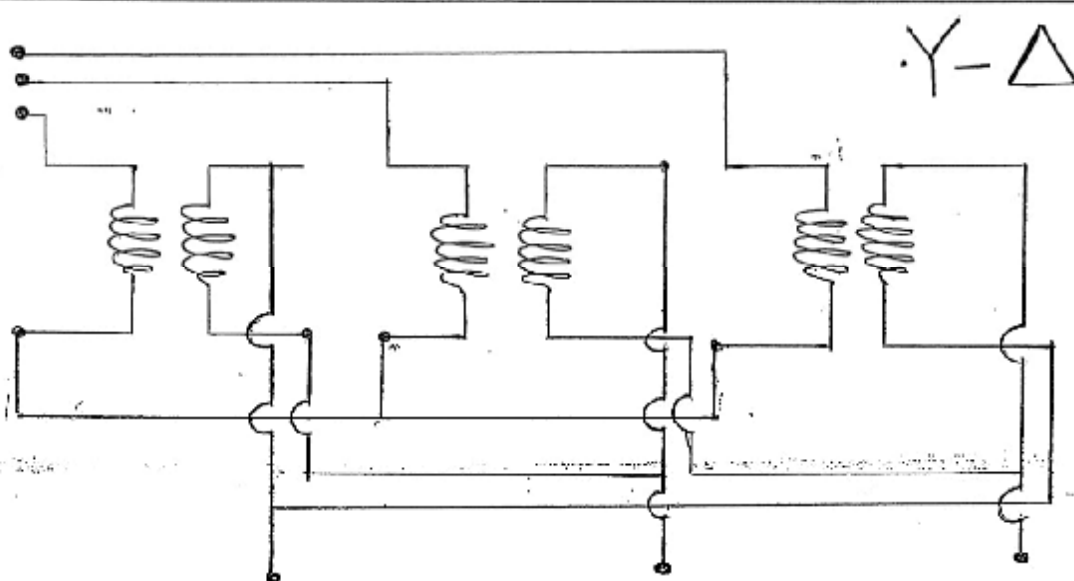
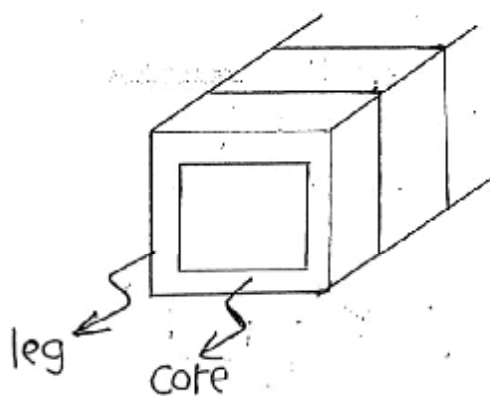
الدائرة الشكل الهندسي
التي لها أقل محيط عند
نفس المساحة لباقي
الأشكال الهندسية .

$$R = \frac{\rho L}{A}$$

Top view of the leg is very closed circular, because the circle has the lowest circumference compared with square and triangle .

note: lamination is
to reduce eddy
current losses.

↳ $\rho \uparrow$



note: discribed of step up and
step down based on phase voltages

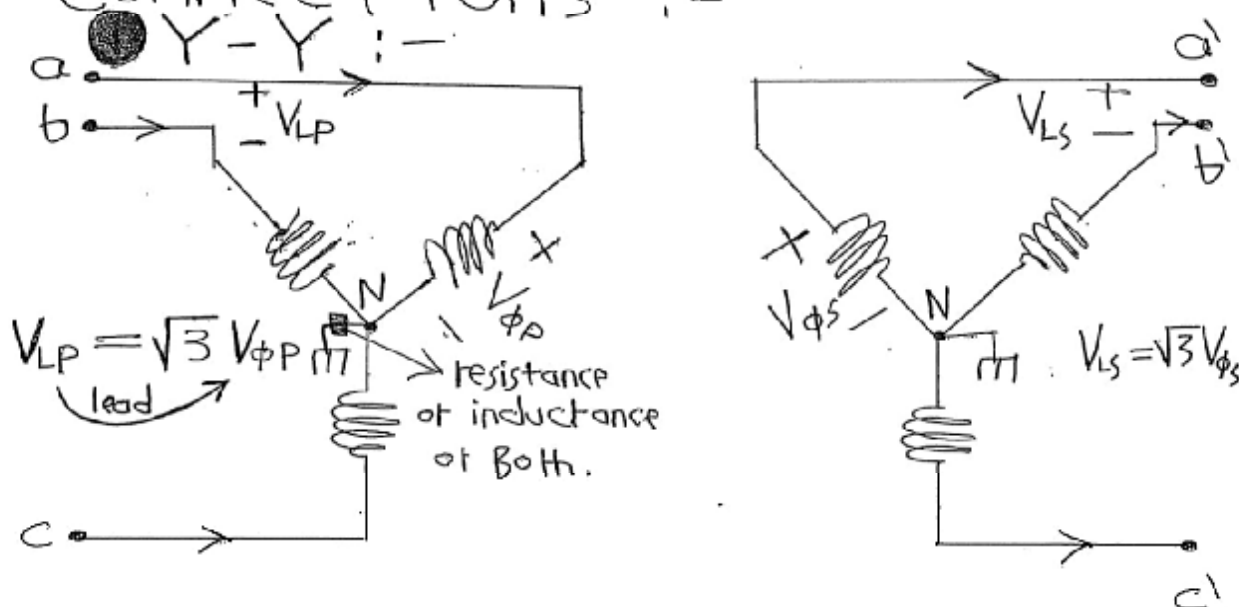


$$a = \frac{V_{P-\phi}}{V_{S-\phi}} = \frac{\# \text{ of terms } P.}{\# \text{ of terms } S.} = \frac{N_{P-\phi}}{N_{S-\phi}} = \text{turns ratio}$$

B

Three phase Transformer

Connections :-



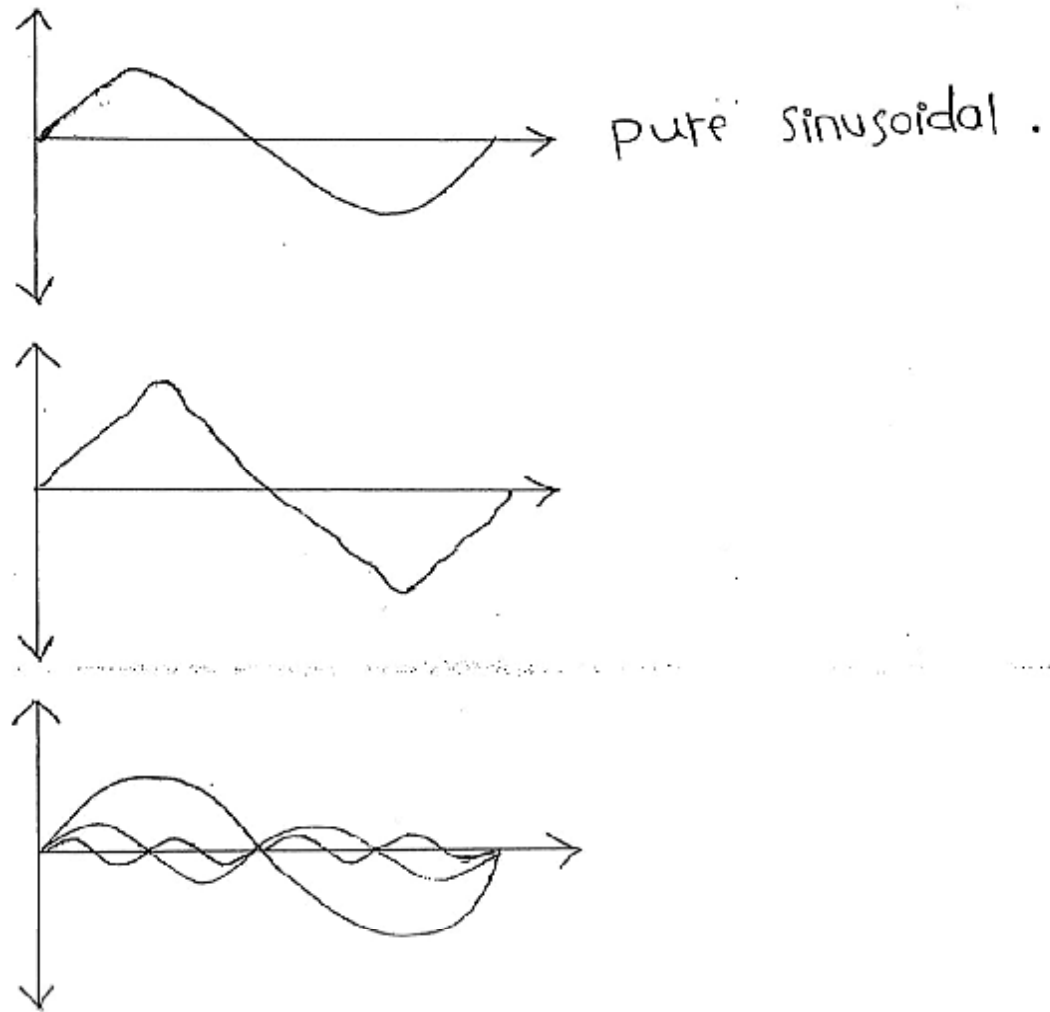
Y-Y connection has 2 problems:

1- unbalanced problem :-

- (Source unbalanced)
- (Load unbalanced)

if the load is unbalanced, then the voltages in the phases become largely unbalanced.

2- Third - harmonic Voltages :-



$$V_{3ha} = \hat{V}_{3h} \sin 3\omega t.$$

$$+ V_{3hb} = \hat{V}_{3h} \sin (3(\omega t - \frac{120^\circ}{360^\circ}))$$

$$+ V_{3hc} = \hat{V}_{3h} \sin (3(\omega t + \frac{120^\circ}{360^\circ}))$$

$$\hat{V}_{3h} \neq 0.$$

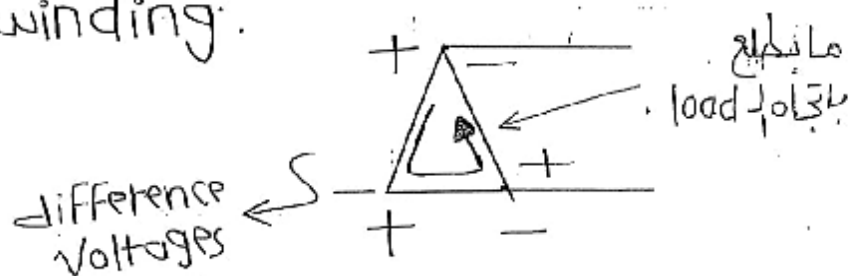
In
Phase

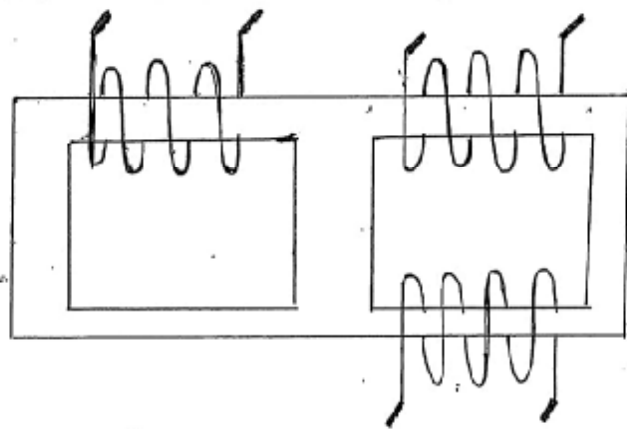
* due to nonlinearity of the material of the core, there are 3rd harmonic components in the phase voltages, which are in phase.

→ Solution of 2 problems:

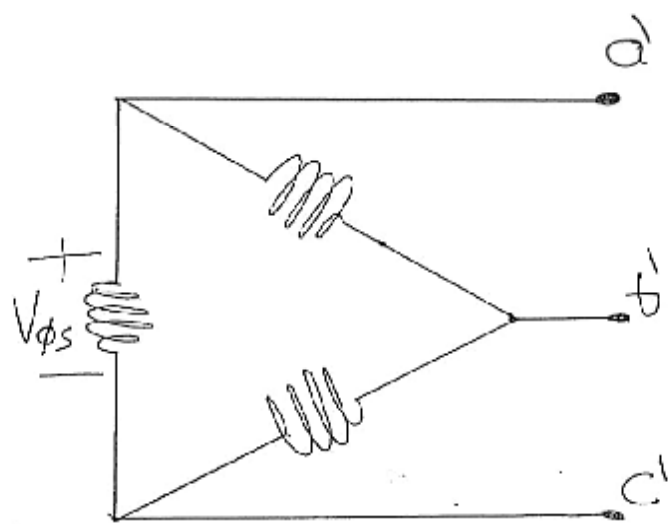
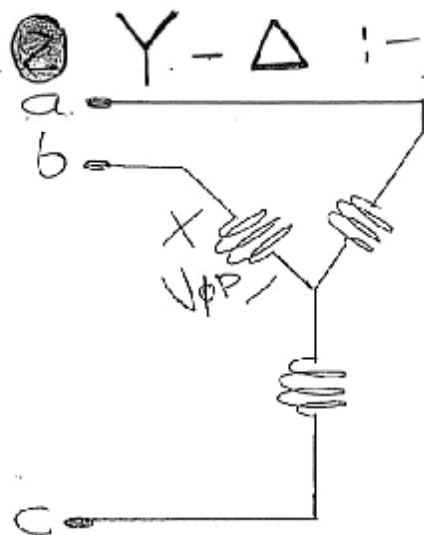
1- Solidly grounding the neutral points. (with zero impedance).

2- Add third (tertiary) winding connected in delta, where the third harmonic components circulate inside this winding.





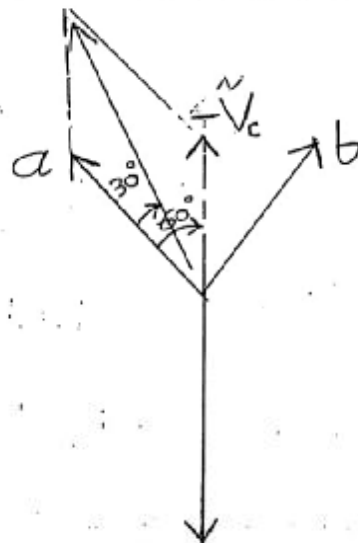
tertiary winding is used to supply light loads like, the lights of substation or the cooling fans.



$$a = \frac{V_{\phi P}}{V_{\phi S}} = \frac{V_{LP}/\sqrt{3}}{V_{LS}} \rightarrow \frac{V_{LP}}{V_{LS}} = a\sqrt{3}$$

This connection has No problem with third harmonic component due to the Δ -connection of the Secondary.

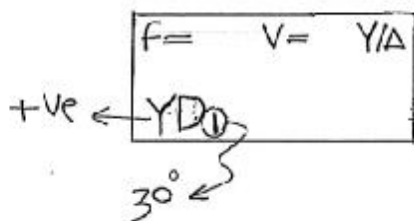
abc +ve
acb -ve



$$\tilde{V}_{ac} = \tilde{V}_a - \tilde{V}_c$$

$\therefore V_{ac}$ lags V_a by 30° .

Group Connection



-ve seq. :-
 V_{ac} leads V_a by 30° .

YD0
330° lagging

The Group Connection is important if the Transformers are to be parallel.

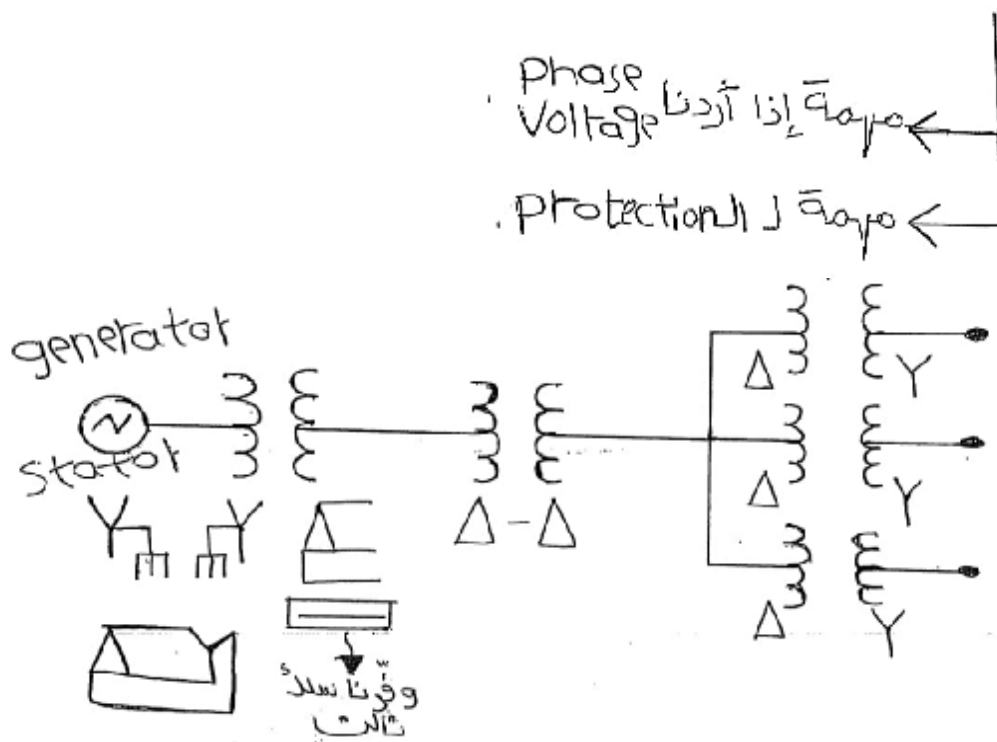
The Phase angles of the transformers Secondary voltage must be equal.

note:

DC \rightarrow Current if difference Voltage.

AC \rightarrow Current if difference Voltage
or difference phase
or both .

This connection has no unbalanced or Third harmonic problems, but has no Neutral point.



* The Per-Unit System of
3 ϕ Transformers :-

S_{base} = total Voltampere Value of
transformer Bank.
MVA/KVA/VA

$$S_{1-\phi base} = \frac{S_{base}}{3}$$

$$I_{1-\phi base} = \frac{S_{1-\phi base}}{V_{1-\phi base}} = \frac{S_{base}}{3V_{1-\phi, base}}$$

$$Z_{base} = \frac{V_{1-\phi, base}}{I_{1-\phi, base}} = \frac{(V_{1-\phi, base})^2}{S_{1-\phi, base}} = \frac{3(V_{1-\phi base})^2}{S_{base}}$$

Voltage current Power Δ Y ϕ phase quantity ϕ على الج ϕ على الج * ϕ phase quantity

$$V_{L, base} = V_{1-\phi, base} \quad \Delta$$

$$V_{L, base} = \sqrt{3} V_{1-\phi, base} \quad Y$$

Example :- 50 kVA, 13800 / 208V
rated 3- ϕ apparent power. line voltages

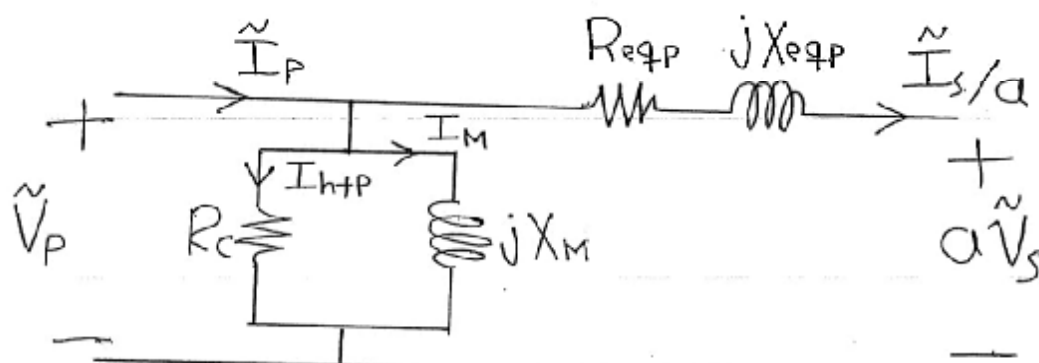
Δ -Y distribution transformer

$$R = 1\% = 0.01 \text{ pu} \quad \left\{ \begin{array}{l} \text{for both} \\ \text{sides.} \end{array} \right.$$

$$X = 7\% = 0.07 \text{ pu}$$

@ What is the phase impedance referred to primary (high voltage) side. ?

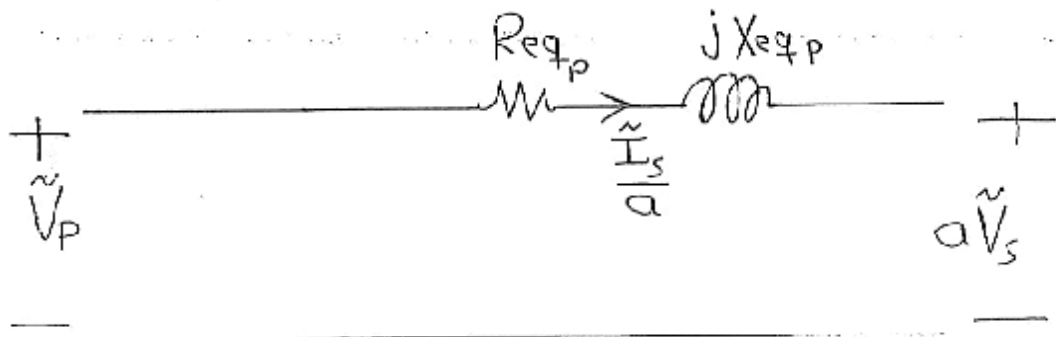
$$\tilde{Z}_{eqp} = \underbrace{R_{eqp}}_{R_1 + a^2 R_2} + j \underbrace{X_{eqp}}_{X_1 + a^2 X_2} \rightarrow \text{actual values ?}$$



Per-phase equ. cct. of Transformer referred to primary side.

$$Z_{\text{base}} = \frac{3(V_{1-\phi, \text{base}})^2}{S_{\text{base}}} \\ = \frac{3(13800)^2}{50k} = 11426 \Omega$$

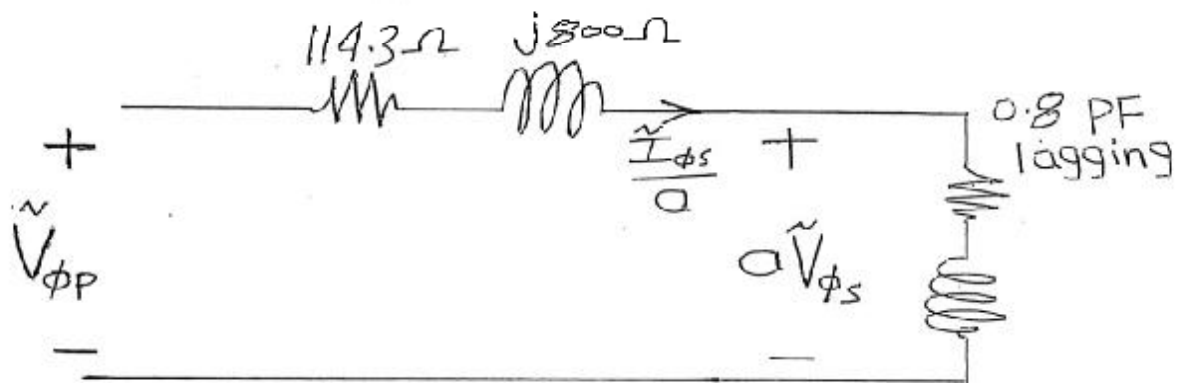
$$\tilde{Z}_{\text{eq}} = 0.01 + j0.07 \text{ pu.} \\ \text{actual value} \quad (\text{per unit}) \quad (\text{base}) \\ \tilde{Z}_{\text{eqp}} = (0.01 + j0.07)(11426) \\ = 114.26 + j800 \Omega$$



R_c and $X_m \gg R_{\text{eq}}$ and X_{eq} .

∞ neglected. \neq

⑥ V_R % at full load and 0.8 PF lagging using actual values.



* Per-phase equ. cct. referred to primary

$$V_R \% = \frac{|\tilde{V}_{\phi P}| - |a\tilde{V}_{\phi S}|}{|a\tilde{V}_{\phi S}|} * 100\%$$

$\leftarrow |a\tilde{V}_{\phi S, \text{no-load}}|$

$$\tilde{V}_{\phi P} = a\tilde{V}_{\phi S} + \left(\frac{\tilde{I}_{\phi S}}{a}\right) [114.3 + j8000]$$

$$a = \frac{V_{\phi P}}{V_{\phi S}} = \frac{13800}{\left(\frac{208}{\sqrt{3}}\right)} = 114.8$$

$$S = \sqrt{3} V_L I_L$$

$$50 * 10^3 = \sqrt{3} (208) \overset{\text{rated}}{I_{\phi S}}$$

$$\rightarrow I_{\phi S} = 139 \text{ A.} \rightarrow \text{at full load.}$$

$$\rightarrow \text{at } \frac{1}{2} \text{ load} \rightarrow \frac{I_{\phi S, \text{rated}}}{2} = 69.5 \text{ A}$$

$$\rightarrow \text{at no load} \rightarrow I_{\phi S} = 0$$

6

$$\tilde{V}_{\phi S} = \frac{208}{\sqrt{3}} \angle 0. \rightarrow \text{reference, دائماً حيزه } \nearrow \cos^{-1} 0.8$$

$$\tilde{V}_{\phi P} = (114.8) \left(\frac{208}{\sqrt{3}} \right) \angle 0 + \left(\frac{139 \angle 0 - \cos^{-1} 0.8}{114.8} \right) *$$

$$= 14500 \angle 2.73^\circ \text{ V.} \quad [114.3 + j800]$$

$$\rightarrow V_R \% = \frac{14506 - 13800}{13800} * 100$$

$$= 5.1\%$$

© Repeat part (b) using per-unit.

$$\tilde{V}_P = \tilde{V}_S + \tilde{I}_S (R + jX)$$

$$= 1 \angle 0 + \left(1 \angle -\cos^{-1} 0.8 \right) (0.01 + j0.07)$$

↑
at full load
* if $\frac{1}{2}$ load $\rightarrow \frac{1}{2} \angle \cos^{-1} 0.8$

$$= 1.051 \angle 2.73 \text{ pu.}$$

$$V_R \% = \frac{1.051 - 1.0}{1.0} * 100 \%$$

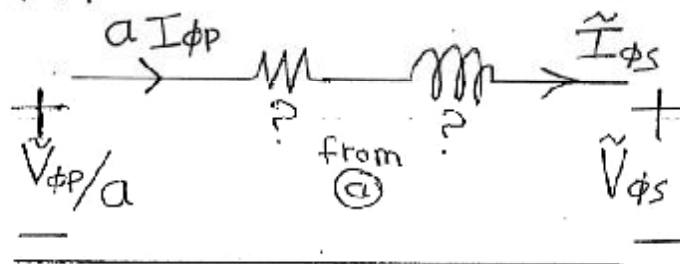
$$= 5.1\%$$

* Advantages of per-unit :-

- 1- Reflected impedance has no meaning.
- 2- line and phase Voltages has no meaning.
- 3- Square root of 3 has no meaning.
- 4- Δ and Y connections has no meaning.
- 5- more sense in numbers.

• Simplifying Analysis.

! [b] if referred to Secondary :



→ Y-connection → $V_{LL} = \sqrt{3} V_{\phi}$
 $I_L = I_{\phi}$

→ Δ -connection → $V_{LL} = V_{\phi}$
 $I_L = \sqrt{3} I_{\phi}$

Example :- 100 MVA, 230/115 kV,

Δ - Δ , 3- ϕ , power transformer

$$R = 0.02 \text{ pu.}$$

$$X = 0.055 \text{ pu.}$$

$$R_c = 120 \text{ pu.}$$

$$X_M = 18 \text{ pu.}$$

□ if the transformer supplies

80 MW at 0.85 P.F lagging load.

draw the phasor diagram in pu?

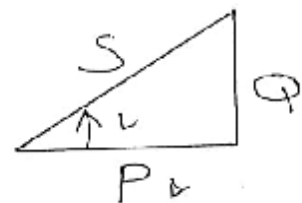
$$S_{\text{rated}} = S_{\text{base}} = 100 * 10^6 \text{ VA.}$$

$$P = \frac{80}{100} = 0.8 \text{ pu.}$$

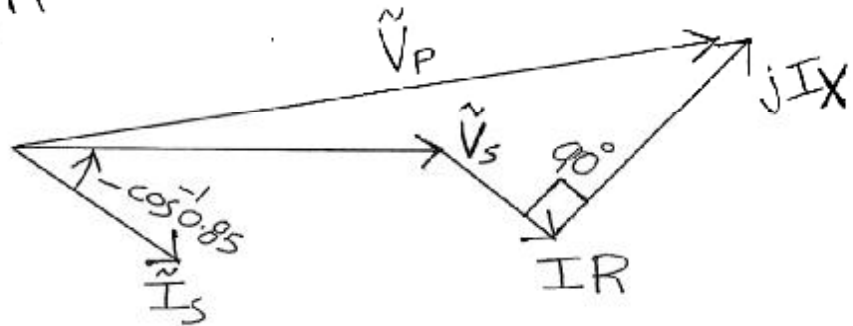
$$P = VI \cos \phi \text{ pu.}$$

$$0.8 = (1) I (0.85)$$

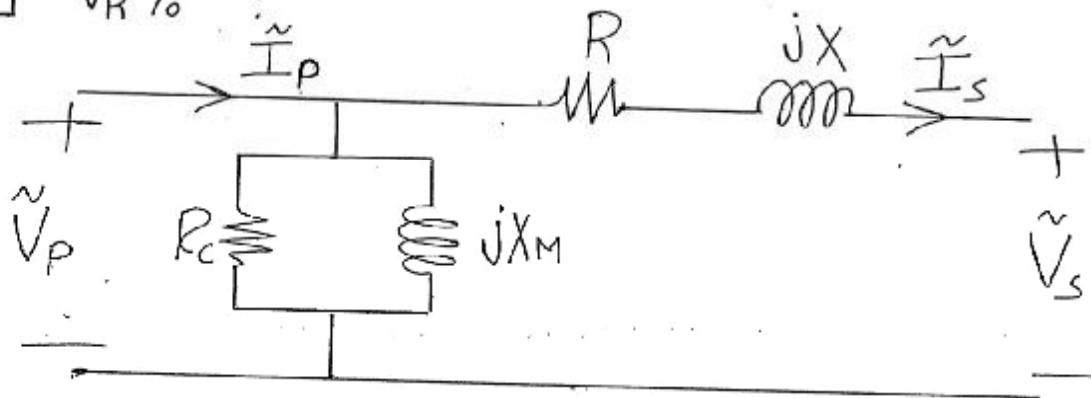
$$\rightarrow I = 0.94 \text{ pu.}$$



“phaset diagram”



b) $V_R \% ?$



$$\tilde{V}_P = \tilde{V}_S + \tilde{I} (0.02 + j0.055)$$

$$= 1 \angle 0 + (0.94 \angle -\cos^{-1} 0.85) (0.02 + j0.055)$$

$$= 1.0476 \text{ pu.}$$

$$V_R \% = 4.76 \% .$$



X_m, R_c ليس لهما
علاقة بحسابان V_R
ولهما علاقة بحسابان I .

c) %?

$$\% = \frac{P_{out}}{P_{in}} * 100$$

$$= \frac{0.8}{0.8 + P_{losses}} * 100$$

$$= \frac{0.8}{0.8 + |\tilde{I}_s|^2 R + \frac{1V_{pl}^2}{R_c}} * 100\%$$

$$= 96.75\%$$

d) actual impedance values referred to primary?

→ Z_{base} at primary side.

$$Z_{base} = \frac{V_{\phi, base}}{I_{\phi, base}} \Big|_{\text{primary}}$$

$$V_{\phi, base} \Big|_{\text{primary}} = 230 * 10^3 V$$

$$S = \sqrt{3} V_L I_L$$

$$100 * 10^6 = \sqrt{3} (230 * 10^3) I_L \rightarrow I_L = 251.02 A_3$$

$$I_{\phi, \text{base}} = \frac{I_L}{\sqrt{3}} \\ = \frac{251.02}{\sqrt{3}} = 145 \text{ A.}$$

$$Z_{\text{base}} = \frac{230 \times 10^3}{145} = 1586.2 \Omega.$$

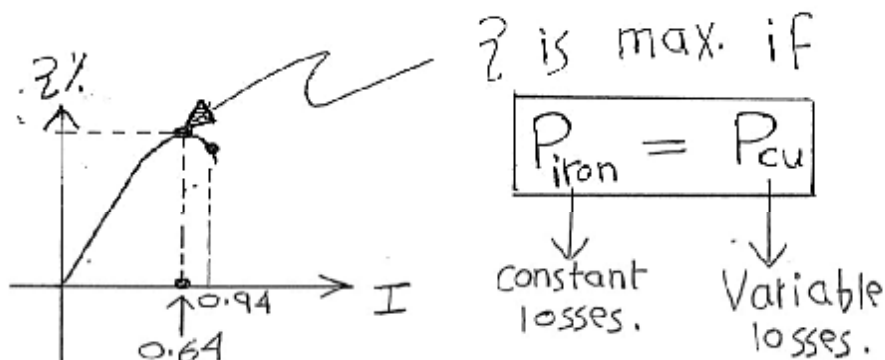
$$R_{\text{eq}, p} = (0.02)(1586.2 \Omega) = \underline{\underline{31.7 \Omega.}}$$

$$X_{\text{eq}, p} = (0.055)(1586.2 \Omega) = \underline{\underline{87.2 \Omega.}}$$

$$R_c = (120)(1586.2 \Omega) = \underline{\underline{190.3 \text{ k}\Omega.}}$$

$$X_M = (18)(1586.2 \Omega) = \underline{\underline{28.5 \text{ k}\Omega.}}$$

e) Calculate the load when the Transformer run at the max. η condition ?



$$\rightarrow P_{iron} = P_{cu}$$

$$\frac{|V_p|^2}{R_c} = |\tilde{I}|^2 (R_{eq})$$

$$\frac{(1.0476)^2}{120} = |\tilde{I}|^2 (0.022)$$

$$\rightarrow I = 0.64 \text{ A}$$

$$\eta\% = \frac{0.8}{0.8 + (2) \left(\frac{(1.0476)^2}{120} \right)} * 100\%$$

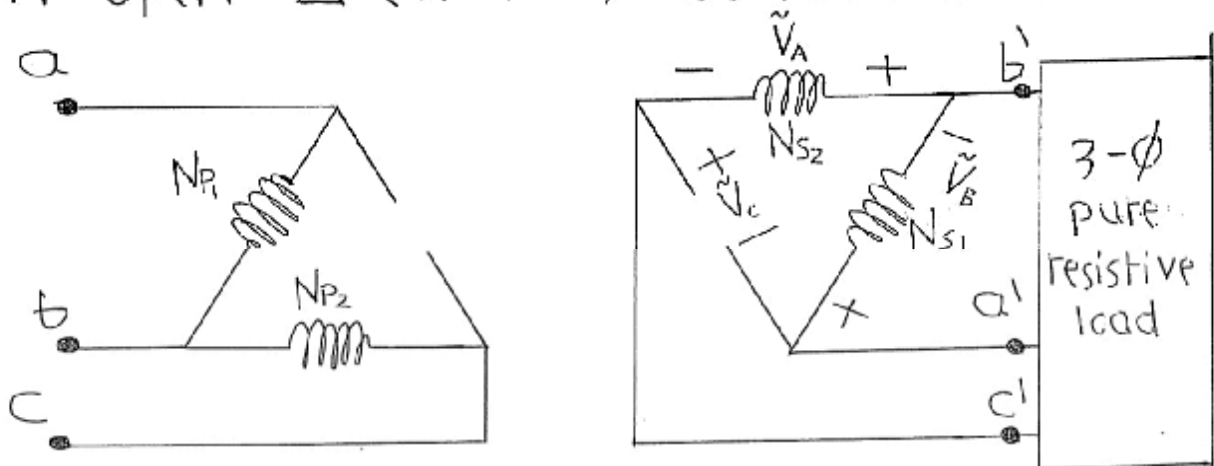
$$= 97.7\%$$

جينا قيمه load
اللي بتضمنه
و بتضمنه max η

Three-Phase Transformations using Two Transformers :-

- 1] Open - Δ (or V-V) Connection.
- 2] Open - Y - Open - Δ Connection.
- 3] Scott + T Connection.
- 4] The Three-phase T Connection.

Open - Δ (or V-V) Connection :-



$$-\tilde{V}_c - \tilde{V}_A - \tilde{V}_B = 0$$

$$\tilde{V}_c = -\tilde{V}_A - \tilde{V}_B = V \angle 120^\circ (V) \rightarrow \text{balanced as Voltage, but not balanced as current.}$$

$$P_1 = \frac{\sqrt{3}}{2} V_\phi I_\phi$$

$$P_2 = V_\phi I_\phi \cos 30^\circ = \frac{\sqrt{3}}{2} V_\phi I_\phi$$

$$P_1 + P_2 = \sqrt{3} V_\phi I_\phi$$

if System is healthy :

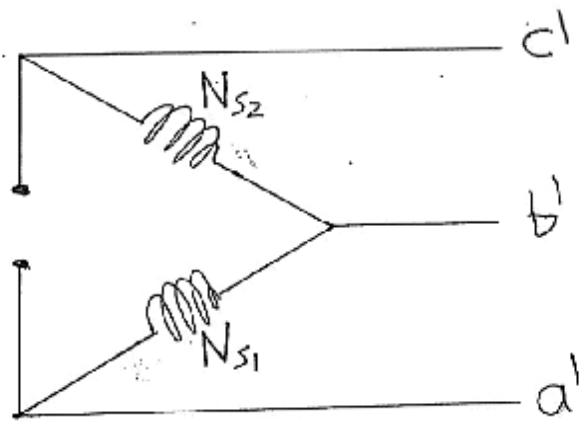
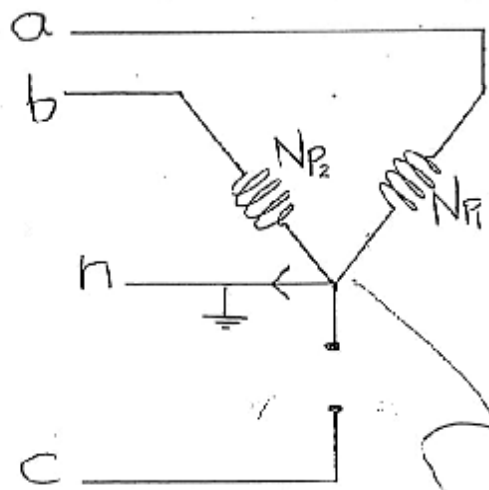
$$P = \sqrt{3} V_L I_L \\ = 3 V_\phi I_\phi$$

$$\frac{P}{P_1 + P_2} = \frac{3 V_\phi I_\phi}{\sqrt{3} V_\phi I_\phi} = \sqrt{3}$$

$$P_1 + P_2 = \frac{1}{\sqrt{3}} P \\ \uparrow \\ 57.7\%$$

→ This connection is used when one of the phases needs maintenance, And if we have 2 Transformers and we want to connect them as 3- ϕ system.

1 # Open - Y - Open - Δ Connection :



The main difference between this connection and the previous one is : The current will not be equal zero.

→ The 3- ϕ Voltages are balanced.

→ The 3- ϕ Currents are not balanced.

$$\rightarrow P_1 + P_2 = \frac{1}{\sqrt{3}} P$$

↑
57.7 %

→ The same usage of the Open- Δ Connection

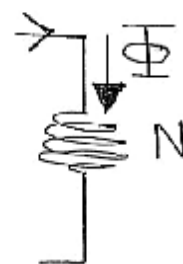
Transformer Ratings and Related Problems:

S, V, I, and f \rightarrow Basics.

* The Voltage and-freq. Ratings of a Transformer :-

$$E_{rms} = 4.44 \Phi f N$$

↑
of turns



$$E_{rms_1} = 4.44 \Phi_1 f_1 N$$

$$E_{rms_2} = 4.44 \Phi_2 f_2 N$$

assuming $E_{rms_1} = E_{rms_2}$:

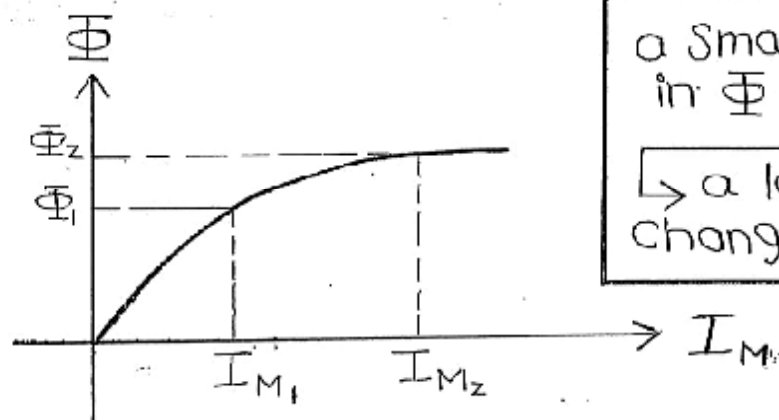
$$\Phi_1 f_1 = \Phi_2 f_2$$

$$\rightarrow \Phi_2 = \frac{f_1}{f_2} \Phi_1$$

$f_1 = 60 \text{ Hz}$ (American Sys.)

$f_2 = 50 \text{ Hz}$ (Europe Sys.)

$$\Phi_2 = 1.2 \Phi_1$$



$$\underline{\underline{I_{M2} \gg I_{M1}}}$$

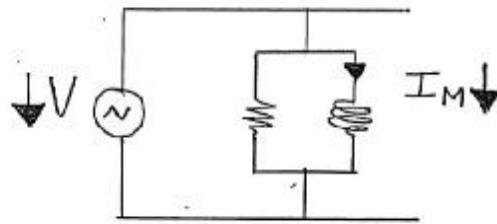
→ Change from high freq. to low freq. is very dangerous, because the saturation of the magnetic cct.

→ Change from low freq. to high freq. is not dangerous.

→ To overcome this problem, derate the apply voltage with the same ratio of the two frequencies.

$$E_{rms2} = \frac{50}{60} E_{rms1}$$

→ We have to reduce the voltage on the primary winding in order to reduce I_M .



* The apparent power and Current Ratings of a Transformer :

→ Transformer with 100 MVA.

↑
rated: max. power that the trans. can produce.

→ Most loads are RL loads.

$$\hat{S} = P + jQ$$

if $P.F._{load} = 1$.
 → (pure resistive).

$$S = P = \sqrt{3} V_L I_L$$

↑
const.

→ (pure inductive),
 → (pure capacitive).

$$Q = S = \sqrt{3} V_L I_L$$

↑
const.

→ load
واحد ع
Trans.

↳ نظر به حال current → the value of the current indicates the kind of the load (full load → over load, ...).



→ half load → half current.

→ full load → full current.

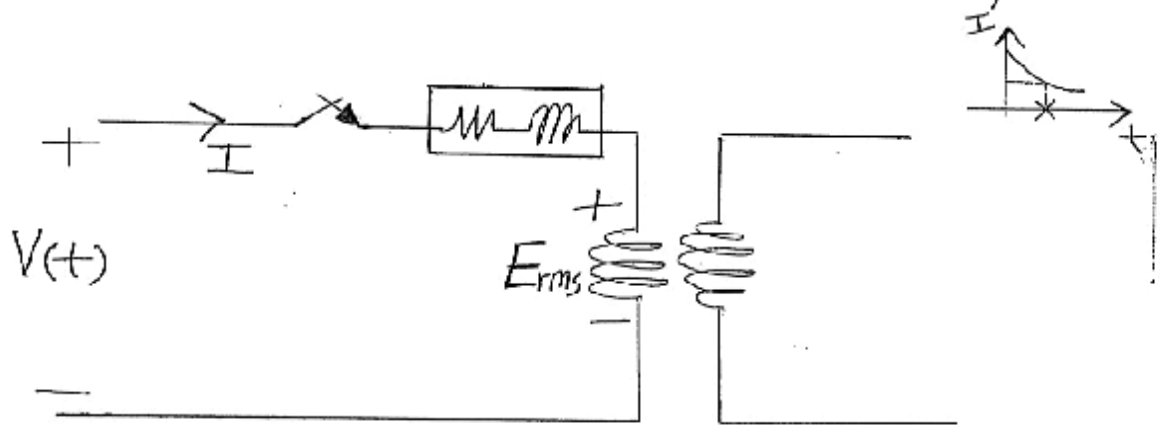
→ load ↑ → current consumed ↑,

$P_{\text{losses}} (I^2 R) \uparrow$, $P_{\text{series impedance}} \uparrow$, heat ↑,

→ then cooling is needed.

* Inrush Current :-

the current drawn by the transformer at starting. (5-7 times load current).



$$I = \frac{V - E_{rms}}{R + jX}$$

→ At the moment of the switching there is no E_{rms} , so there is no Φ , and the current is very large.

No current → No Φ → No E_{rms} .

→ Inrush current depends on:

$$V(t) = V \cos(\omega t + \phi)$$

↳ depends on the instant of switching.

the ϕ shift of the voltage.

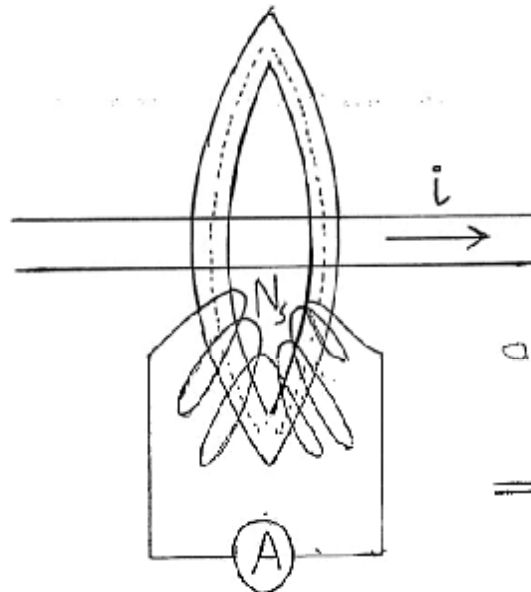
Instrument Transformers

V.T

- 1- Step down.
- 2- accurate.
- 3- Very low power rating (Small Size).
- 4- Used for measurements.

C.T

- 1- Step down.
- 2- accurate.
- 3- Very low power rating (Small Size)
- 4- Used for measurements



Very high current and very high voltage,
 ⇒ من الصعب تركيب
 (A)

لذلك يتم استخدام VT و CT

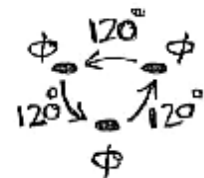
* V.T \rightarrow Short cct. on Secondary \rightarrow Very dangerous.

* C.T \rightarrow Open cct. on Secondary \rightarrow Very dangerous:

Notes On AC Machines :-

The effect of coil pitch on AC machines
Stators :-

→ phase shift between 3ϕ in space, gives 120°
phase shift in time.



* In General, the air gap flux density distribution is never pure sinusoidal.

Because :-

- 1] Saturation of ferromagnetic material.
- 2] Unbalanced load, Stator Υ , non grounded.

↳ reflection on flux.

↳ never pure sinusoidal.

→ To reduce the unwanted harmonics
 Several techniques are used of these:
 techniques, "fraction pitch winding"

↑
 effect of $\frac{1}{3}$ from $\frac{1}{3}$ of the 3rd harmonic

The pitch of the pole :-

The angular distance between two adjacent poles on a machine.

$$\rho_p = \frac{360^\circ}{P} \quad (\text{mechanical degrees}).$$

The pitch
 of the
 pole.

#. of poles.

* Regardless the #. of poles on the machine, the pole pitch is always equals 180° electrical.

ex. :- $P = 4$:-

→ mechanical : $\rho_p = \frac{360^\circ}{4} = 90^\circ$.

→ electrical : always = 180° .

* Full pitch coil :- the Stator Coil Spans (extends) across the same angle of the Pole pitch.

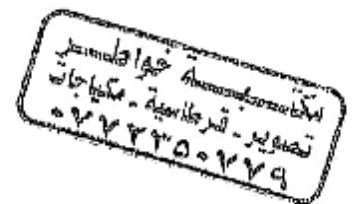
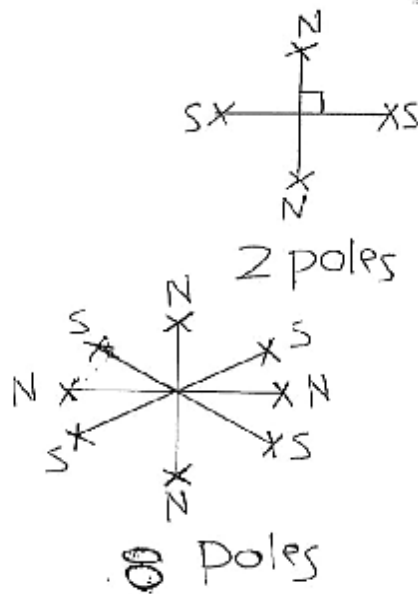
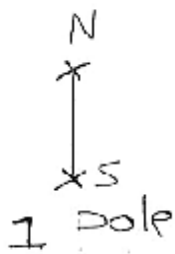
* Fractional pitch coil : the Stator Coil Spans across an angle less than the pole pitch.

ex \rightarrow (coil pitch is $\frac{5}{6}$ of the pole pitch).

in respect of the #. of poles : ($p=2$)

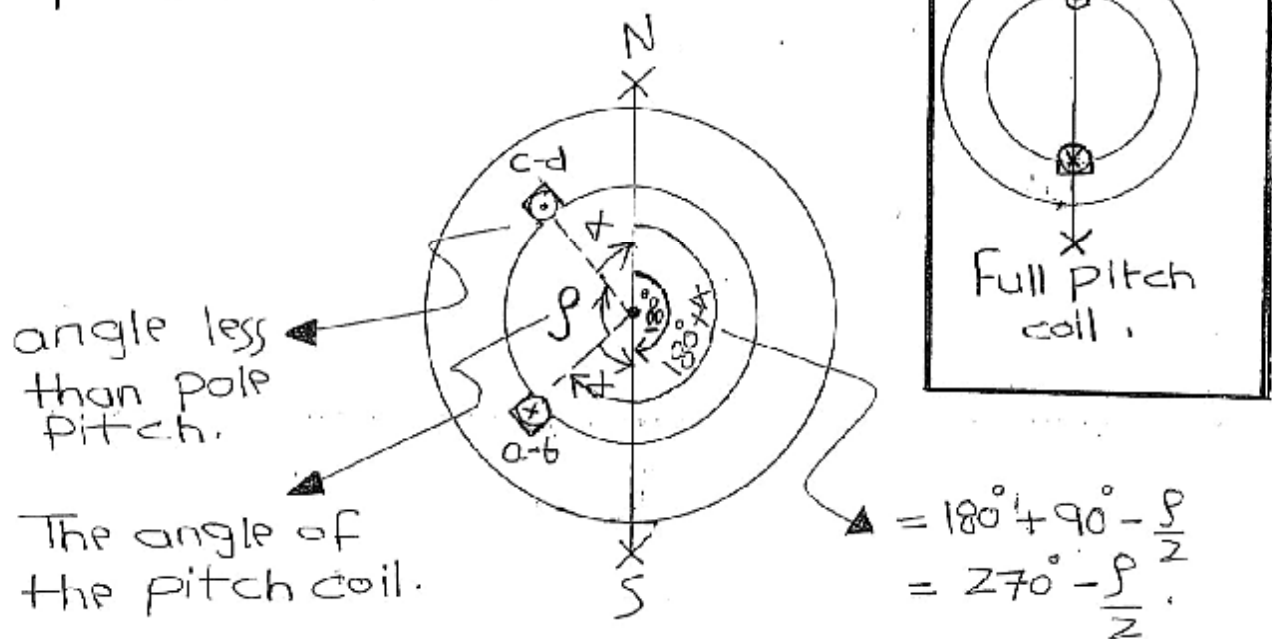
$$\rightarrow \rho_p = \frac{5}{6} * 180^\circ \text{ electrical} = 150$$

$$\rightarrow \rho_p = 150 * \# \text{ of poles} = 150 * 2 = 300 \text{ mechanical.}$$



* Windings with fractional pitch coils are known as Chorded windings.

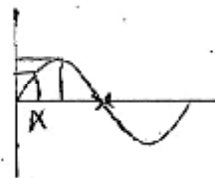
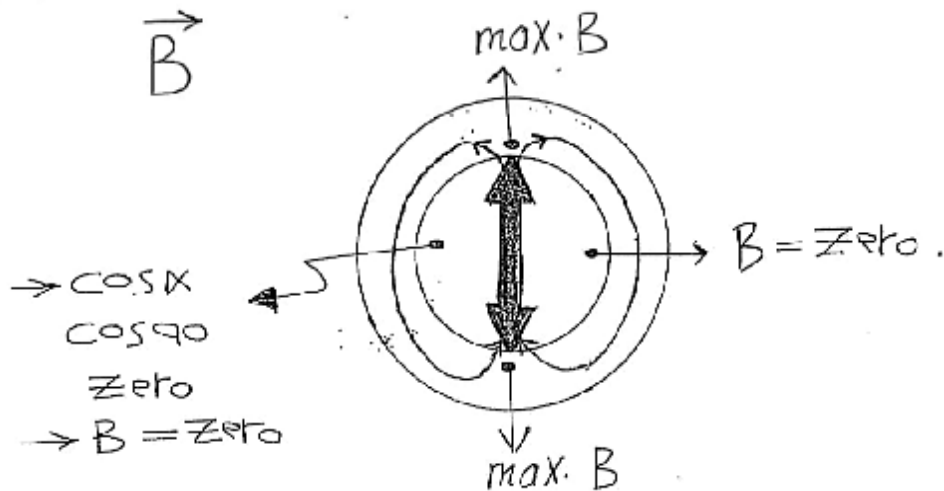
* The Induced Voltage of a fractional Pitch coil :-



$$p + 2\alpha = 180^\circ \rightarrow \alpha = 90^\circ - \frac{p}{2}$$

→ Flux density (from rotor to Stator)

\vec{B}



$$B = B_M \cos \alpha$$

$$B = B_M \cos (\omega t - \alpha) \text{ ----- Travelling wave.}$$

↑
time

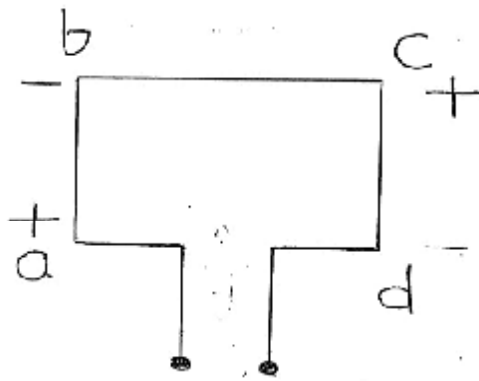
$$e_{\text{ind}} = (\vec{v} \times \vec{B}) \cdot \vec{l}$$

↑ Voltage induced in any coil

↑ velocity of the conductor relative to the magnetic field

↑ magnetic flux density.

→ length of the wire. (with current direction).



Segment ab :

$$\begin{aligned}
 e_{ba} &= (\vec{v} \times \vec{B}) \cdot \vec{I} = VB l = VB_M \cos \alpha l \\
 &= VB_M \cos(\omega t - \alpha) l \\
 &= VB_M \cos(\omega t - (90 - \frac{\theta}{2})) l
 \end{aligned}$$

Segment bc and ad :-

$$e_{cb} = e_{da} = 0 \quad , \quad \vec{v} \times \vec{B} = 0 \quad , \quad (\vec{v} \perp \vec{B})$$

Segment cd :-

$$e_{dc} = VB_M \cos(\omega t - (90 - \frac{\theta}{2})) l$$

$$\rightarrow e_{ind} = e_{ba} + e_{cd}$$

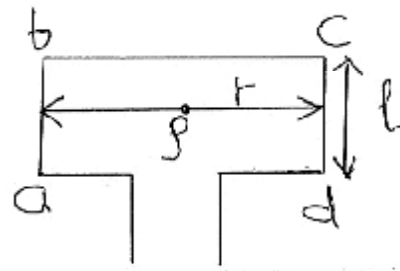
$$= 2VB_M \cos(\omega t - (90 - \frac{\rho}{2})) l$$



$$e_{ind} = 2 \underline{V} B_M l \cos \omega t \sin \frac{\rho}{2}$$

$$* V = r\omega$$

$$* \phi = Brl$$



$$\Rightarrow E_{ind} = 4.44 \phi f N k_p$$

$$k_p = \sin \frac{\rho}{2}$$

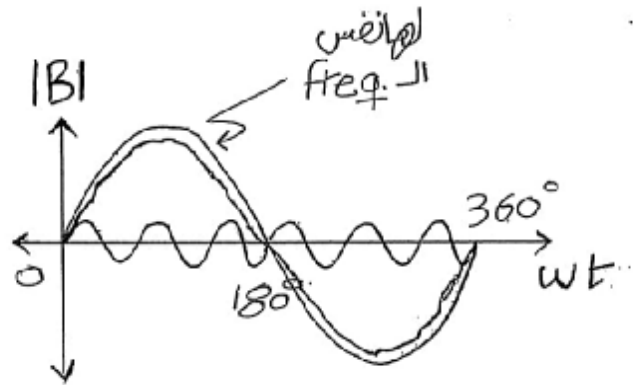
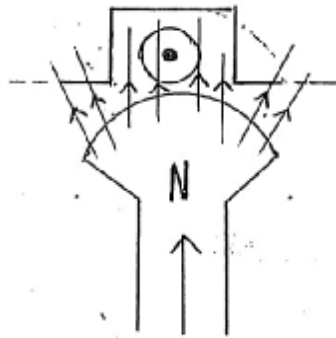
↑
coil pitch factor.

if $k_p = 0.9 \rightarrow$ this will

reduce the induced voltage (disadvantage).



Harmonic problems and fractional pitch windings :



* The waveform of Flux follow the waveform of voltage.

* Voltage lags flux by 90° .

$$e_{ind} = -N \frac{d\Phi}{dt}$$

الفريز الـ Φ هو سبب
induced
voltage.

ظفر في البداية الـ Φ وبعدها بـ 90° الـ (V) ،
إنا حار في Load (يكني وصلة C, L, R) بيه
يصير في I ، والـ I بهل Φ اتجاهه يتكاس
رتباه الـ Φ الأمان.

at no load \rightarrow no current. \rightarrow only one flux.

Only odd harmonics exist in the waveform of the induced Voltages.
 why? because Signals are Symmetric on X-axis.
 $\rightarrow 3^{\text{rd}}, 5^{\text{th}}, 7^{\text{th}}, \dots$

The 3^{rd} harmonic components and its multiples do not appear in the O/P

$$3 \times 2 = 6 \times$$

$$3 \times 3 = 9 \times$$

$$3 \times 4 = 12 \times$$

$\Rightarrow 7^{\text{th}}, 5^{\text{th}}$ تصبح ال 5th و ال 7th تظهر
 لا زبق تقاليم او تلغيره

Voltage in case of Solidly grounded

Υ and Δ Stator windings.

Only $5^{\text{th}}, 7^{\text{th}}, 11^{\text{th}}, 13^{\text{th}}, \dots$ harmonic components are there.

* [As the index of harmonic increases, the magnitude decreases.]

* The Pitch factor of the Coil at the harmonic frequency is: $k_p = \sin \frac{\gamma}{2}$
 where γ : index of the harmonic to be examined.

Example :- 3ϕ - 2 Poles. Stator has Coils with a $\frac{5}{6}$ pitch, what are the pitch factors for the harmonics presented in the Coils of the machine?

$$k_{p_1} = \sin \frac{\rho}{2} \Rightarrow \text{Fundamental (first harm. component)}$$

$$k_{p_3} = \sin \frac{3\rho}{2} \Rightarrow 3^{\text{rd}}$$

$$k_{p_5} = \sin \frac{5\rho}{2} \Rightarrow 5^{\text{th}}$$

$$k_{p_7} = \sin \frac{7\rho}{2} \Rightarrow 7^{\text{th}}$$

⋮

$$* \rho_p = \frac{360^\circ}{2} = 180^\circ.$$

$$\rho = \frac{5}{6} * 180^\circ = 150^\circ.$$

⇒

3

$$k_{p_1} = \sin \frac{150^\circ}{2} = 0.966 \Rightarrow \text{reduced by } 3.4\%$$

$$k_{p_3} = \sin \frac{3(150^\circ)}{2} = -0.707 \Rightarrow \text{do not exist.}$$

$$k_{p_5} = \sin \frac{5(150^\circ)}{2} = 0.259 \Rightarrow \text{reduced by } 75\%.$$

$$k_{p_7} = \sin \frac{7(150^\circ)}{2} = 0.259 \Rightarrow \text{reduced by } 74\%.$$

$$k_{p_9} \Rightarrow \text{do not exist.}$$

The advantage of fractional pitch winding is to reduce the magnitude of the most dangerous harmonic component (5^{th} and 7^{th}) \Rightarrow to get more sinusoidal wave form.

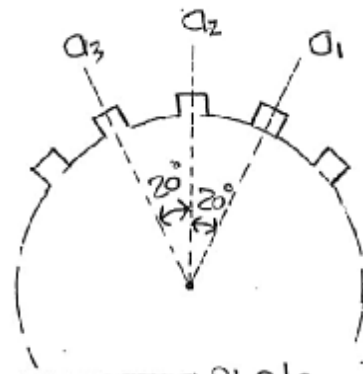
The disadvantage is slightly reduce the magnitude of the fundamental component.

! Distribution windings and Distribution factor :-

$$\text{Slot pitch} = \frac{360^\circ}{18} = 20^\circ$$

$$\frac{360^\circ}{20^\circ} = \underline{\underline{18 \text{ Slots}}}$$

$$\div 3 = \underline{\underline{6 \text{ Slots/phase}}}$$



→ Each coil needs 2 slots Single layer

$$\# \text{ of coils} = \frac{6}{2} = \underline{\underline{3 \text{ Coils}}}$$

→ each coil have 10 turns ! :-

$$\begin{aligned} \# \text{ of turns per phase} &= 10 * 3 \\ &= \underline{\underline{30 \text{ turns/phase}}} \end{aligned}$$

$$\theta = \omega t$$

$$20^\circ = 2\pi f t$$

$$t = \frac{20^\circ}{2\pi f} * \frac{\pi}{180^\circ}$$

double layer → # of slots = # of coils.

$$\tilde{E}_{a_2} = E \angle 0^\circ \rightarrow \text{reference.}$$

$$\tilde{E}_{a_1} = E \angle -20^\circ$$

$$\tilde{E}_{a_3} = E \angle 20^\circ$$

$$\begin{aligned} \rightarrow \tilde{E}_a &= \tilde{E}_{a_1} + \tilde{E}_{a_2} + \tilde{E}_{a_3} \\ &= E \angle -20^\circ + E \angle 0^\circ + E \angle 20^\circ = \dots \end{aligned}$$

$$|\tilde{E}_a| = 2.879 E \rightarrow \text{reduction}$$

$$\frac{2.879}{3} = 0.95 = k_d \quad \text{: distribution Factor.}$$

* توزيع الـ winding في Slots يقل الـ induced Voltage

* # of Slots أكبر ← angle (بين كل Slot و Slot) أصغر ← k_d أقل

$$\rightarrow k_d = \frac{\sin(n\delta/2)}{n \sin(\delta/2)} \quad , \text{ where :}$$

n : no. of slots per pole per phase.

δ : angle between each two successive slots. 6

→ If $n=3$, $\gamma=20^\circ$!

$$k_d = \frac{\sin(3 * 20^\circ/2)}{3 \sin(10^\circ)} = 0.96$$

* $k_w = k_p k_d \Rightarrow$ winding factor.

* $E_{rms} = 4.44 \Phi f N k_w$

→ $k_p \approx 0.96 \approx k_d$: typical.
(0.94 → 0.98)

1 AC Machines power flows and losses:

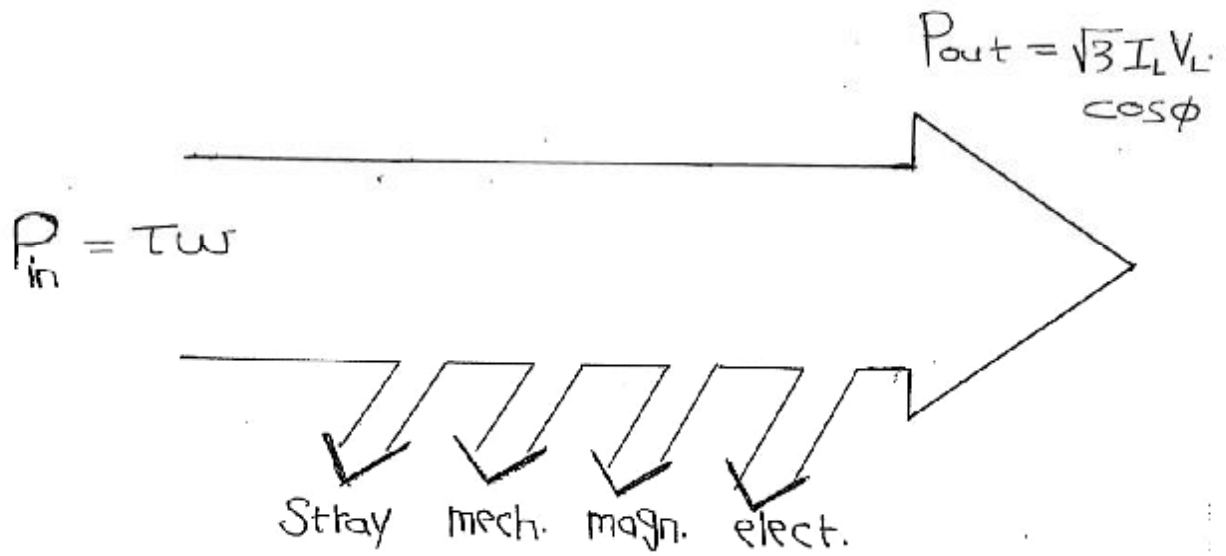
* Machines Convert P_m generator $\rightarrow P_e$ through the action of magnetic field.
 P_e motor $\rightarrow P_m$ magnetic field.

\rightarrow magnetic losses \rightarrow hysteresis losses \rightarrow at Core, Stator and Rotor.
and eddy current losses.
(Mostly at stator)

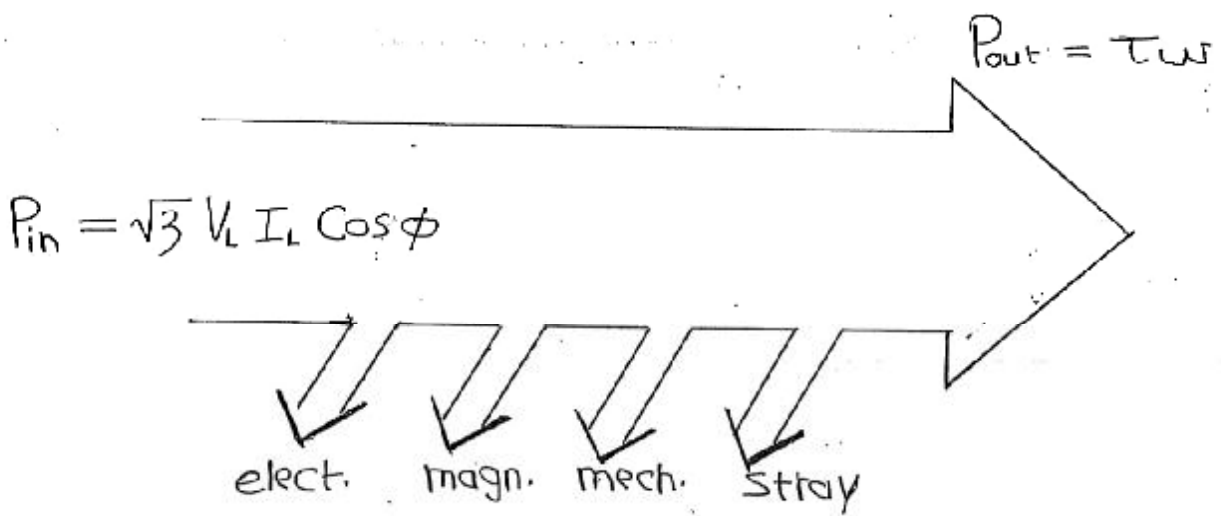
\rightarrow electrical losses \rightarrow at rotor and stator.
($I^2 R$).

\rightarrow mechanical losses \rightarrow friction $\left\{ \begin{array}{l} \text{with air} \rightarrow \text{windage losses.} \\ \text{with Bearings} \rightarrow \text{friction losses.} \end{array} \right.$

\rightarrow Stray losses \rightarrow difference in calculations, sometimes cases by aging.



* Power flow of generator.



* Power flow of motor.

#. typical Practical \leftarrow Size of the machine \leftarrow $\frac{1}{L} \frac{dL}{L}$

Example : $p=2$, 3- ϕ , Y-connected
 Synch. generator, Double layer,
 4 Coils/Phase, 10 turns/Coil, Coil pitch = 150°
 $n_s = 3000$ rpm, $\Phi = 0.019$ Wb/pole.

a) Slot pitch ?

Double layer : #. of Slots = #. of Coils \therefore
 $= \underline{\underline{4 \text{ Slots/phase}}}$

in 3- $\phi \rightarrow 4 \times 3 = \underline{\underline{12 \text{ Slots}}}$

$$\frac{*360}{12} = 30^\circ \text{ elec.}$$



\therefore

$\rightarrow 30^\circ$ mech.

b) How many Slots do the Coil Span
 Coil pitch in terms of Slots ?

$$\frac{150^\circ}{180^\circ} = \frac{5}{6} \rightarrow \therefore 5 \text{ Slots.}$$

Coil Pitch \swarrow
 elect. angle \searrow
 \searrow # of Slots per pole.



C) E_A ?

$$E_A = 4.44 \Phi f N k_w$$
$$= 4.44 \Phi f N k_p k_d$$

flux per pole.

$$f = \frac{n_s P}{120} = \frac{(3000)(2)}{120} = 50 \text{ Hz.}$$

of turns per phase = N

$$\begin{aligned} &= 4 \text{ coils} \times 10 \text{ turns} \\ &\text{if single layer} \leftarrow \text{phase coil} \\ &\rightarrow 2 \text{ coils} \\ &\text{phase} = 40 \text{ turns/phase.} \end{aligned}$$

$$k_p = \sin \frac{p}{2} = \sin \frac{150^\circ}{2} = 0.966.$$

$$k_d = \frac{\sin(n\alpha/2)}{n \sin(\alpha/2)} = \frac{\sin(2(\frac{30^\circ}{2}))}{2 \sin(\frac{30^\circ}{2})} = 0.966.$$

$$n = \# \text{ of slots per pole per phase} = 2$$

6 / 3 = 2

$$E_A = (4.44)(0.019)(50)(40)(0.966)(0.966)$$
$$= 157 \text{ V} \rightarrow \text{phase voltage.}$$

(Y-connected).

d) Line Voltage ?

$$\begin{aligned}\rightarrow V_T &= \sqrt{3} V_\phi = \sqrt{3} (157) \\ &= 272 V.\end{aligned}$$

e) How much Suppression does the fractional pitch winding give for the 5th harmonic component ?

$$\rightarrow k_{p_5} = \sin \frac{(5)(150)}{2} = 0.259.$$

Given 4 Poles, 80 Slots, 5th harmonic pitch factor = Zero

→ coil pitch in terms of slots?

$$\sin \frac{5\beta}{2} = 0$$

$$\frac{5\beta}{2} = 180^\circ \rightarrow \beta = 1 : \beta = 72$$

$$\rightarrow \frac{X}{20} = \frac{72}{180}$$

$$\frac{80}{4} \triangleleft \frac{80}{4} \rightarrow X = 8 \checkmark$$

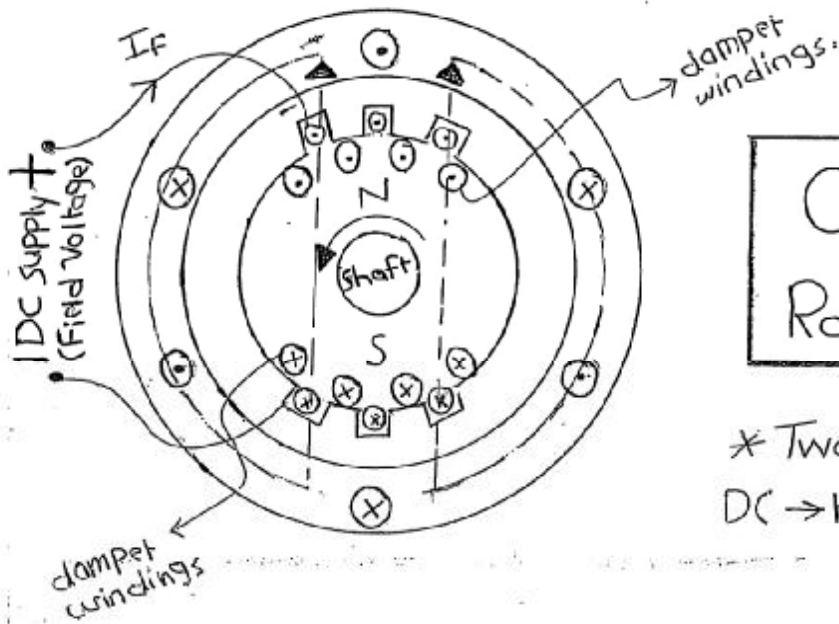
$$\rightarrow \beta = 2 : \beta = 144$$

$$\rightarrow \frac{X}{20} = \frac{144}{180} \rightarrow X = 16 \checkmark$$

...

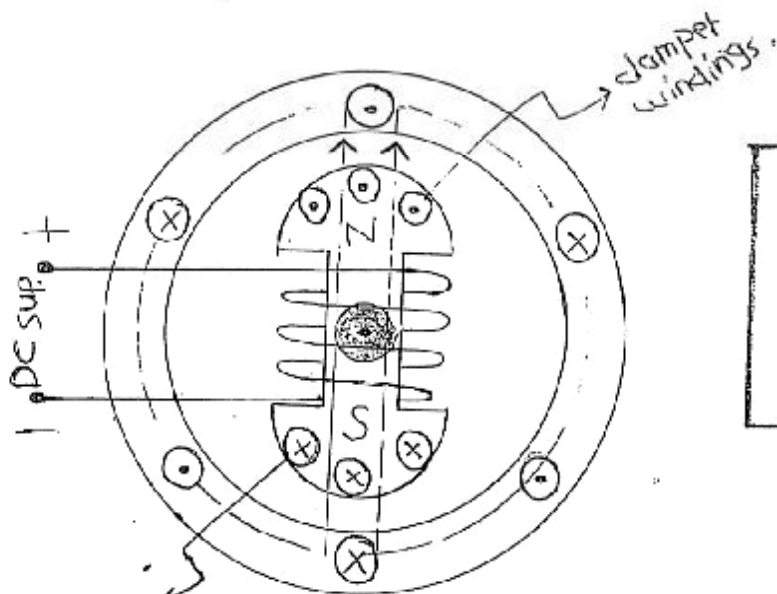
3

Chapter 4 :- Synchronous Generators :



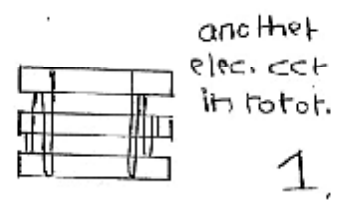
Cylindrical Rotor Type.

* Two electrical ccts :
DC → rotor, AC → Stator.



Salient pole Type.

→ copper bars short circuited from the two ends at the pole faces.



Damper windings: are windings placed in slots curved at the pole face of the rotor of synchronous generator, they are used to damp oscillation in case of transients,

In steady-state the induced voltage and current inside them are zero.

* Why we use damper windings?

① In transient case the current induced in them produces anti-torque which damp oscillation of the rotor.

② To start-up synchronous motor.

* There are two common approaches to supply dc power to field windings on the rotor :-

- 1) Supply dc power from an external dc source to the rotor by means of slip rings and brushes.
- 2) Supply dc power from dc power source mounted directly on the shaft of the Synch. generator.

* Slip rings and brushes problems :-

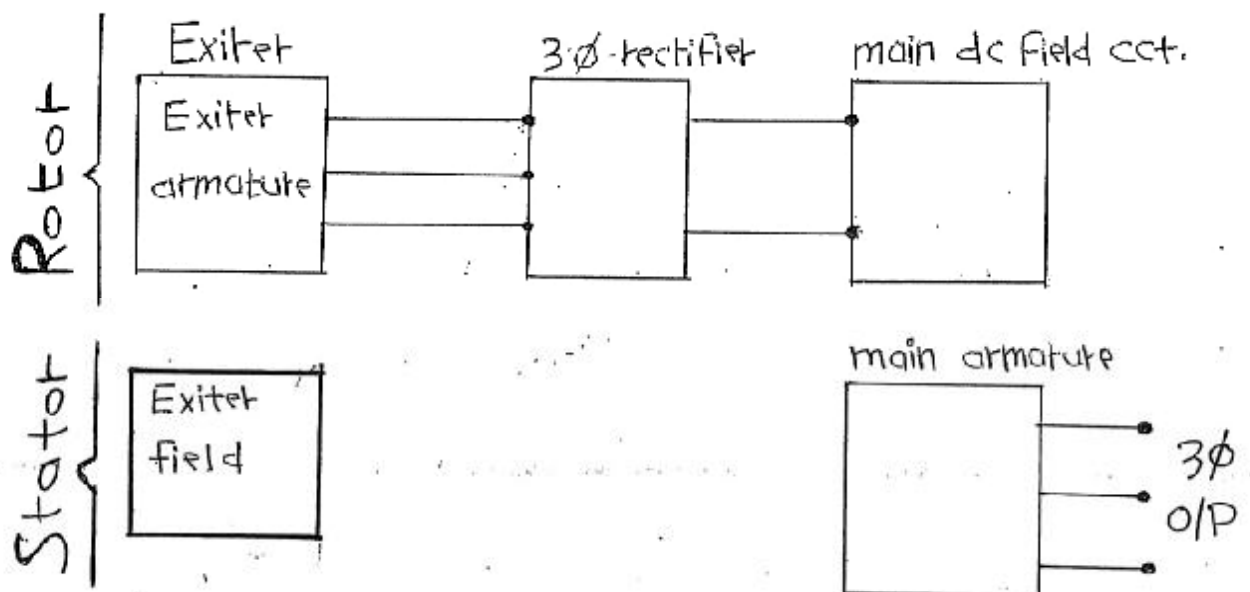
- 1) They increase the amount of maintenance required on the machine, since the brushes checked for wear regularly.
- 2) Brush voltage drop can be the cause of significant power losses on machines with large amount of field currents.

⇒ So, they are used in smaller Synch. machines, because no other method of supplying dc-field current is cost-effective

Brushless exciter is a small AC generator,

- ↳ its field cct. mounted on the Stator.
- ↳ its armature cct. mounted on the rotor.

□ main exciter :-



* By controlling the small dc field current of the exciter generator (located on stator), we can adjust the field current on the main machine without slip rings and brushes.

* brushless exciter requires much less maintenance than slip rings and brushes, since no mechanical contacts between rotor and stator.

2] Pilot exciter :- Small ac generator.

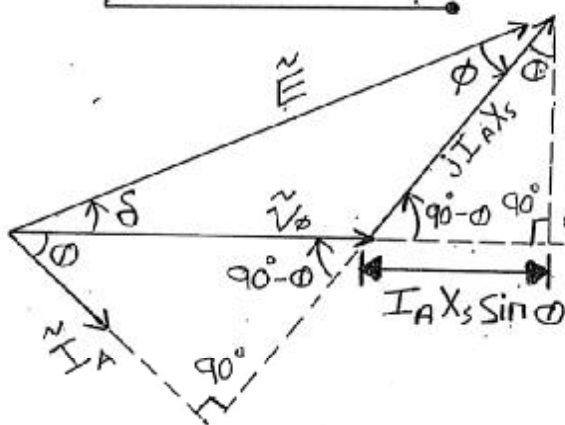
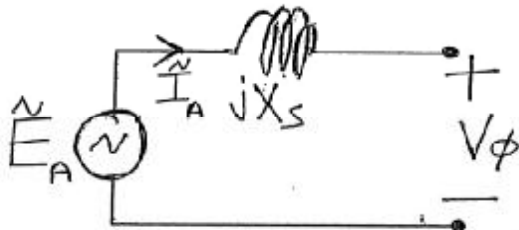
- ↳ Permanent magnets mounted on the rotor
- ↳ 3 ϕ winding on the stator.

Why we use pilot exciter ?

- 1) To make the excitation of a generator completely independent of any external power source.
- 2) It produces power for the field ckt. of the exciter, which in turn controls the field circuit of the main machine.

* IF the pilot exciter is included on the generator shaft, then no external electric power is required to run the generator.

* Power and Torque of Synchron. gen. :-



$$E_A \sin \delta = I_A X_s \cos \phi$$

$$= I_A X_s \sin (90 - \phi)$$

$$I_A \cos \phi = \frac{E_A \sin \delta}{X_s}$$

$$P_{out} = 3 V_\phi I_A \cos \phi$$

$$= 3 V_\phi \frac{E_A \sin \delta}{X_s} \quad \text{--- Power angle equ.}$$

$$P_{out} = \frac{3 V_\phi E_A}{X_s} \sin \delta$$

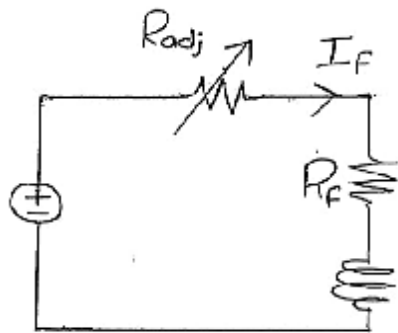
Static stability limit.

δ \rightarrow delta angle / Rotor angle
($15^\circ \rightarrow 65^\circ$).

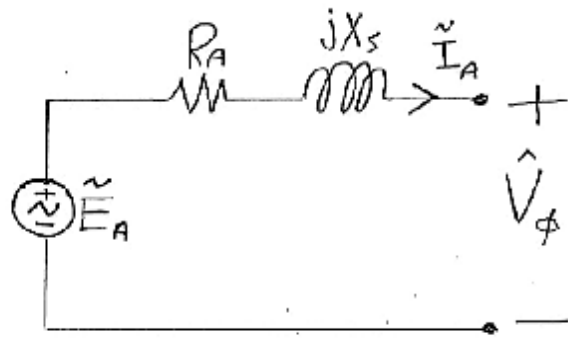
$$T_{ind} = \frac{3 V_\phi E_A \sin \delta}{\omega_s X_s}$$

When $R_A = 0$.

Synchronous Generator Equivalent Circuit :



Field cct.

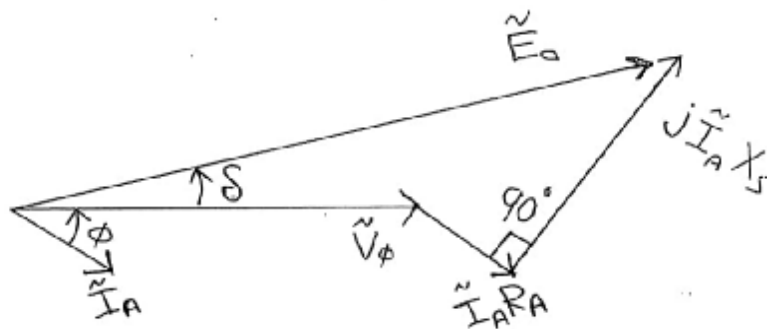


Armature cct.

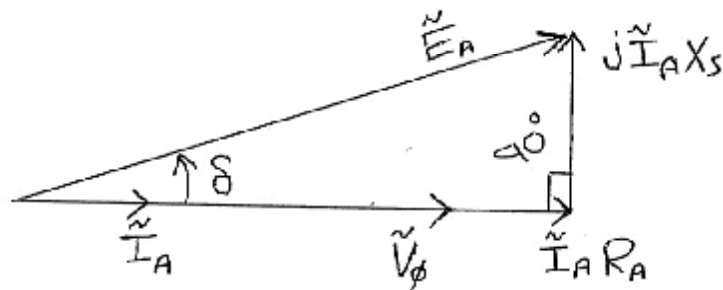
Per-phase Equ. cct.

$$\tilde{E}_A = 4.44 \Phi f N k_w$$

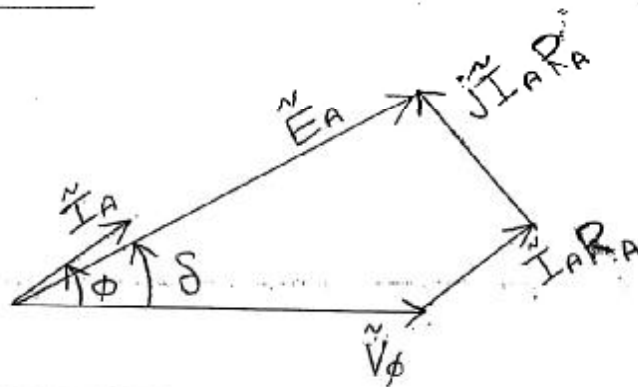
* Lagging PF Load :



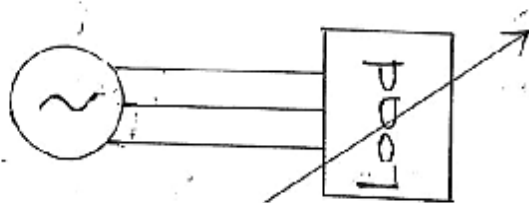
* Unity PF :



* Leading PF :-



Synchronous Generator Operating Alone :



⇒ operating alone : not connected to the net work, (Separate Power System).

Assumptions :

① Fixed \tilde{E}_A (Fixed Φ).

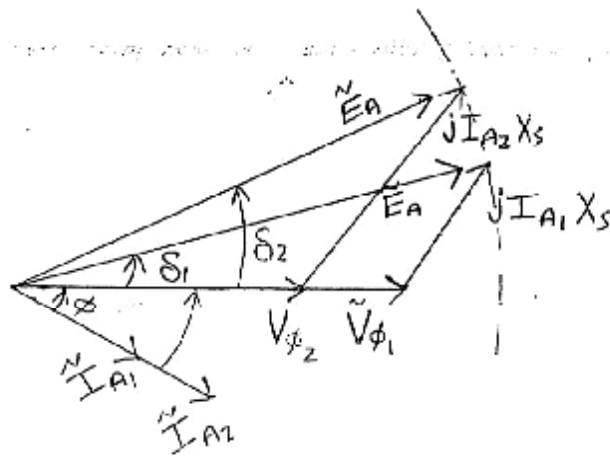
② No Change in the P.F of the load. $\rightarrow I_F$ must be const.

③ Constant ω . ④ R_A is neglected.

* Variable load.

← زيادة load \rightarrow تكبير زيادة current
 (Parallel loads)

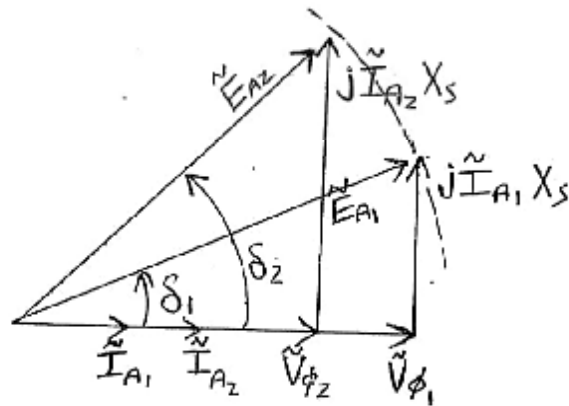
□ with lagging P.F Load :



$$V_R\% = \frac{|\tilde{E}_A| - |\tilde{V}_\phi|}{|\tilde{V}_\phi|} 100\% > 0$$

$\delta \uparrow, V_\phi \downarrow, I_A \uparrow \Rightarrow$ if the load increase.

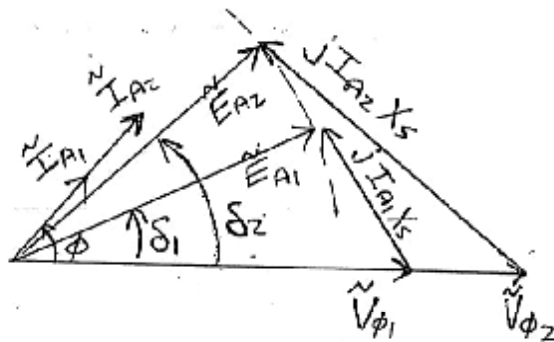
2] Unity P.F load :



$$V_R\% > 0.$$

$$\delta \uparrow, V_\phi \downarrow, I_A \uparrow.$$

3] with leading P.F load :



$$V_R\% < 0.$$

$$\delta \uparrow, I_A \uparrow, V_\phi \uparrow.$$

* If the load is capacitive $\rightarrow V_R\% < 0$.

* If the load is inductive $\rightarrow V_R\% > 0$.

How to keep V_ϕ constant?

↑
terminal
Voltage

This is done by Automatic Voltage

Regulator (AVR) : (negative feedback control system) which changes the field current (Φ) in order to keep

V_ϕ constant, by changing the R_{adj} .
increasing and decreasing.

if $V_\phi \uparrow \longrightarrow I_f \downarrow$

if $V_\phi \downarrow \longrightarrow I_f \uparrow$

} due to R_{adj}

Example: 480V, 60Hz, Y-connected
 Stator, $P = 6$, $X_s = 1 \Omega$, $I_A = 60A$,
 0.8 p.f lagging load, Full load

$P_{F\&W} = 1.5 \text{ kW}$, $P_{\text{core}} = 1 \text{ kW}$, $R_A = 0$,
 mech. losses.

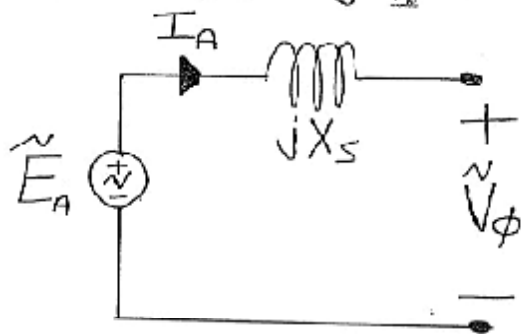
I_f is adjusted such that the terminal
 Voltage is 480V at no load.
operating point.

a) $n_s = ?$

$$f = \frac{n_s P}{120} \rightarrow 60 = \frac{n_s (6)}{120}$$

$$n_s = 1200 \text{ rpm}$$

b) ① $\tilde{V}_\phi = ?$ if it is loaded with Full-load
 and 0.8 lagging p.f ?



$$\rightarrow \tilde{E}_A = \tilde{V}_\phi + j \tilde{I}_A X_s$$

at phase

$$\frac{480}{\sqrt{3}} \angle \delta = V_\phi \angle 0 + j(60 \angle -\cos^{-1} 0.8)(1)$$

$$\frac{480}{\sqrt{3}} \cos \delta + j \frac{480}{\sqrt{3}} \sin \delta = V_\phi + j[60 \times 0.8 - j60 \times 0.6]$$

$$= V_\phi + j60 \times 0.8 + 60 \times 0.6$$

$$\rightarrow \frac{480}{\sqrt{3}} \cos \delta = V_\phi + 60 \times 0.6 \text{ ---- (1)}$$

$$\rightarrow \frac{480}{\sqrt{3}} \sin \delta = 60 \times 0.8 \text{ ---- (2)}$$

from (2) :- $\sin \delta = 0.17 \rightarrow \delta = 9.97^\circ$.

from (1) :- $V_\phi = 236.9 \text{ V}$.

علينا معرفة المبروفة لأنه القطر E_A
 والمطلوب حساب V_ϕ .

② $\tilde{V}_\phi = ?$ if it is loaded with full-load at unity P.F. Load ?

$$\frac{480}{\sqrt{3}} \angle \delta = V_\phi \angle 0 + j(60 \angle 0)(1)$$

$$\frac{480}{\sqrt{3}} \cos \delta + j \frac{480}{\sqrt{3}} \sin \delta = V_{\phi} + j 60$$

$$\rightarrow \frac{480}{\sqrt{3}} \sin \delta = 60 \rightarrow \delta = 12.5^{\circ}$$

$$\rightarrow \frac{480}{\sqrt{3}} \cos \delta = V_{\phi} \rightarrow V_{\phi} = 270.5 \text{ V}$$

③ $\tilde{V}_{\phi} = ?$, if it is loaded with leading P.F load ?

$$\frac{480}{\sqrt{3}} \cos \delta + j \frac{480}{\sqrt{3}} \sin \delta = V_{\phi} + j(48 + j36)$$

$$\rightarrow \frac{480}{\sqrt{3}} \sin \delta = 48 \rightarrow \delta = 9.97^{\circ}$$

$$\rightarrow \frac{480}{\sqrt{3}} \cos \delta = V_{\phi} - 36 \rightarrow V_{\phi} = 308.9 \text{ V}$$

c) $\% \text{Z}$ for the case b: ① ?

$$\% \text{Z} = \frac{P_{\text{out}}}{P_{\text{in}}} * 100\%$$

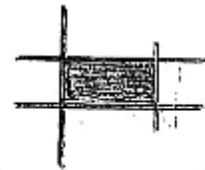
$$P_{\text{out}} = \sqrt{3} \underset{\substack{\uparrow \\ \sqrt{3} V_{\phi} = 410 \text{ V}}}{V_L} I_L \cos \phi = \sqrt{3} (410) (60) (0.8) = 34113.6 \text{ W}$$

e) Voltage Regulation ?

$$V_{R_1} \% = \frac{\frac{480}{\sqrt{3}} - 236.8}{236.8} 100\% = 17\% .$$

$$V_{R_2} \% = \frac{\frac{480}{\sqrt{3}} - 270.4}{270.4} 100\% = 2.48\% .$$

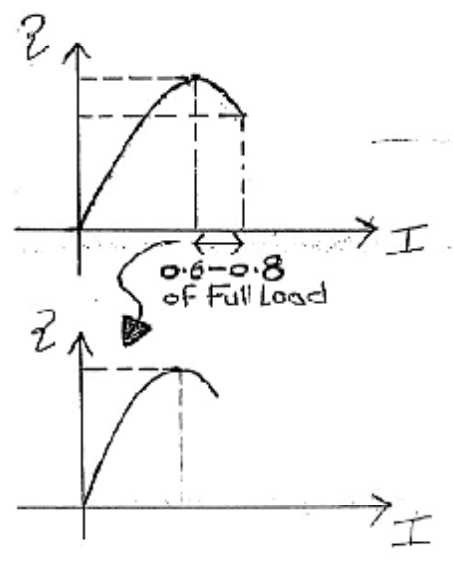
$$V_{R_3} \% = \frac{\frac{480}{\sqrt{3}} - 308.8}{308.8} 100\% = -10.25\% .$$



Parallel Operation of AC Generator :-

why Connect generators in parallel ?

- To Supply more power if the demand increases.
 ↑
 active & reactive
- Increased Reliability of the Power System.
- To have a more efficient Power System:



> operating Pnt. الفكرة تكون في
 max. point { عند Generator }
 η
 # Δ, losses ↓

if only one generator is used and it is not operating at near full load, then it will be inefficient. With several smaller machines in parallel, it is possible to operate only a fraction of them. The ones that do operate are operating near full load and thus more efficiently.

Conditions of Paralleling :-

- 1] The rated line voltages, must be equal otherwise, Circulating current flow between generators.
- 2] They must have the same phase sequence, otherwise, \rightarrow Voltage difference \rightarrow Circulating current
generator \downarrow الثاني motor .
- 3] The same phase angles, otherwise, \rightarrow Circulating current.
- 4] The frequency of the oncoming generator must be slightly higher. [1Hz, 1.5Hz] than the freq. of the running system.

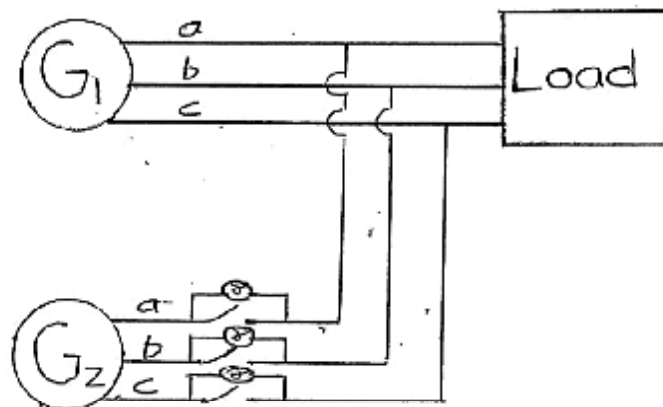
\rightarrow When the generator is loaded after switching the CCT. breakers on, the rotational speed decreases (freq. decreases).

generator بحركه 1, prime mover 6 4 نوصيه ع لشبكة بصير loaded و عليه 50Hz

تتنزل سرعه 1 prime mover \leftarrow وكاستجابة لازم يرجع لسرعة \leftarrow بس ما راح يقدر اذا كان زي

delay كبير \leftarrow فبتعطين 1, gener. الثاني
* or 51Hz
52Hz

Paralleling procedure :-



When the three light-bulb become dark
[difference voltages = 0]

↳ G_1, G_2 have the Same Voltages,
Same Freq, Same phase Shift,
Same Seq.



↳ Then Switch the
cct breaker (G_2) ON.

at Small systems
or PV's systems.

#

→ Big Stations → Big cct. breakers



We use Synchroscope
as a switch for G_2
When 4 conditions
are met.

* automated
* need high power.

Frequency-Power and Voltage-Reactive Power Characteristics of a Synchronous Generator:

Prime movers \rightarrow any Type of mech. engines,
 \rightarrow wind turbines,
 \rightarrow water turbines,
 \rightarrow gas turbines ...

Speed \downarrow
When Load \uparrow
(freq. \downarrow) .

\rightarrow Variation in Speed with increasing load is non-linear.

ترجع تزوّد كمية Fuel \rightarrow الحل

demand \uparrow \rightarrow Reactive power from gen. \uparrow

Load \uparrow \rightarrow Reactive comp. current \uparrow

Voltage \downarrow .

ترجع تكبيره \rightarrow induced Voltage

$$SD\% = \frac{n_{NL} - n_{FI}}{n_{FI}} * 100\%$$

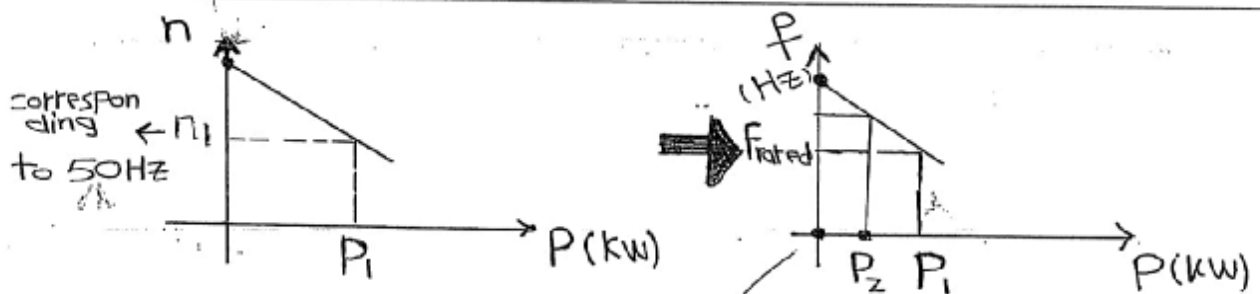
↑
Speed regulator.
Speed drop.



كلما اقتربن لا ديفر يكون أحسن ✓

✓ فقط Sync. motor لا SD = speed

$$SD\% = 2\% - 4\%$$



فجأة $P = 0$ (Short cct. or open cct.)

تقل السرعة
تقل fuel للengine

تتغير السرعة

Turbine governor

due to -ve feedback control speed
↳ Time delay → (1-2 second).

$$P = S_p (f_{nl} - f_{sys}) \text{ --- Freq. Power}$$

Characteristic.

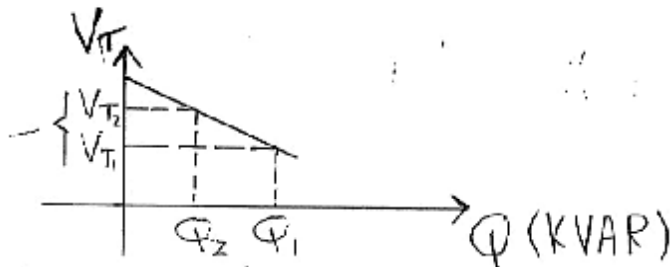
P : output power from generator.

S_p : Slope of the curve (KW/Hz) or (MW/Hz)

f_{nl} : no-load freq.

f_{sys} : System operating freq.
(constant: 50 or 60 Hz).

A similar relation for the voltage as function of the output reactive power can be derived:



$$V_T \downarrow, Q \uparrow$$

parallel generator have the same V_T .

demand $\downarrow \Rightarrow Q \downarrow \Rightarrow$

\Rightarrow AVR \rightarrow (-ve) feedback controller.

\hookrightarrow Time delay (in msec.) \rightarrow measuring process
order: Field current due to ϕ

* ولوقت لا يوقف لا Line Field Curr.

ϕ due to Z

Note :-

In Synchronizing generators operating alone, P & Q supplied are the amount demanded by the load.

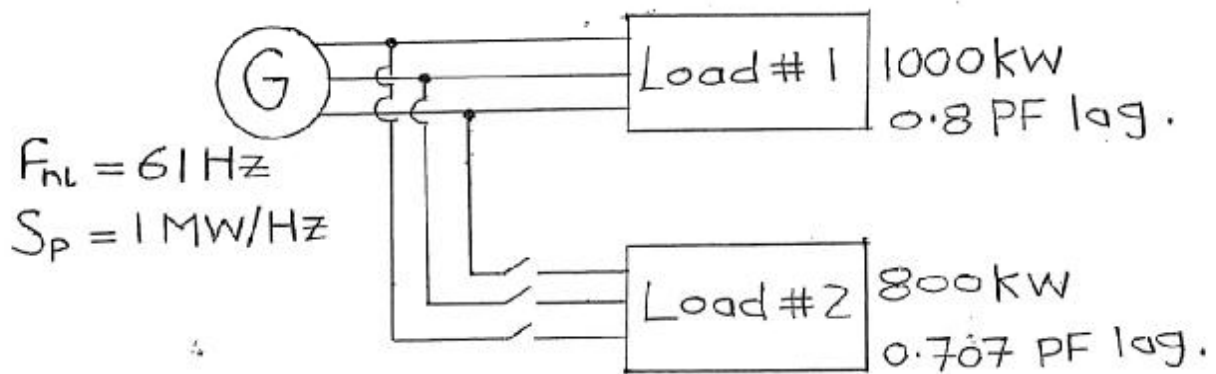
~~~~~  
all the power generated must be consumed, otherwise, → difference voltage and losses  
سبب اختلاف الجهد

---

$P \uparrow \rightarrow S \downarrow \xrightarrow{\text{الحل}} \text{Governor}$  } -V<sub>r</sub> control system.  
 $Q \uparrow \rightarrow V \downarrow \xrightarrow{\text{الحل}} \text{AVR}$  }



Example :-



a) Before switching ON,  $f_{sys}$  ?

$$P = S_P (f_{nl} - f_{sys})$$

$$1 \times 10^6 = 1 \times 10^6 (61 - f_{sys}) \rightarrow f_{sys} = 60 \text{ Hz}$$

b) with load #2 connected,  $f_{sys}$  ?  
demand  $\uparrow$

$$P = S_P (f_{nl} - f_{sys})$$

$$1 \times 10^6 + 0.8 \times 10^6 = 1 \times 10^6 (61 - f_{sys})$$

$$f_{sys} = 59.2 \text{ Hz}$$

c) What action should be taken to restore the nominal freq.?

The fuel input to the prime mover must increase, (no-load freq. must increase)

$$\rightarrow 1 \times 10^6 + 0.8 \times 10^6 = 1 \times 10^6 (f_{nl} - 60)$$

$$f_{nl} = 61.8 \text{ Hz}$$

---

\* Summary for Generator operating alone:

1] Real power and Reactive power must be equal demand of the load.

(the same power generated by generator)

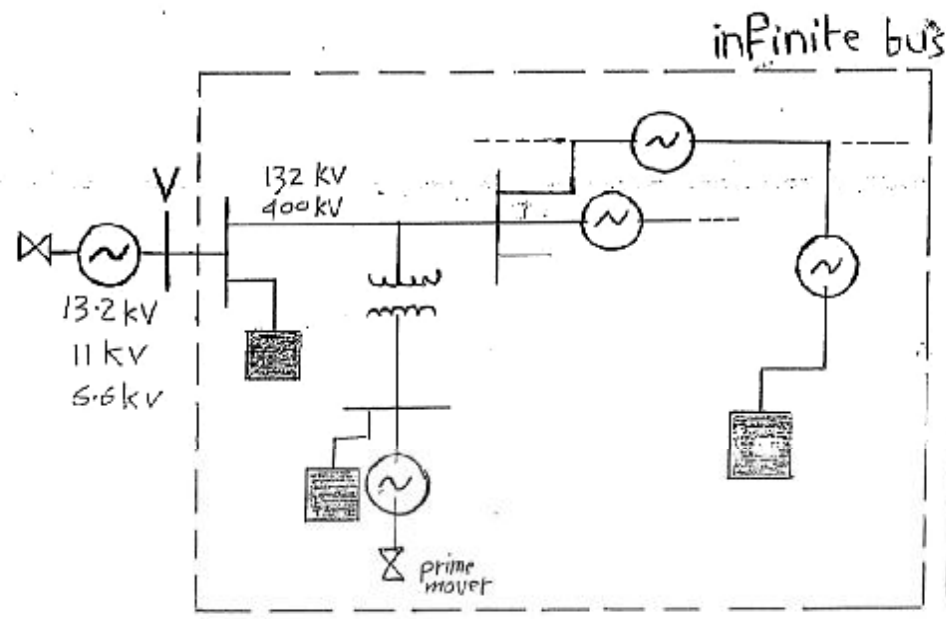
2] The Governor set point controls the operating freq.

3] The Field current controls the terminal voltage.

---

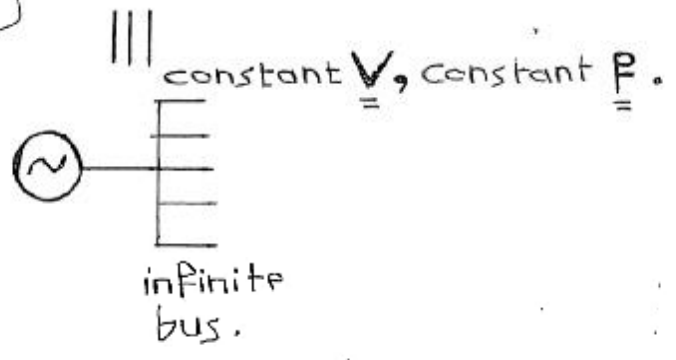
# Operation of Generators in Parallel with large power System :

Infinite bus : So large Power System where its Voltage and Frequency do not vary regardless <sup>بغض النظر</sup> the amount of active and reactive Power consumed.

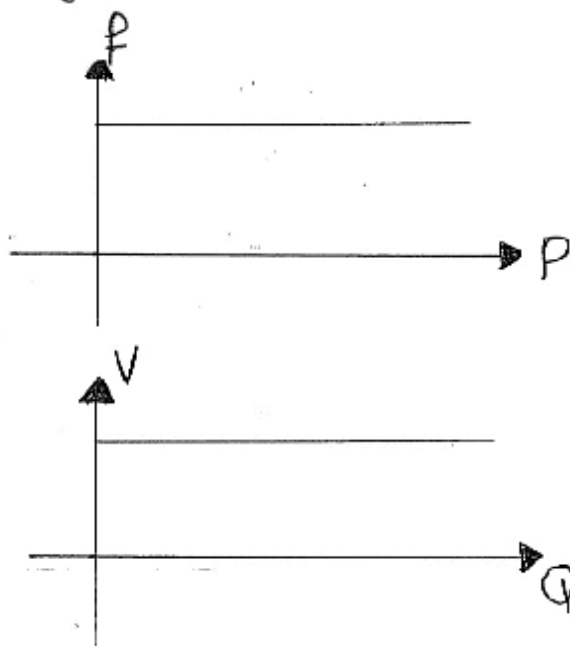


- large # of Generators.
- large # of Transm. lines
- large # of Transformer.
- large # of Loads.

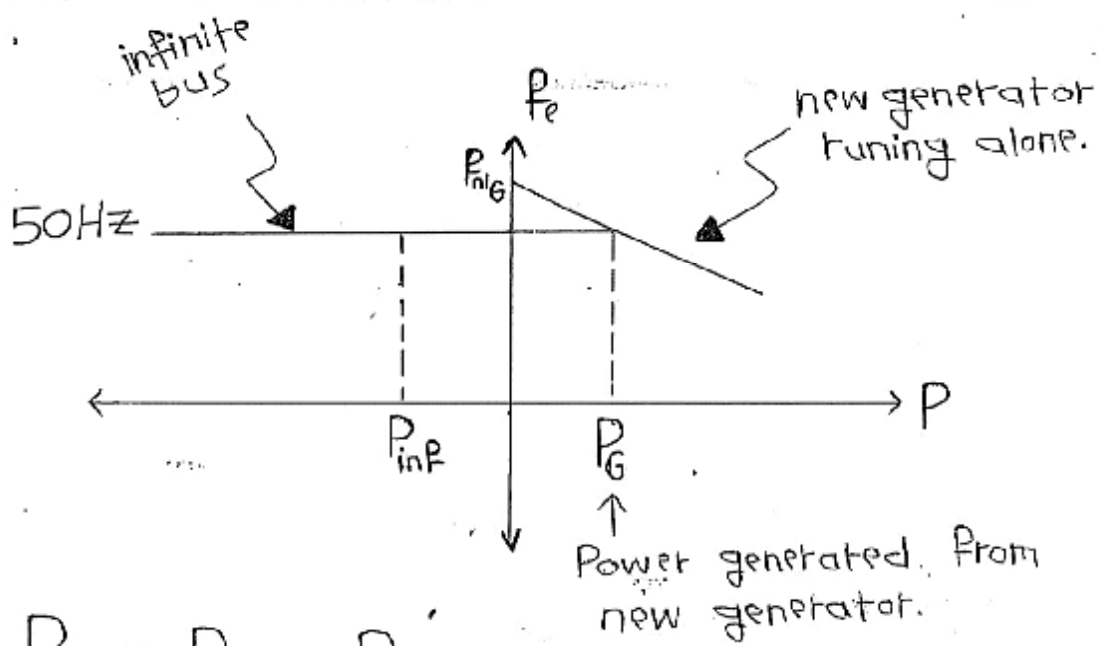
جبرال جن.  
الجبرال انه  
لشغل بنفس  
ال V و P.



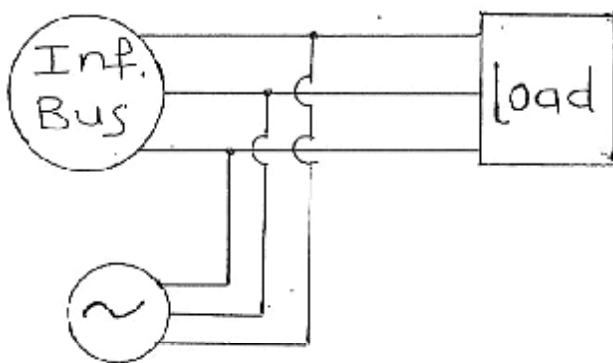
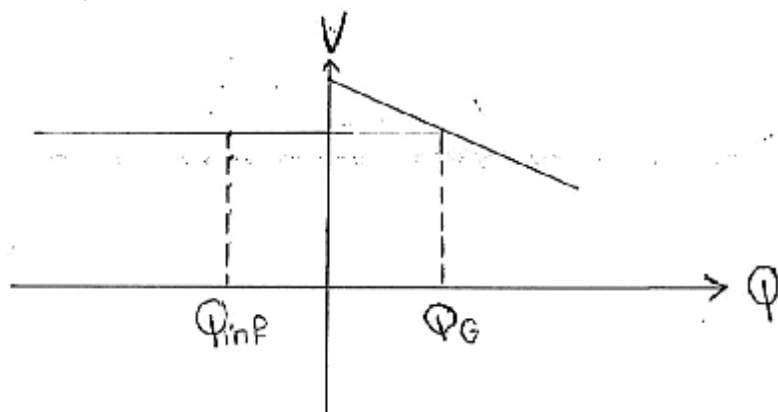
When we add a new generator by using Synchroscope, initially the original generator of the power system is still bigger than the new one. Then the new one takes the same voltage and the same freq. of the power system becoming similar to the original generator.



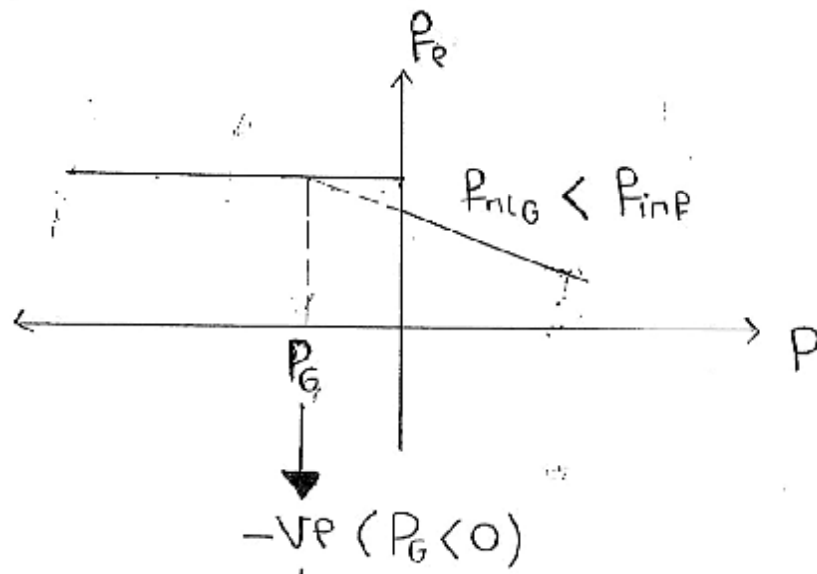
How the new generator takes the same network's voltage even though there are a lot of transformers inside the infinite bus? \* Because we use the per unit values.



$$P_{tot} = P_{inf} + P_G$$



if  $P_{mLG}$  was not slightly higher than  $P_{imp}$ :



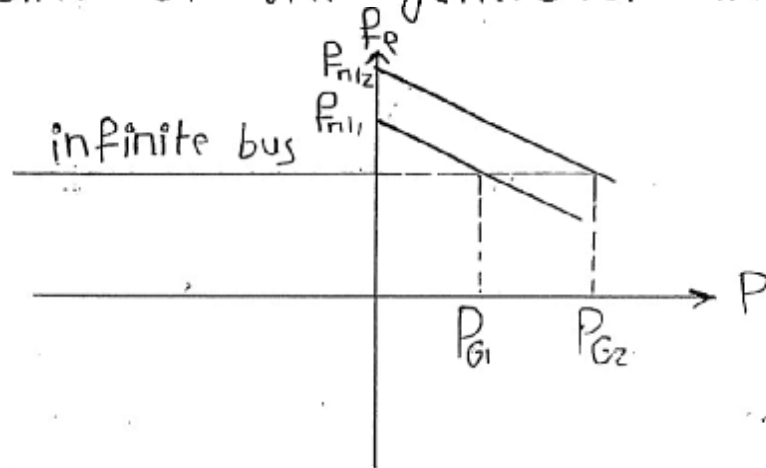
The generator will  
run as motor.  
(consumed power)

- to turn on protection system;
- due to → directional of the power.
- or due to current.

Power System ↑

- Voltage constant المستقر
- cost التكلفة

What happens when the governor set point of the generator increase?



The effect of increasing the governor set point.

→ value of Pule أكبر

- no load freq. of the generator increases,
- Power Contribution of the generator increases,
- Running Freq. of the power system remains constant.

\* at no load → the speed of the generator greater than the synchronous speed.

∴ If you have a generator connected to infinite bus, and you need to increase the active power,

→ Provide the amount of fuel in governor.

---

# In phase diagram :-

assumptions :-

$I_f$  is constant  $\Rightarrow E_A$  is constant.

$\omega$  is constant  $\Rightarrow$  tuning freq. is const.

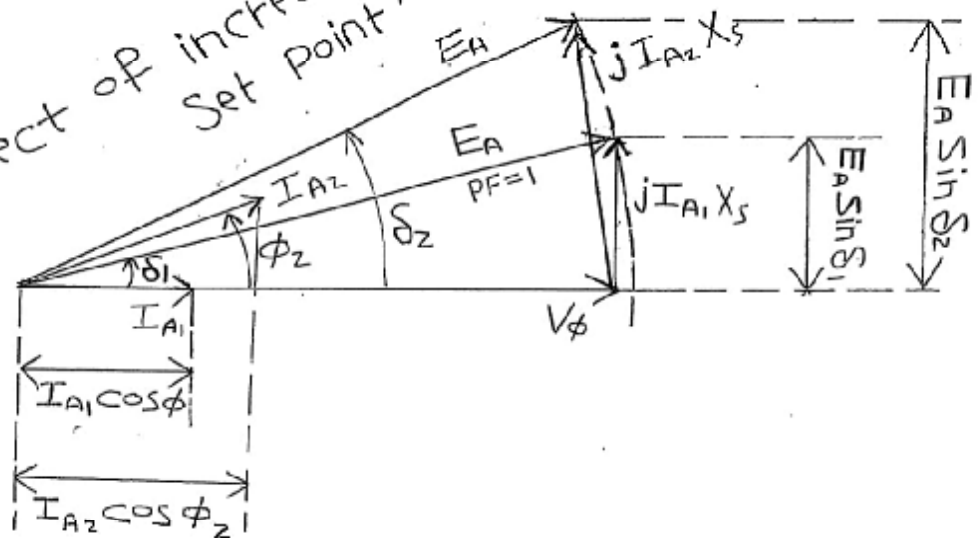
$V_t$  is constant  $\Rightarrow$  terminal voltage is const.

} connection with infinite bus.

$R_A$  is neglected.



"The effect of increasing the governor set point"



$$P \uparrow = \frac{3 V_{\phi} I_A \cos \phi}{\text{const.}} \uparrow$$

$$P \uparrow = \frac{3 V_{\phi} E_A \sin \delta}{X_s} \uparrow$$

\* Conclusion:

1]  $I_A$  increases. (phase current  $\uparrow$ )

2]  $V_{\phi}$  constant.

3]  $I_A X_s$  increases.  
(Voltage drop)

4  $E_A$  constant by assumption.

5  $E_A \sin \delta$  increases.

6  $\delta$  increases.

7 The p.f of the generator becomes  
more leading. ( $\phi \uparrow$ ).

يتحرك باتجاه leading.

8  $I_A \cos \phi$  increases.

9 Power generated ( $P$ ) increases.

10  $Q = 3 V_\phi \underline{I_A \sin \phi} \downarrow$

The variation of  $Q$  is very small  
compared with  $P$ .

( $Q$  consumes power) because the  
generator becomes more leading.

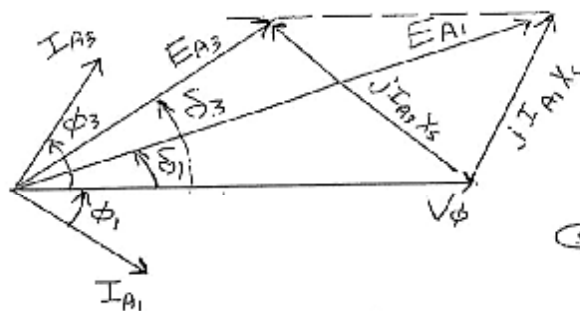
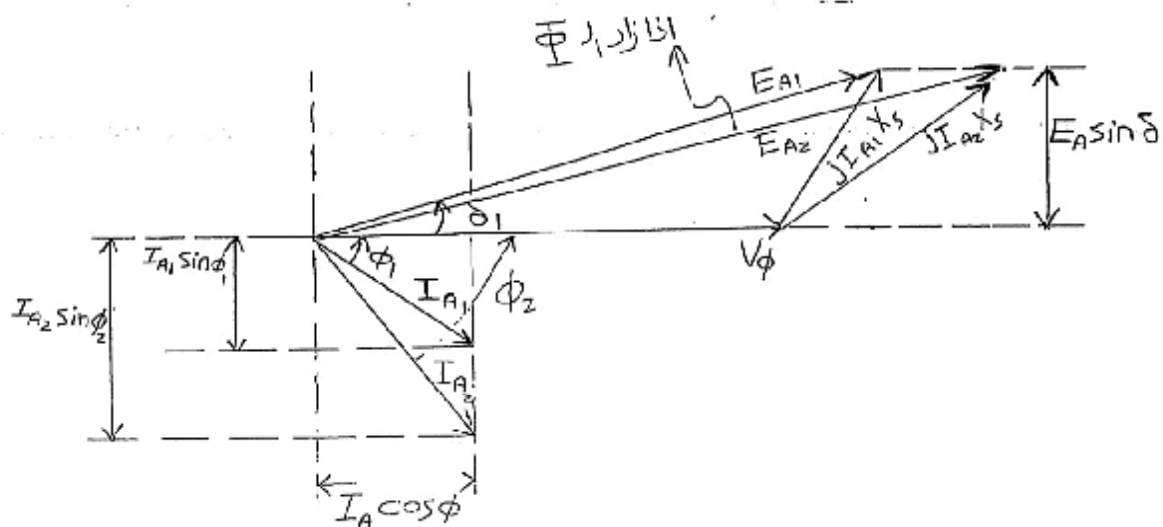
What happens if the field Current increases and the governor Set point remains Unchanged?

\* The input active power remains Constant,  
 $\hookrightarrow I_A \cos \phi$  &  $E_A \sin \delta$  are constant.

\*  $i_f \uparrow \rightarrow \Phi \uparrow \rightarrow E_A = 4.44 \Phi F N K \omega \uparrow$ ,

\*  $V_\phi$  remains constant.

\*  $P$  remains constant  $\rightarrow \omega$  is constant.



$$P = 3V_\phi I_\phi \sin \phi$$

c

$$\boxed{1} \quad I_f \uparrow, \Phi \uparrow, E_A \uparrow,$$

PF becomes more lagging,

$$\delta \downarrow, I_A \sin \phi \uparrow,$$

$$Q_{\text{out put}} \uparrow$$

$P_{\text{out put}}$  remains constant.

$$\boxed{2} \quad I_f \downarrow, \Phi \downarrow, E_A \downarrow,$$

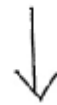
PF becomes more leading,

$$\delta \uparrow, I_A \sin \phi \downarrow,$$

$$Q_{\text{out put}} \downarrow$$

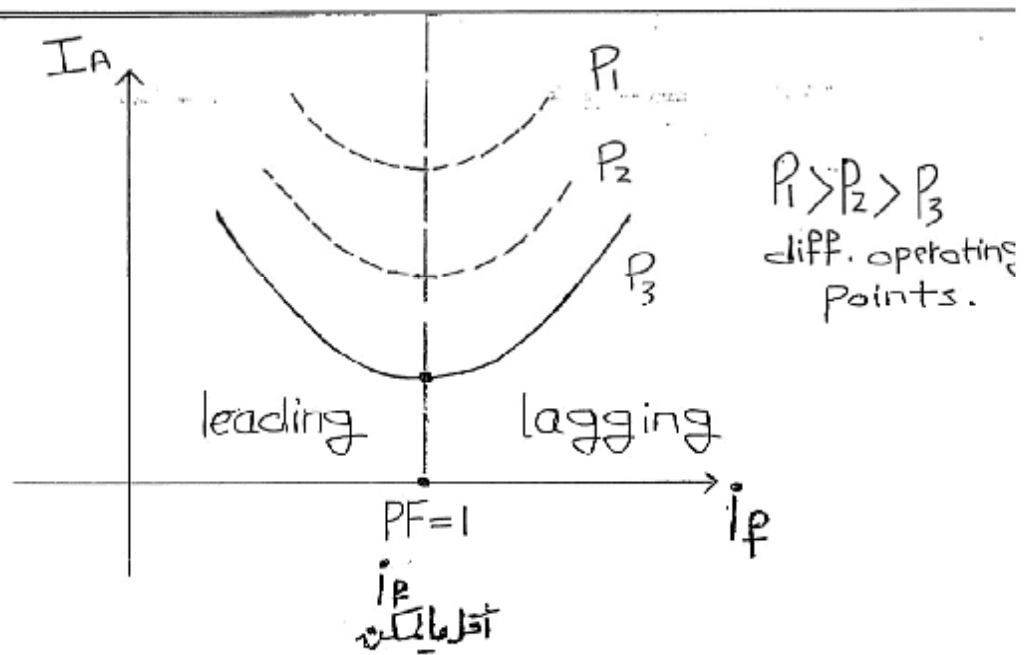
$P_{\text{out put}}$  remains constant.

$$\boxed{3} \quad \text{@ certain value of } I_f \rightarrow PF = 1$$



$I_f$

نقطه یابی است

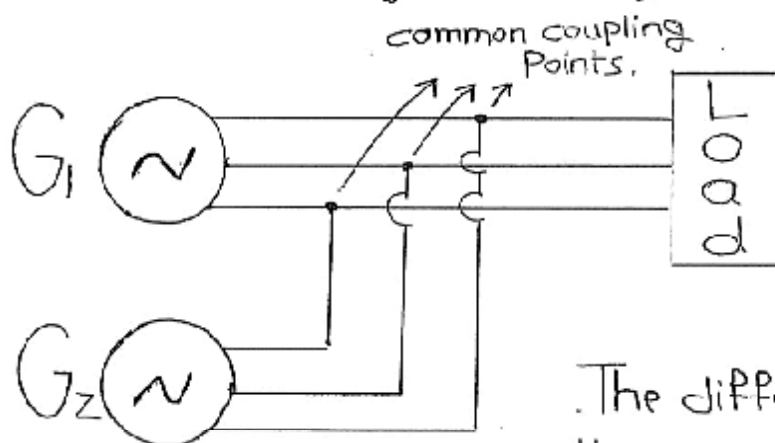


## V-curves of Synch. gen.

### Summary:

- 1) When a generator is connected to an infinite bus, the frequency and terminal voltage are controlled by the power system (infinite bus).
- 2) The governor set point controls the active power generated by the generator.
- 3) The reactive power supplied by the generator is controlled by the field current.

Operation of Generators in parallel with other generators of the same size;



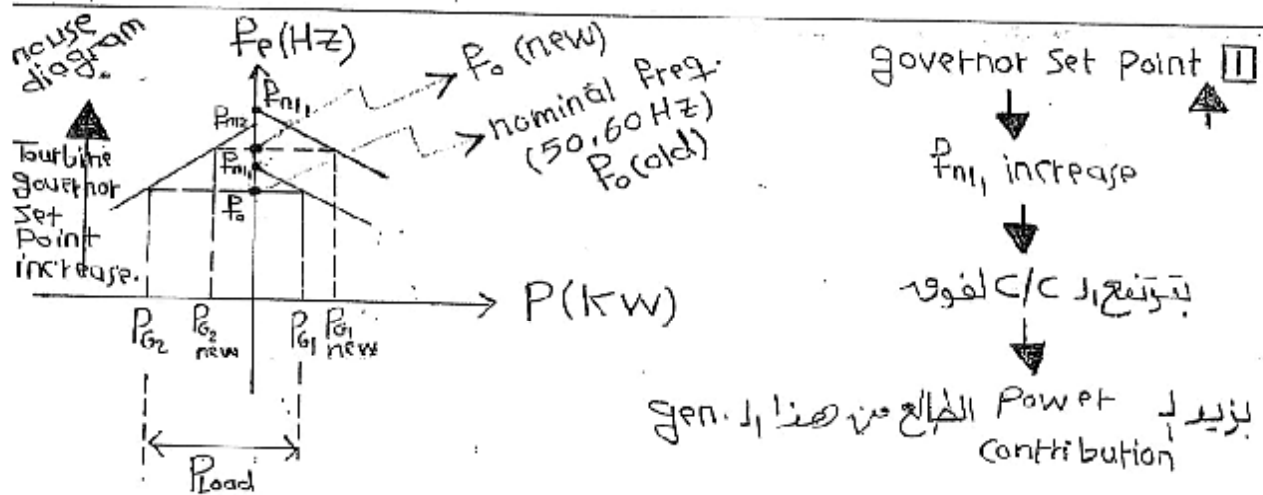
The difference between this case and the previous cases is that here, the Voltage & Freq. are not constant, while in the other cases they were Const

The most important point here, is the Power Sharing, which is the process of how the power is shared between generators with the need to avoid the overload that exceeds the 10% at a certain period of time, as long as there is no overheat.

$$P_{\text{Load consumed}} = P_{\text{Tot}} = P_{G1} + P_{G2}$$

$$Q_{\text{Load consumed}} = Q_{\text{Tot}} = Q_{G1} + Q_{G2}$$

As the previous cases, this case has the same principle and the same equations (مبدأ التوازن), but they differ in the load: here, the loads are lamped, and there the loads are distributed on large areas.



Governor Set Point  $\uparrow$

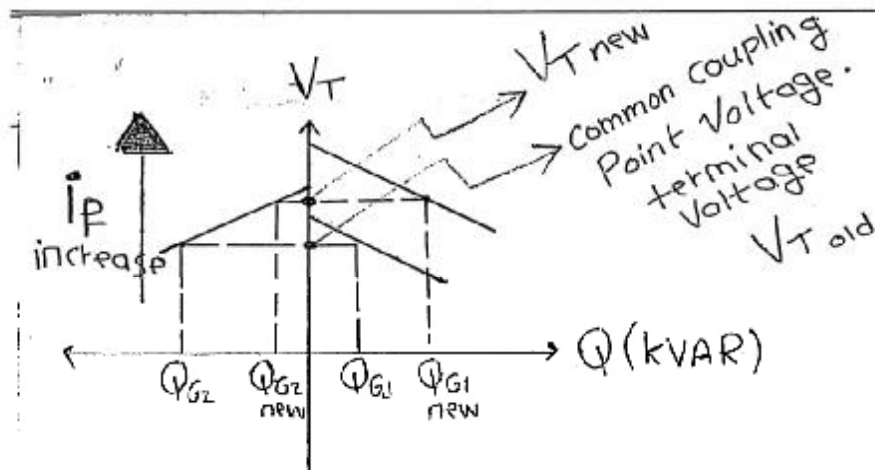
$f_{n1}$  increase

يترفع الـ C/C لقوة

Power  $\uparrow$  Power  $\downarrow$  Contribution  $\uparrow$   $\downarrow$

note: if the governor set point of the first generator increases, for the same load:  $P_{G1} \uparrow, P_{G2} \downarrow, f_{n1} \uparrow$ .  
وتنقل المخرج من الثاني.

\* Governor  $\leftrightarrow$  P \*



note: if the field current of the first generator increases,  $\therefore Q_{G1} \uparrow \rightarrow Q_{G2} \downarrow$

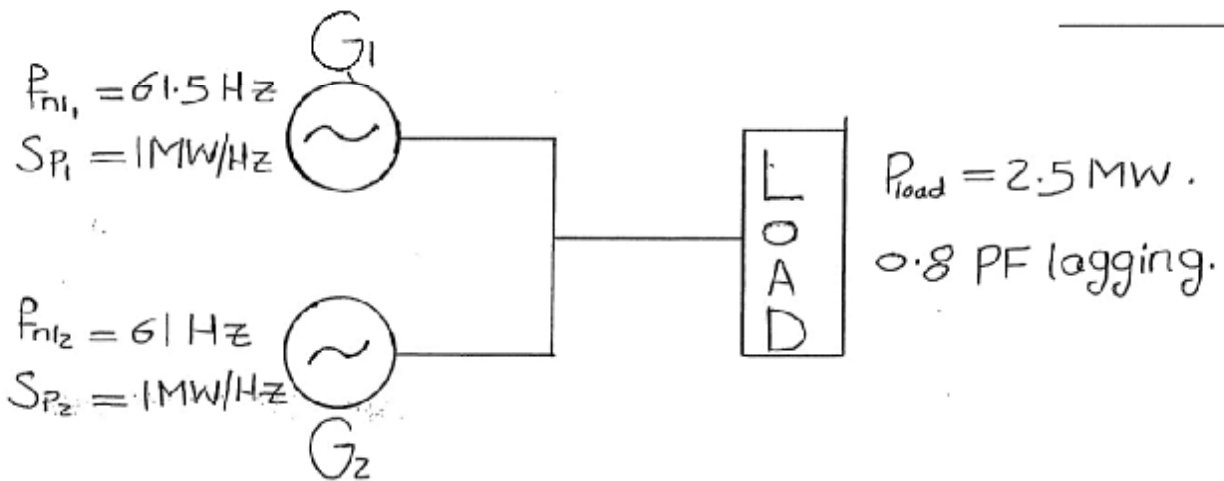
Terminal Voltage  $\uparrow$

Why we do this ?!

Because of the overload : So, if we have two generators, one of them running at the 20% of the overload and the other at 40%, then one turns off and the other runs at 60% of the over load. (توزيع الحمل)



Example :



a)  $f_{sys} = ?$

$$P_{TOT} = P_1 + P_2$$

$$= \underline{S_{P1} (f_{n1} - f_{sys})} + \underline{S_{P2} (f_{n2} - f_{sys})}$$

$$2.5 \times 10^6 = 1 \times 10^6 (61.5 - f_{sys}) + 1 \times 10^6 (61 - f_{sys})$$

$$\rightarrow f_{sys} = 60 \text{ Hz}$$

$$P_1 = 1.5 \times 10^6 \text{ W}$$

$$P_2 = 1 \times 10^6 \text{ W}$$

b) If an additional 1 MW is connected

$$P_{\text{Tot}} = P_1 + P_2$$

$$3.5 \times 10^6 = 1 \times 10^6 (61.5 - f_{\text{sys}}) + 1 \times 10^6 (61 - f_{\text{sys}})$$

$$\rightarrow f_{\text{sys}} = 59.5 \text{ Hz.}$$

$P \downarrow$ , as load  $\uparrow$

$$P_1 = 2 \text{ MW.}$$

$$P_2 = 1.5 \text{ MW.}$$

c) If the governor set point of  $G_2$  is increased by such that to have  $f_{n12}$  equals to 61.5 Hz.

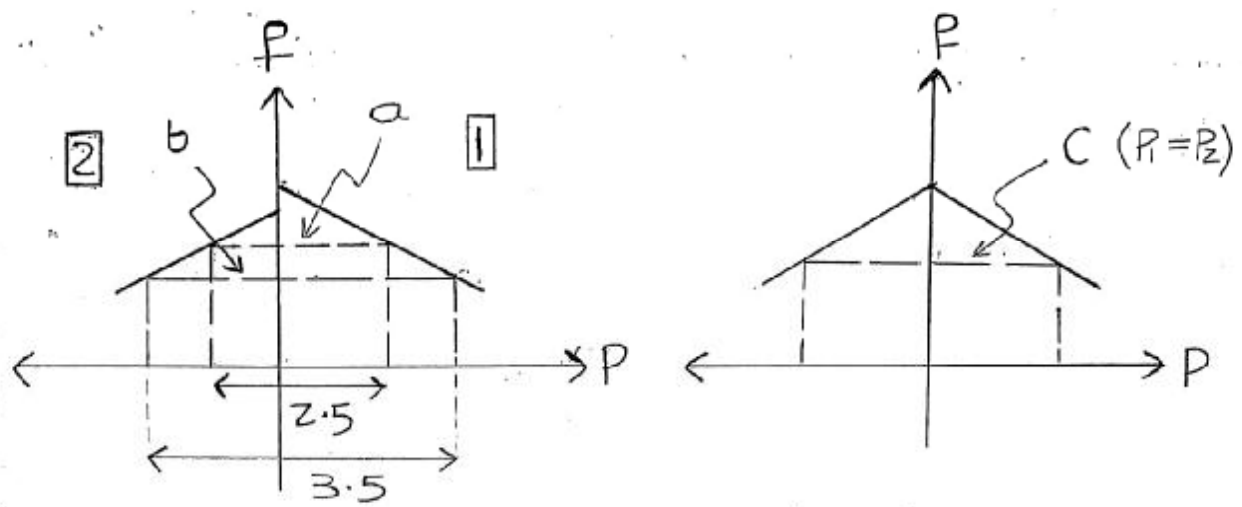
$$3.5 \times 10^6 = 1 \times 10^6 (61.5 - f_{\text{sys}}) + 1 \times 10^6 (61.5 - f_{\text{sys}})$$

$$\rightarrow f_{\text{sys}} = 59.75 \text{ Hz.}$$

$$P_1 = 1.75 \text{ MW.}$$

$$P_2 = 1.75 \text{ MW.}$$

} Same Sp, Same  $f_{n1}$   
 $\rightarrow$  Same P.



d) How to restore the nominal freq.?

→ increase the (no load freq.) of the

first generator or the second generator

or the Both.

e) How to restore the nominal freq.

keeping the "power sharing" equally?

$$3.5 \times 10^6 = [1 \times 10^6 (f_{n11} - f_{sys})] \times 2$$

$\uparrow$  60Hz       $\uparrow$  equals Power

$$f_{n11} = f_{n12} = \text{▲}$$

→ two generators connected in parallel have the same  $f_{sys}$ .

1] The Total Power Consumed by the load is supplied by the generators.

2] The System Freq. and the terminal Voltage (Common Coupling Point Voltage) depends on the Power demands.

3] To adjust the real Power Sharing between the generators without changing  $P_{sys}$  :

a- if the load increases, the governor set point of the first or the second or the both of two generators should increase.

b- if the load decreases, the governor set point of the first or the second or the both of two generators should decrease.

---

4] To adjust  $f_{sys}$  without changing the real power sharing between the generators :

a- if the  $f$  is to be increases , increase the governor set point of two generators.

b- if the  $f$  is to be decreases , decrease the governor set point of two generators.

5] To adjust the Reactive Power Sharing between generators without changing  $V_T$  :

a- if the  $Q$  demand increases , increase the field current of the first generator or the second or the both

b- if the  $Q$  demand decreases,  
decrease the field current of the  
first generator or the second or the both.

6] To adjust the  $V_T$  between generators  
without changing the reactive power  
sharing :

a- if the  $V_T$  is to be increased,  
increase the field current of the  
two generators.

b- if the  $V_T$  is to be decreased,  
decrease the field current of the  
two generators.

## Synchronous Generator Transients :

\* Transients Come as a result of :

1] Shaft torque is applied .

( i/P mech. power Changed ) .

$$\uparrow P_m = \uparrow T_m \cdot \omega$$

↑  
const.

↳ by governor set point. (Pule).

2] Output load Suddenly Changes .

\* demand Changes on P or Q .

\* Switching : P or Q ↓ , and

Connecting : P or Q ↑ .

3] Connecting generator in parallel with a running power Sys. (or disconnecting)

↳ \* Changes in the power Sharing .

4] Short circuits in any part of the Power System.

\* during fault  $\rightarrow$  S.C. between  <sup>$P=0$</sup>  3 lines or  
2 lines or line-ground.  
 $\rightarrow$  Changing the power flow.

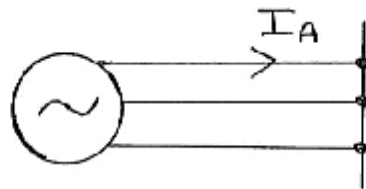
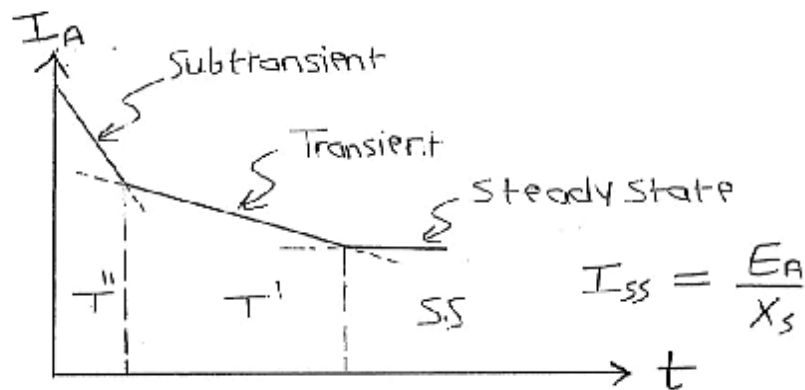
5] Field current or Field voltage is applied.  $\rightarrow$  [Field power is applied].

Also, loss of excitation  $\rightarrow$  rotor of the syn. gen. is open.  
( $I_f = 0$ )  
(fault).



# # Short CCT. Transient :-

(See the figure in book Pages:- 246-249, 5<sup>th</sup> Ed.)



rated power پر ماکسimum  
 generator آگبر  
 time کم

مکتبہ اسلامیہ خواتین  
 تعمیر - قمر عباسیہ - مکہ مکرمہ  
 ۰۷۷۲۲۵۰۷۷۹

4

## Synchronous Generator Ratings :

Ratings (Design Specifications) : Certain basic limits like speed and power that maybe obtained from the machine.

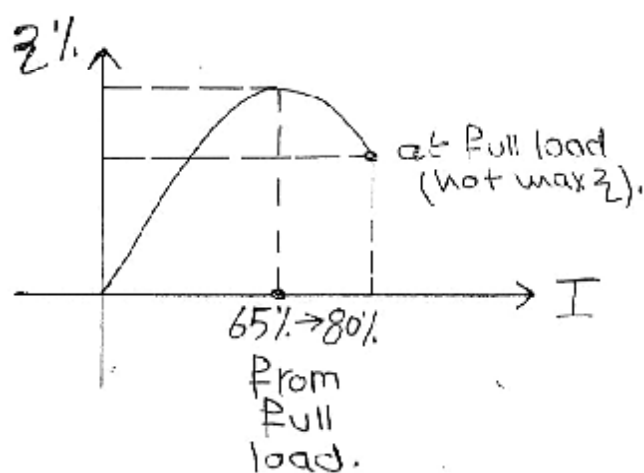
The purpose of the Ratings is to protect the machine from important operation.

Rated Power  $\uparrow$  , Size & weight  $\uparrow$

Rated Voltage  $\uparrow$  , Size & weight  $\uparrow$

Current  $\uparrow$  , Size, cost, weight & Volume  $\uparrow$

Rated Power  $\uparrow$  ,  $\eta \uparrow$  , losses  $\downarrow$  , as a ratio of  $\eta \uparrow$



\* Typical Rated Values :

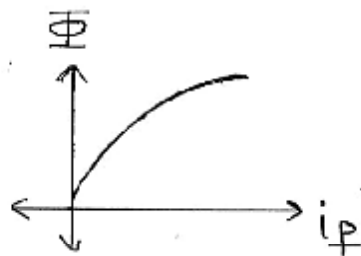
- 1 Voltage.      2 Frequency.      3 Speed.
- 4 Apperant Power.      5 P.F → load بحسب
- 6 Field current.      7 Armature Current.
- 8 Service factor.

→ most of them are coupled together.

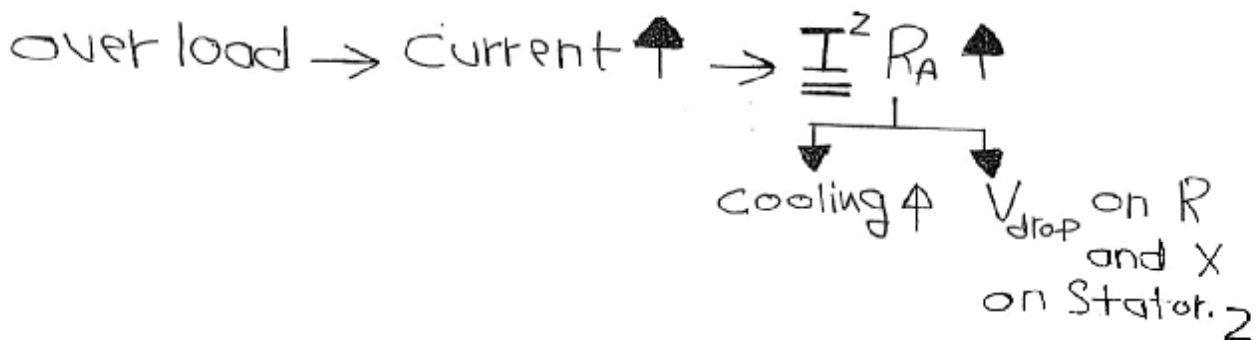
$$E = 4.44 \Phi f N k_w$$

$$f = \frac{n P}{120}$$

$$S = \sqrt{3} V_L I_L$$



# Full load Power الفكرة لانتجى ←





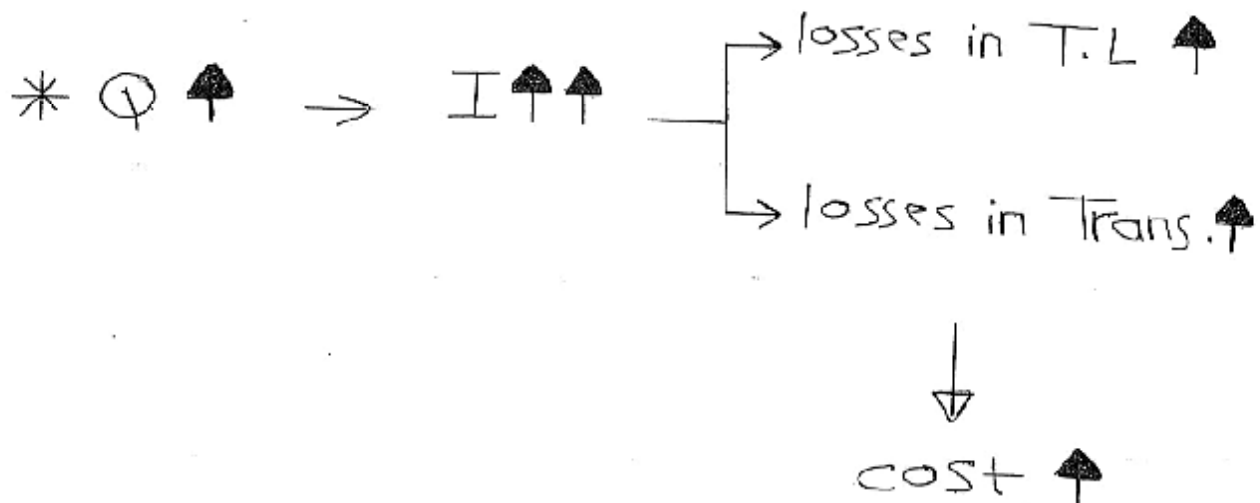
$$S = \sqrt{3} V_L I_L$$

@ Full load:

→  $P = 0$  : Pure inductive or capacitive load.

الاجزاء  
→  $\phi = 0$  : pure Resistive load.

$$[S = P]$$



Service Factor : The ratio of actual max. Power to its Name plate Rating. a generator with service factor of 1.1 can run at 110% of the rated load without damage (problems), if the overheat is accounted.

→ up normal conditions → underload تشغيل عادية  
مثلاً فوهة حره وتبريد سريع ⇒ 90% Full load.

\* because of that, Service factor related with Temp. ##.

SF=1.5 ← Full load 150% تبريد كويس وبدي اشغل ع 150% من Full load

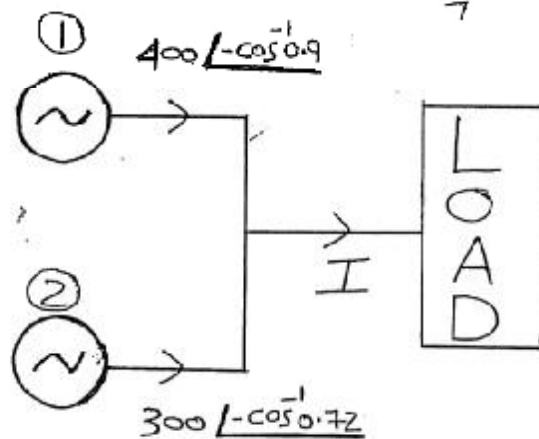
→ Transformer → due to Tap Changers.  
change #. of turns  
→ Generator → a problem is:  
saturation of the magn. cct.  
5

Question :-

both have the same  
Field current, and  
the same voltage.

600 kVA  
480 V

600 kVA  
480 V



Both Generator prime mover have  
difference Speed drop c/c .

a)

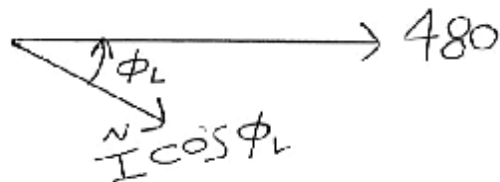
$$P_1 = \sqrt{3} V_L I_L \cos \phi = 0.3 \text{ MW.}$$

$$P_2 = \sqrt{3} V_L I_L \cos \phi = 0.18 \text{ MW.}$$

$$Q_1 = \sqrt{3} V_L I_L \sin \phi = 0.14 \text{ MVAR.}$$

$$Q_2 = \sqrt{3} V_L I_L \sin \phi = 0.173 \text{ MVAR.}$$

b) The overall P.F. on the load.?



$$\begin{aligned}\tilde{I} &= 400 \angle -\cos^{-1} 0.9 + 300 \angle -\cos^{-1} 0.72 \\ &= 691.5 \angle -33.6^\circ \text{ A.}\end{aligned}$$

$$\text{PF} = \cos(-33.6) = 0.833 \text{ lagging.}$$

c) Is the  $G_1$  run in rated power.?

$$P_1 = 300 \text{ k.}, \quad Q_1 = 144 \text{ k}$$

$$\sqrt{(P_1)^2 + (Q_1)^2} \stackrel{?}{=} S_{\text{rated}}$$

$$\underline{332.7 \text{ k}} \neq \underline{600 \text{ k}}$$

\* No,  $G_1$  not run in rated power.



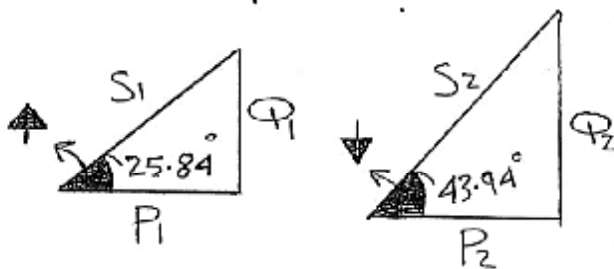
d) Is one Generator is enough?

$$\sqrt{(P_1 + P_2)^2 + (\Phi_1 + \Phi_2)^2} \stackrel{?}{=} S_{\text{rated}}$$

$$\underline{581.9k} < \underline{600k}$$

→ ∴ one G is enough.

e) In what direction must the field current on each generator be adjusted in order for them to run at the same P.F.?



$$\begin{aligned} \cos \phi_1 &= \cos \phi_2 \\ \cos \left[ \tan^{-1} \frac{\Phi_1}{P_1} \right] &= \cos \left[ \tan^{-1} \frac{\Phi_2}{P_2} \right] \\ \frac{\Phi_1}{P_1} &= \frac{\Phi_2}{P_2} \quad \text{--- [1]} \end{aligned}$$

# increase field current of the 1<sup>st</sup> gen. →  $\Phi_1 \uparrow$

$$\Phi_1 + \Phi_2 = 0.144 + 0.173$$

or: # decrease field current of the 2<sup>nd</sup> gen. →  $\Phi_2 \downarrow$

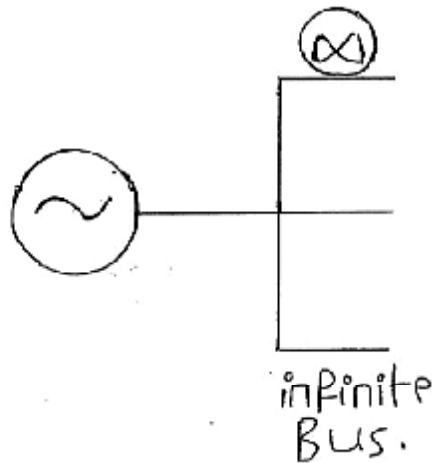
$$\Phi_1 + \Phi_2 = 0.317 \quad \text{--- [2]}$$

⋮

Question :

20 MVA , 13.8 kV , 0.8 PF lagging -

Y connected , Synchron. gen. ,  $R_A = 0$  ,  $X_s = 0.7$  pu



rated values  
استعملنا  
basis 1

a)

$$S_{\text{base}} = \sqrt{3} V_{L,\text{base}} I_{L,\text{base}}$$

$$20 \times 10^6 = \sqrt{3} (13.8 \times 10^3) I_{\phi,\text{base}}$$

$$\rightarrow I_{\text{base}} = 836.74 \text{ A}$$

$$Z_{\text{base}} = \frac{V_{\phi,\text{base}}}{I_{\phi,\text{base}}} = \frac{13.8 \times 10^3 / \sqrt{3}}{836.74} = 9.5 \Omega$$

$$X_s = (0.7)(9.5) = 6.67 \Omega$$

$$b) \tilde{E}_A = ? \text{ in p.u. !}$$

$$\tilde{E}_A = \tilde{V}_\phi + j \tilde{I}_\phi X_s$$

$$= 1 \angle 0 + j(1 \angle -\cos^{-1} 0.8)(0.7)$$

$$= 1.526 \angle 21.52^\circ \text{ p.u.}$$

$$c) |\tilde{E}_A| = (1.526) V_{\phi, \text{base}}$$

$$= (1.526) \times \frac{13.8 \times 10^3}{\sqrt{3}} = 12.15 \text{ kV.}$$

d) If the internal generated voltage is reduced by 5%, Then calculate

$I_\phi$  ?

↑  
or  $I_p$   
(نفس الاستيعاب)

$$E_{A2} = 0.95 E_{A1}$$

$$\rightarrow E_{A1} \sin \delta_1 = E_{A2} \sin \delta_2$$

$$E_{A1} \sin(21.52^\circ) = 0.95 E_{A1} \sin \delta_2$$

$$\rightarrow \delta_2 = 22.71^\circ$$

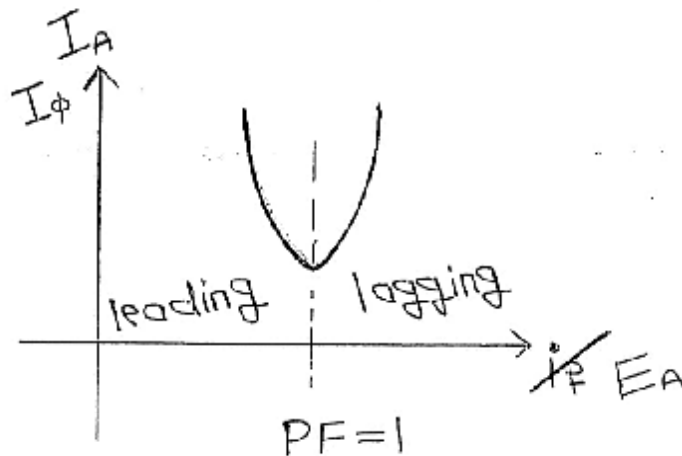
$$* \tilde{E}_A = \tilde{V}_\phi + j \tilde{I}_\phi X_s$$

$$0.95 * 1.526 \angle 22.7^\circ = 1 \angle 0 + j \tilde{I}_\phi (0.7)$$

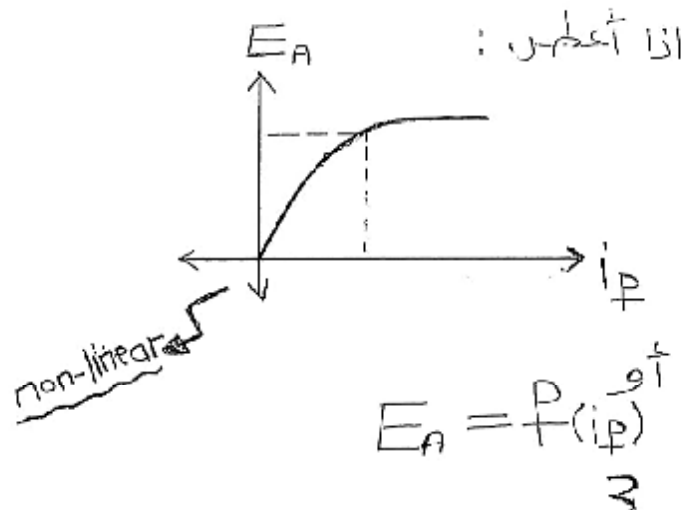
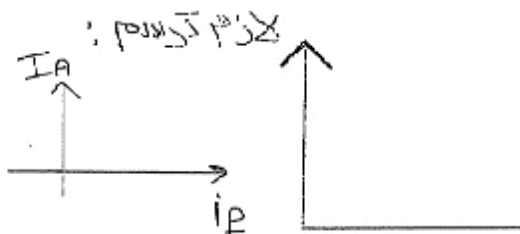
$$\rightarrow I_\phi = 0.93 \angle -31.1^\circ \text{ A.}$$

e) Repeat the previous part (d), for 10%, 15%, 20%, 25%.

\* reduction in  $E_A$



| $I_\phi$ | $E_A$ |
|----------|-------|
|          |       |

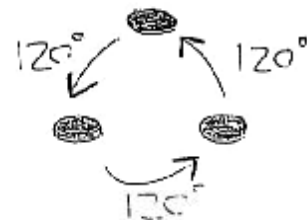


# Synchronous Motors :-

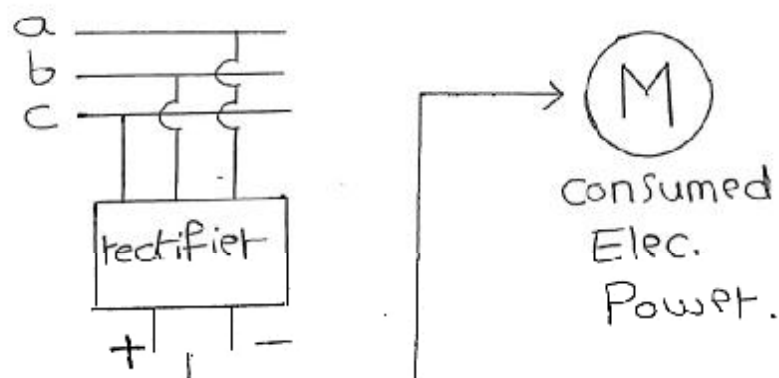
Types:

- 1] Salient pole rotor.
- 2] Cylindrical rotor.

\* The Stator of Synch. motor is the same as that of the induction motor i.e. three phase winding phase shifted by  $120^\circ$  in Space.



\* DC Power Supply is fed to the rotor.



3 $\phi$  Synchron. machines  $\rightarrow$  rotor  $\rightarrow$  DC.

## \*Principle of Operation :-

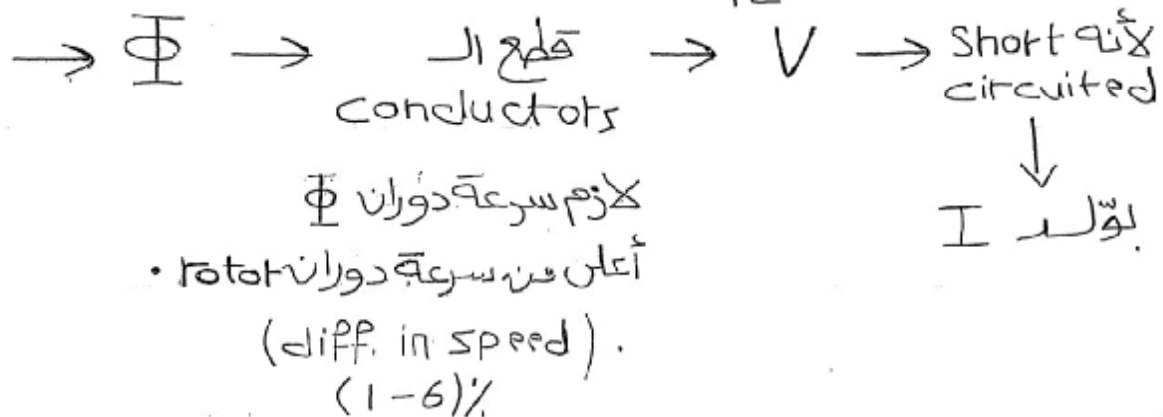
in Induction motor:

3 $\phi$  currents  $\rightarrow$  3 $\phi$  windings  
120 $^\circ$  phase in time      120 $^\circ$  phase in space.

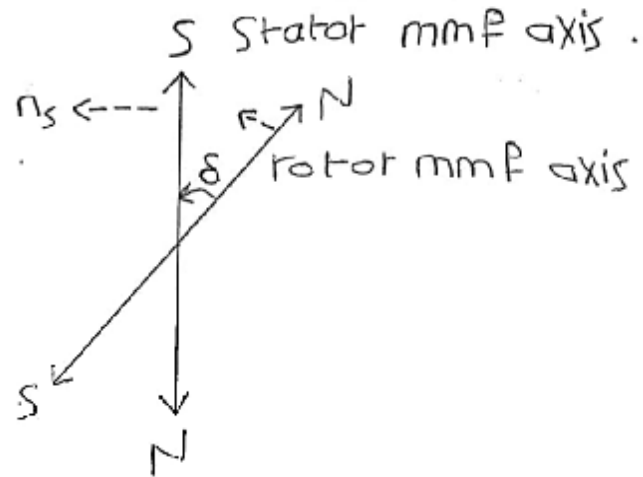


Rotating mag. field  
in Synchron. Speed

$$f = \frac{n_s P}{120^\circ}$$



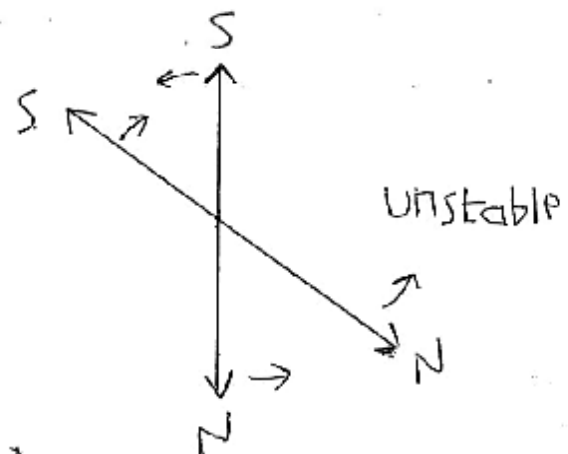
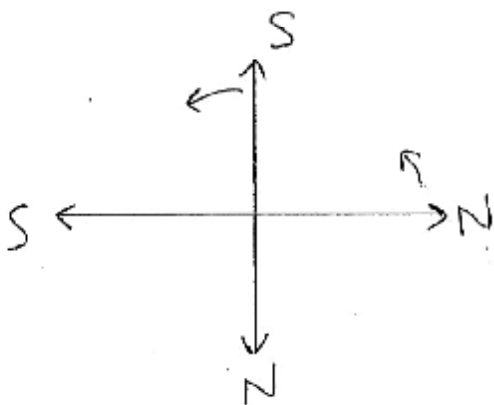
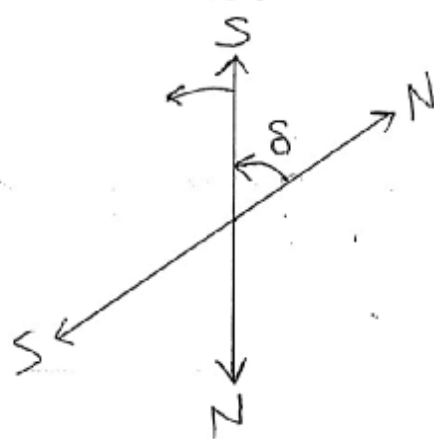




\*  $\delta = 0$  at no-load

كل ما زاد ال load  $\delta$  ↑

والعلاقة بينهم  $\sin$



\* stable  $\rightarrow \delta < 90$  }  $\delta = 90 \rightarrow$  marginally stable  
 \* unstable  $\rightarrow \delta > 90$  } uns.  $\leftarrow$  disturbance also 4



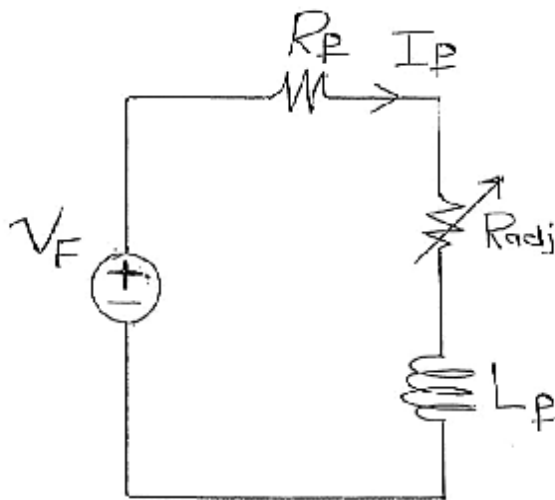
\* This motor is not self-starting  
due to inertia of the rotor.

↑  
[mass + dimension]

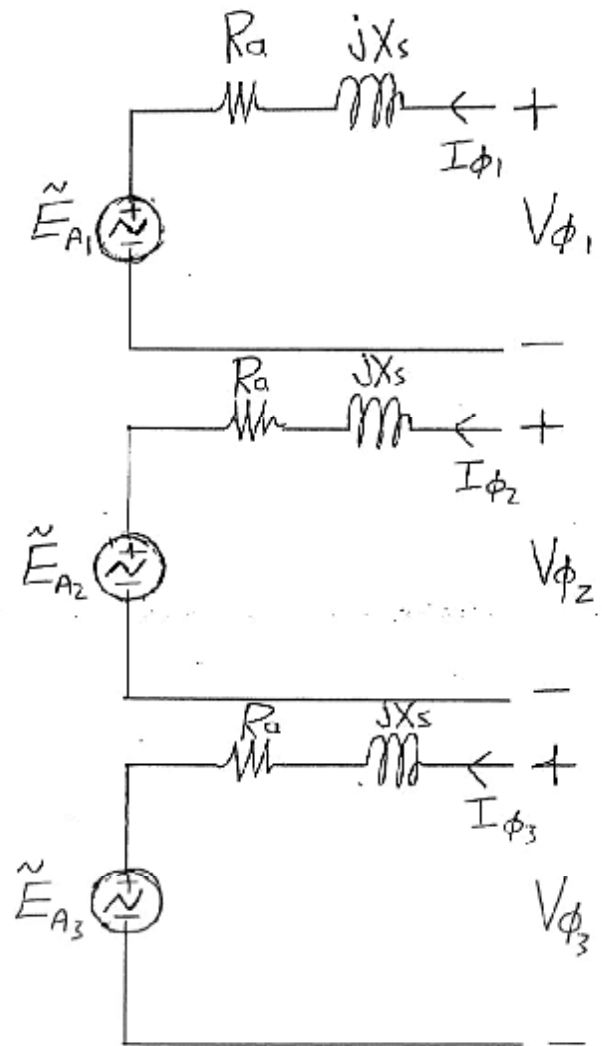
[Step Change  $\omega$   
of Speed]

# Equivalent circuit of Synch. Motor :-

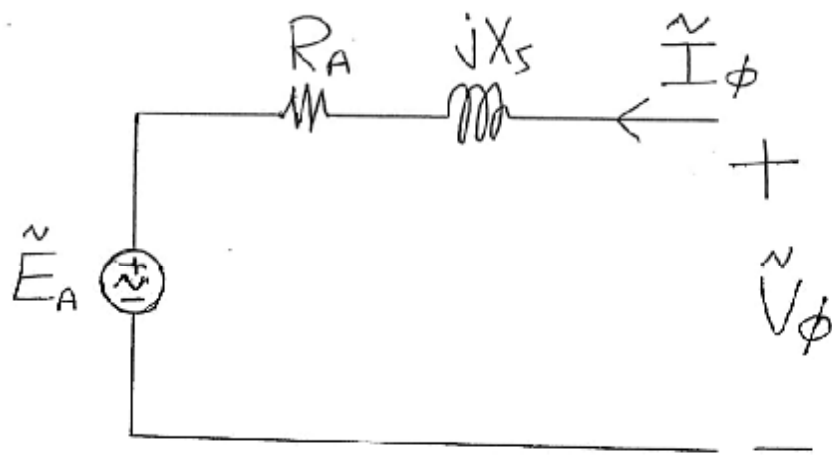
Motor :-



Field Ckt.



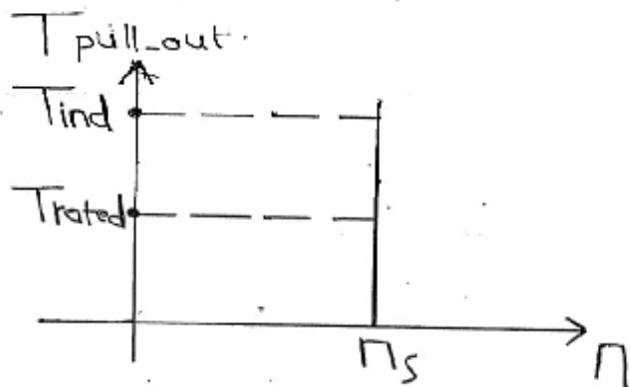
Armature Ckt.



$$\tilde{E}_A = 4.44 \Phi f N k_w$$

$$\tilde{V}_\phi = \tilde{E}_A + \tilde{I}_A (R_A + jX_s)$$

\* Torque - Speed Characteristics:  
[Steady-state]

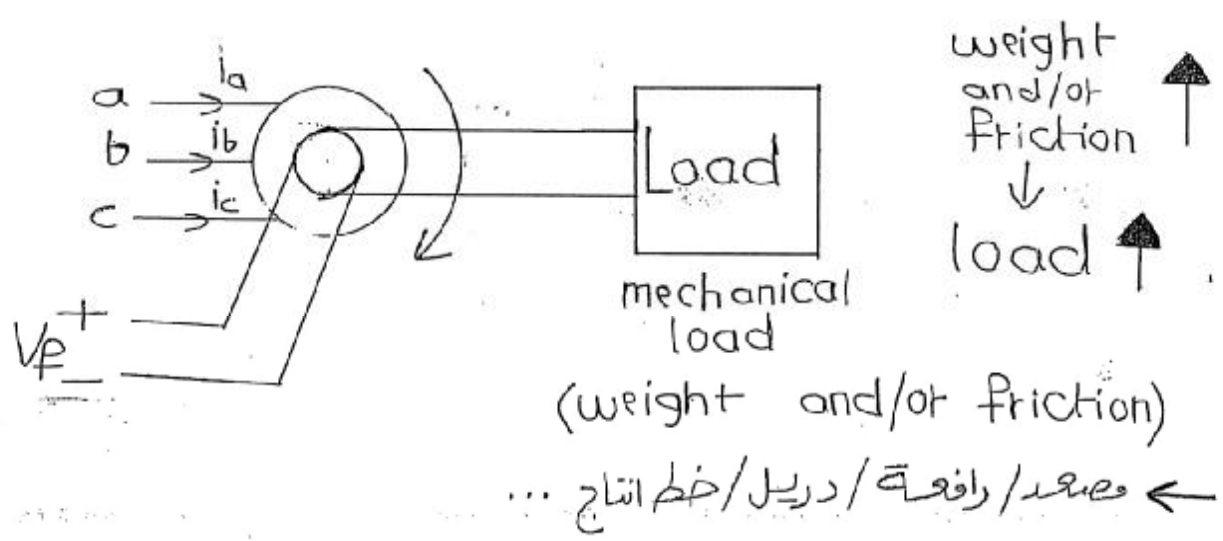


$$T_{\text{pull-out}} = T_{\text{max}} \Big|_{\delta=90^\circ} = \frac{3 V_\phi E_A}{\omega_s X_s}$$

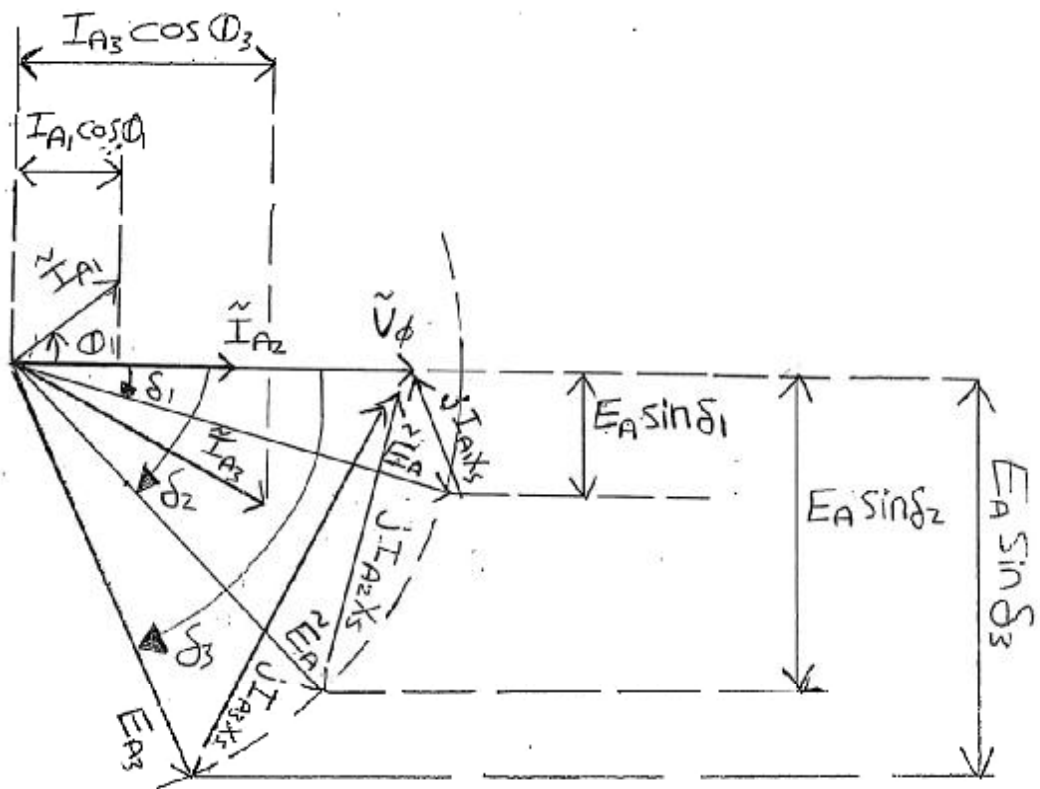
$\delta=90^\circ$  marginally stable.

$$T_{\text{ind}} = \frac{3 V_\phi E_A}{\omega_s X_s} \sin \delta$$

# \* Effect of load Changes on Synchron. motor :-



- 1-  $E_A$  is constant  $\rightarrow$   $I_f$  is constant
- 2-  $V_\phi$  is constant.
- 3-  $T_{ind} \uparrow = \frac{3 V_\phi E_A \sin \delta}{\omega X_s} \uparrow \rightarrow E_A \sin \delta$   
 $\hookrightarrow$  const.



as the load increases  $\left[ (T_L) \uparrow \right]$ ,

$T_{ind} \uparrow$ ,  $\sin \delta \uparrow$ ,  $I_A \uparrow$ ,

$E_A \sin \delta \uparrow$ , P.F. motor moves towards lagging.

\* ممکن ہے leading لیکن یقیناً نہیں

Example: 208 V, 45 kVA, 0.8 PF  
 leading,  $\Delta$ -connected Stator 60Hz  
 $X_s = 2.5 \Omega$ ,  $R_A = 0$ ,  $P_{F\&W} = 1500 \text{ W}$ ,  
 $P_{\text{core}} = 1000 \text{ W}$ .

$P_{\text{out}} = 15 \text{ hp}$  at 0.8 PF leading } initial  
 o.p

a)  $\tilde{E}_A = ?$

$$P_{\text{in}} = P_{\text{out}} + P_{\text{losses}}$$

$$= 15 \times 746 + 1500 + 1000$$

$$= 13.7 \text{ kW}$$

$$P_{\text{in}} = \sqrt{3} V_L I_L \cos \theta$$

$$13.7 \times 10^3 = \sqrt{3} \times (208) I_L (0.8)$$

$$\rightarrow I_L = 47.5 \text{ A} \quad \div \sqrt{3}$$

$$\tilde{I}_\phi = 27.4 \angle \cos^{-1} 0.8$$

$$\tilde{E}_A = \tilde{V}_\phi - j \hat{I}_A X_s$$

$$= 208 \angle 0 - j [(27.4) \angle \cos^{-1} 0.8] (2.5)$$

$$= 255 \angle -12.4^\circ \text{ V}$$

b) if the load Torque becomes  
30 hp ?!

$$P_{in} = 30 \times 746 + 1500 + 1000$$

$$= 24.9 \text{ kW.}$$

$$P_{in} = \sqrt{3} V_L I_L \cos \phi$$

$$24.9 \times 10^3 = \sqrt{3} (208) \underline{I_L} \cos \underline{\phi} \quad \text{--- 2 unknowns}$$

$$\rightarrow P_{in} = \frac{3 V_{\phi} E_A}{X_s} \sin \delta \quad \checkmark$$

$$24.9 \times 10^3 = \frac{3(208)(255)}{2.5} \sin \delta$$

$$\rightarrow \delta = \underset{\substack{\uparrow \\ \text{motor}}}{23}^{\circ}$$

$$\tilde{E}_A = \tilde{V}_{\phi} - j \hat{I}_A X_s$$

$$255 \angle -23^{\circ} = 208 \angle 0 - j \tilde{I}_A (2.5)$$

$$\rightarrow I_A = 41.2 \angle +15^{\circ} \text{ A.}$$

$$\text{PF} = \cos 15^{\circ} = 0.96 \text{ leading.}$$

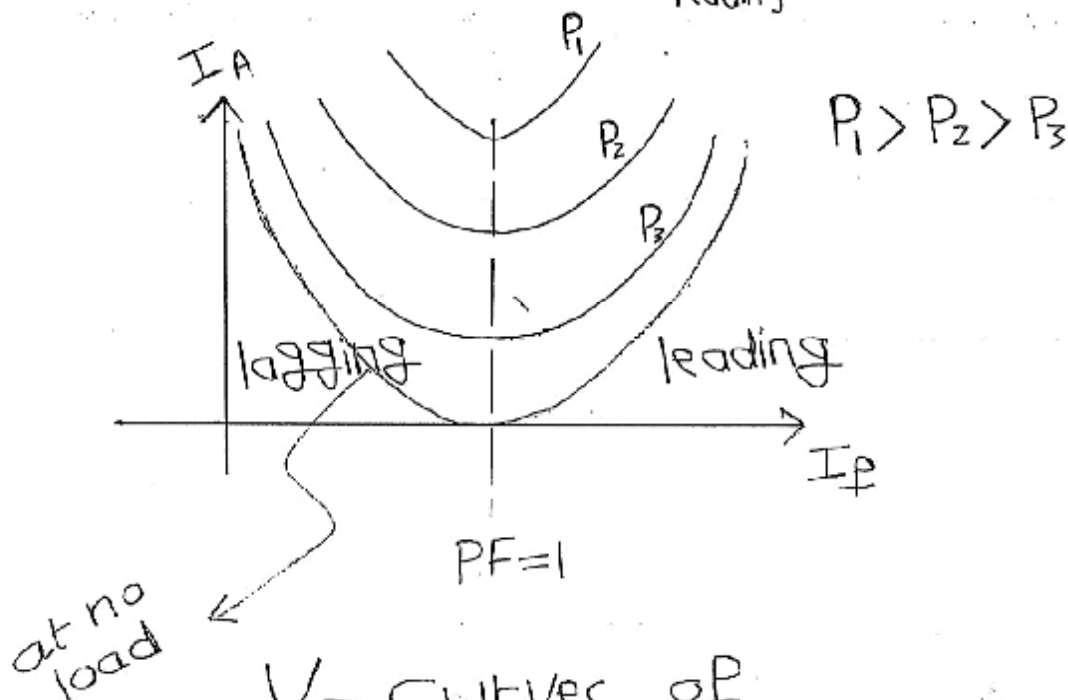
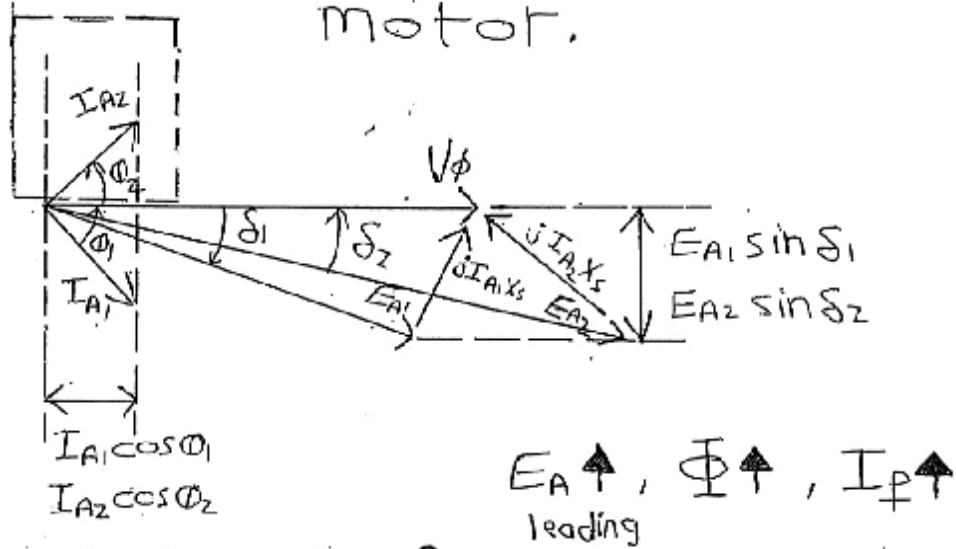




$$P_{in} = 3 V_{\phi} I_A \cos \theta$$

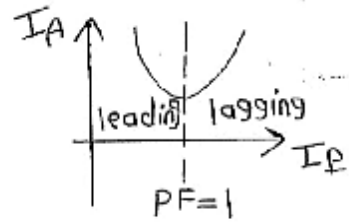
↳  $I_A \cos \theta$  : const.

overexcited synch. motor.

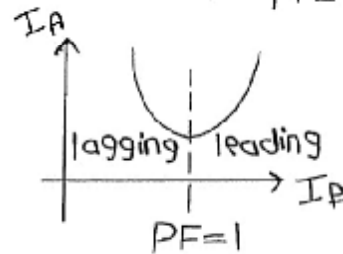


V-curves of  
Synch. motor.

\* Synch. genetrators



\* Synch. motots



→ PF of the motor can be controlled.

: ad v. : leading بالسيارة

→ \* Can be used as P.F correction in overexcited Synch. motor.

motor

corrector

controlable

\* To get V-curves :

- ①  $P_{out}$  Const.
- ②  $V_T$  Const.
- ③ Speed Const.
- ④  $I_f$  Variable.

Example: The Same motor of the previous example.

$P_{out} = 15 \text{ hP} ; \text{ PF} = 0.85 \text{ lagging}$

↑  
initial  
O.P

a)  $\tilde{E}_A = ?$

$$P_{in} = \sqrt{3} V_L I_L \cos \theta$$

$$(15 \times 746) + 1500 + 1000 = \sqrt{3} (208) I_L (0.85)$$

$$\rightarrow \tilde{I}_\phi = 25.8 \angle -31.8^\circ \text{ A.}$$

$$\begin{aligned}\tilde{E}_A &= 208 \angle 0 - j [(25.8) \angle -31.8^\circ] (2.5) \\ &= 182 \angle -17.5^\circ \text{ V.}\end{aligned}$$

b) If the flux is increased by 25%.

$$\tilde{I}_A = ?$$

$$\rightarrow E_{A1} \sin \delta_1 = E_{A2} \sin \delta_2$$

$$E_{A1} \sin(-17.5) = 1.25 E_{A1} \sin \delta_2$$

$$\rightarrow \delta_2 = -14^\circ$$

$$182 \angle -14^\circ = 208 \angle 0 - j \tilde{I}_A (2.5)$$

$$\rightarrow \tilde{I}_A = 22.5 \angle 13.2^\circ \text{ A.}$$

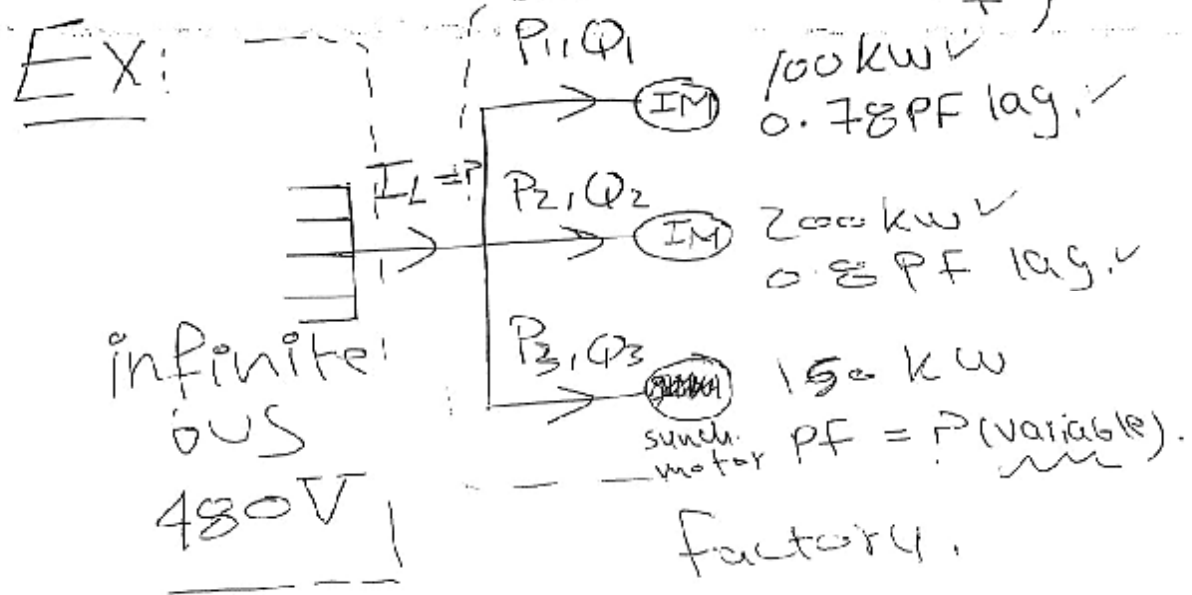
$$\text{PF} = \cos 13.2 = 0.97 \text{ leading.}$$

c) What is the minimum value of  $I_A$ ?

@ unity PF :-

# The Synchronous Motor <sup>الموتور المتزامن</sup> and Power Factor Correction:

\* ~~overexcited~~ Synchron. motor runs with leading Power factor therefore it can be used as ~~power factor corrector~~. The good thing is that its power factor is ~~adjustable~~ (due to change  $i_f$ ).



Q) If the sync. motor is adjusted to run at 0.85 PF lagging. IEP

$$\begin{aligned} \Phi_1 &= \tan \phi_1 P_1 \\ &= \tan [\cos^{-1} 0.78] [100] \\ &= 80.2 \text{ kVAR.} \end{aligned}$$

$$\begin{aligned} \Phi_2 &= \tan \phi_2 P_2 \\ &= \tan [\cos^{-1} 0.8] [200] \\ &= 150 \text{ kVAR.} \end{aligned}$$

$$\Phi_3 = 93 \text{ kVAR.}$$

$$P_{\text{tot}} = P_1 + P_2 + P_3 = 450 \text{ kW}$$

$$\Phi_{\text{tot}} = \Phi_1 + \Phi_2 + \Phi_3 =$$

$$\phi_{\text{tot}} = \tan^{-1} \frac{\Phi_{\text{tot}}}{P_{\text{tot}}}$$

=

$$\cos \phi_{\text{tot}} = 0.812 \text{ lag.}$$

√2

$$P_{\text{tot}} = \sqrt{3} V_L I_L \cos \phi_{\text{tot}}$$

$$450 \times 10^3 = \sqrt{3} (480) I_L (0.812)$$

$$\rightarrow I_L = 667 \text{ A. } \#$$

b) If the synchron. motor is adjusted to run at

0.85 leading P.F.  $I_L = P$

# No change in  $\phi_1, \phi_2$

$$\phi_3 = \tan^{-1} P_3 = -93 \text{ kVAR.}$$

$$\# P_{\text{Tot}} = P_1 + P_2 + P_3 = 450 \text{ kW}$$

$$\# \phi_{\text{Tot}} = \phi_1 + \phi_2 + \phi_3 = 137.2 \text{ kVAR}$$

$$PF_{\text{tot}} = 0.957 \text{ lags}$$

$$P_{\text{tot}} = 450 \times 10^3 = \sqrt{3} (480) I_L$$

$$I_L = 566 \text{ A. } (\cos 0.957) \sqrt{3}$$





Synch. motor  
cost ↑  
bulky size

- i/v of  $P$  is limited - i/v of  $P$



cap. banks  
in parallel.

## # Starting of synch motor :-

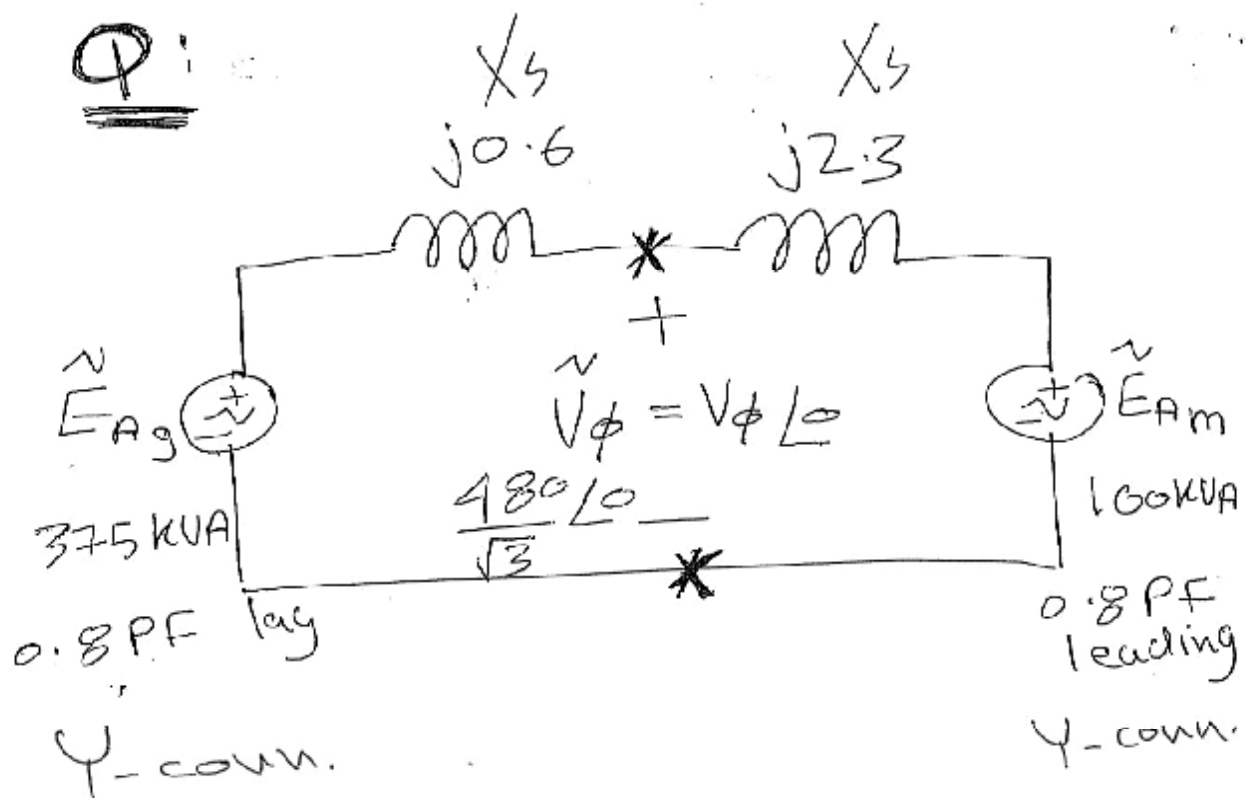
Motor will follow up due to

① ~~starting~~  $\equiv$  starting of all types of mechanical engine  
external prime mover (motor) #

② rotating mmf  $\rightarrow$  due to  $P$  #  
gradual inc. of speed  $\rightarrow$   $P = \frac{nsP}{120}$

③ Utilizing the damper windings.  $\rightarrow$  indu. motor

# Starting @ 5% induction and running @ 5% induction motor



$\tilde{V}_\phi = \frac{480}{\sqrt{3}} L_0$  when the motor  
 draws the rated power @ PF=1  
 D.O.P

$$P_m = \sqrt{3} V_L I_L \cos \phi$$

$$100 \times 10^3 = \sqrt{3} (480) I_\phi$$

$$\tilde{I}_\phi = 120.3 L_0 A.$$

6

$$\begin{aligned} \vec{E}_{Ag} &= \vec{V}_\phi + j \vec{I}_\phi 0.8 \\ &= 286.4 \angle 14.6^\circ \text{ V} \end{aligned}$$

$$\Rightarrow V_\phi = \vec{E}_{Am} + j \vec{I}_\phi 2.3$$

$$\vec{E}_{Am} = 391.6 \angle -44.9^\circ \text{ V}$$

b) If the flux of the motor

increases by 10%, then

calculate the new terminal voltage.

$$\Rightarrow E_{Am1} \sin \delta_1 = E_{Am2} \sin \delta_2$$

$$E_{Am1} \sin(-44.9^\circ) = (1.1) (E_{Am1}) \sin \delta_2$$

$$\delta_2 = -39.9^\circ$$

$$P_2 = P_1 = 100 \times 10^3 = \frac{(3)(V_{\phi 2})(1.1)(391.6)}{2.3}$$

$$V_{\phi 2} = 277.4 \text{ V} = \frac{480}{\sqrt{3}} * \frac{\sin(-39.9^\circ)}{\sqrt{7}}$$

$$I_{\phi} = \frac{277.4 \angle 0 - (1.1)(391.6) \angle -34.9^{\circ}}{2.3}$$

$$\underline{\underline{PF_{\text{new}}}} = \cos \angle =$$

\* Static stability power limit

$$= P_{\text{max}}$$

$$= \frac{3 V_{\phi} E_A}{X_S} \quad \#$$

100

1

Permanent Magnet (PM) materials and Permanent Magnet (PM) machines:

\* The Flux in PM machines is established by Permanent magnets. Both internal generated Voltage and induced Torque depends on the value of the magnetic Flux.

$E_A = k\Phi\omega$  --- internal generated Voltage.

$T_{ind} = k\Phi I_a$  --- induced Torque.

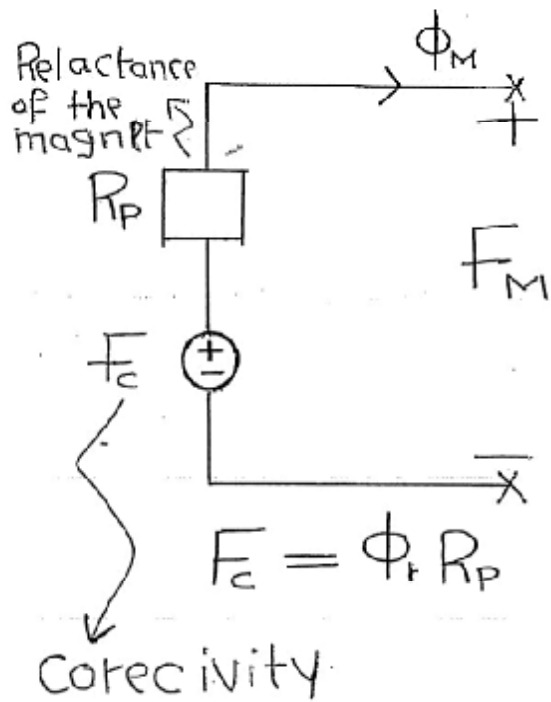
PM machine is Considered as the Source of the magnetic flux  $\Phi$ .

| Magnetic cct.                   | Electrical cct.          |
|---------------------------------|--------------------------|
| Flux (wb)                       | Current (A).             |
| MMF (A·turns)                   | EMF (V).                 |
| Reluctance (A/wb)               | Resistance ( $\Omega$ ). |
| $MMF = NI$<br>or $MMF = \Phi R$ |                          |

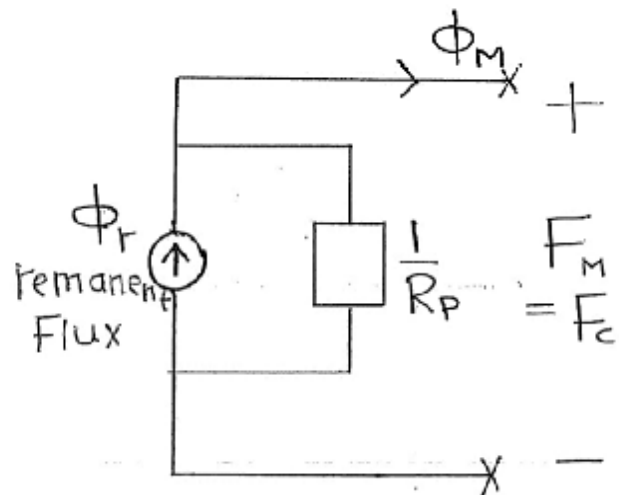
The main difference :  
 Electrical CCT. is linear.  
 Magnetic CCT. is non-linear.

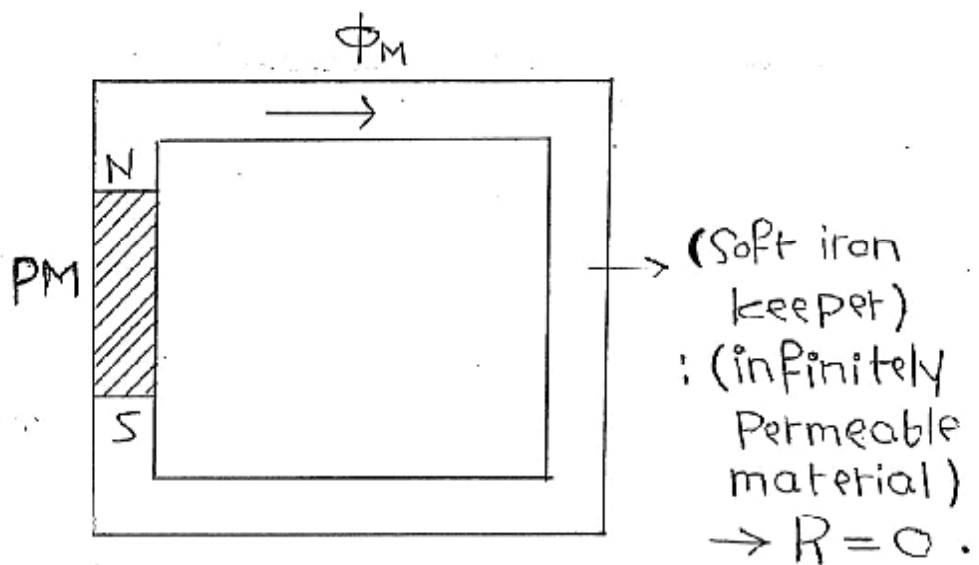
PM Can be represented by :

Thevenin Eq.



Norton Eq.





The magnet can be short circuited by connecting a material with infinite permeability ( $R = 0$ ).

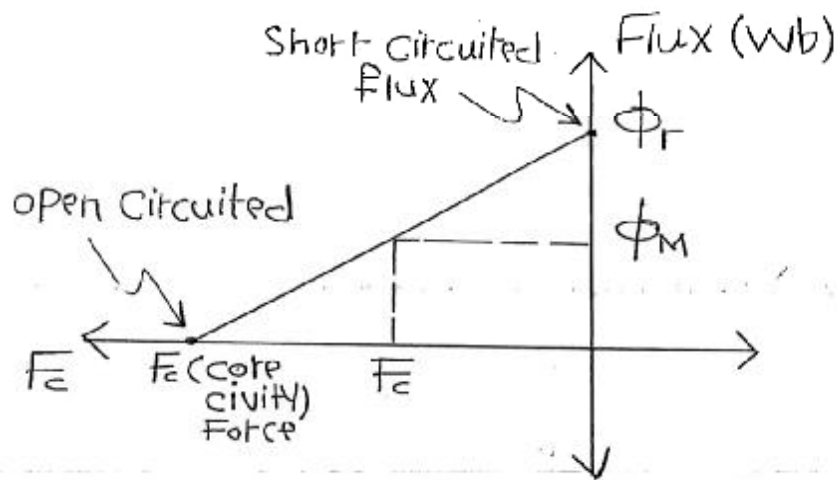
$$\rightarrow \Phi_r = \Phi_M$$

The magnet is open circuited if

$$\Phi_M = 0 \rightarrow R = \infty$$

$F_c$  is called Coercive MMF as it is MMF required to <sup>Force</sup> Coerce the magnet to produce Zero Flux. It directly expresses the resistance of the magnet against demagnetization.

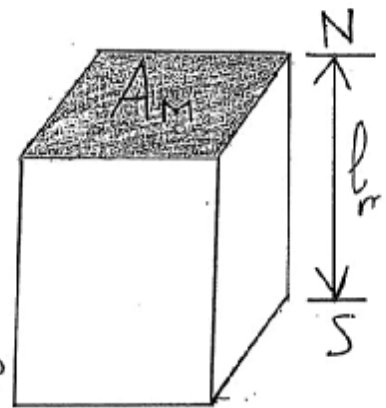
\* The Characteristics of PM :



$\therefore$  PM  $\downarrow$  Field cct. مميزات إستبدال  
 losses  $\downarrow$ , design  $\downarrow$ , Size  $\downarrow$ ,  $Z \uparrow$ ,  
 weight  $\downarrow$ , Copper  $\times$ , winding  $\times$ .  
عيوب :  
 cost  $\uparrow$ ,  $\phi$  const: No controllability.  
4



The remanent Flux  $\phi_r$  & Corecive MMF depend not only on the material properties but also on the dimensions of the magnet.



$$\phi_r = B_r \cdot A_m$$

$\downarrow$  material property       $\hookrightarrow$  magnet Cross Sectional area dimension.

$$F_c = H_c \cdot l_m$$

$\downarrow$  corectivity       $\hookrightarrow$  length of the materia in the direction of magnetization.

$\downarrow$  Corectivity Force

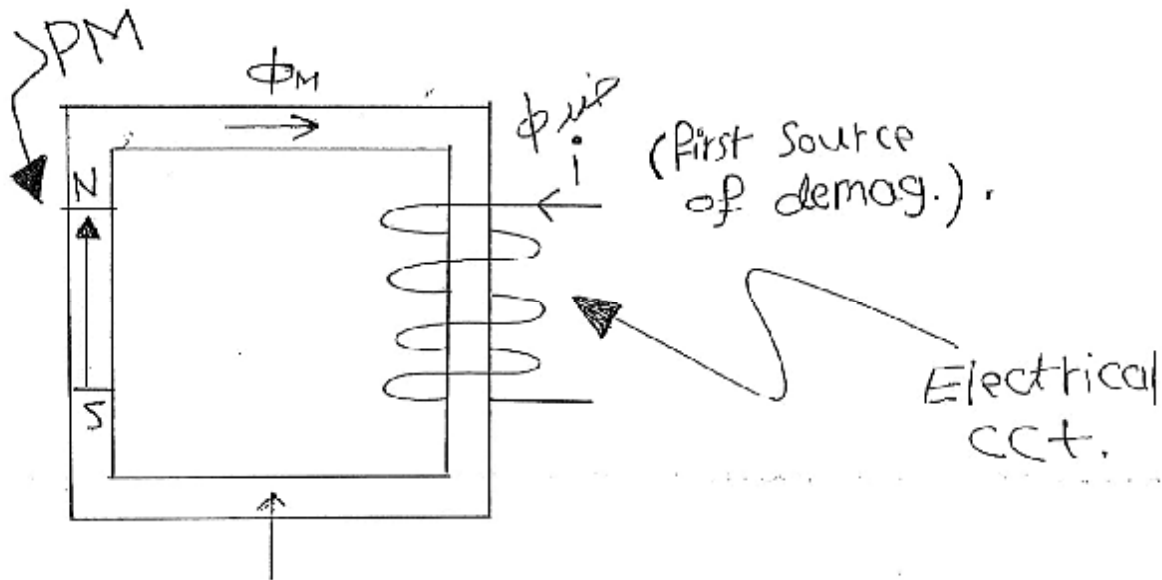
\* Electrical cct.  $\rightarrow$  current in Statot [diff in Voltage]

$\downarrow$   
 $\phi \Rightarrow$  diff in MME.

if Fault or Short cct.

$\rightarrow$  current  $\uparrow$   
 $\hookrightarrow$  First source of demag.

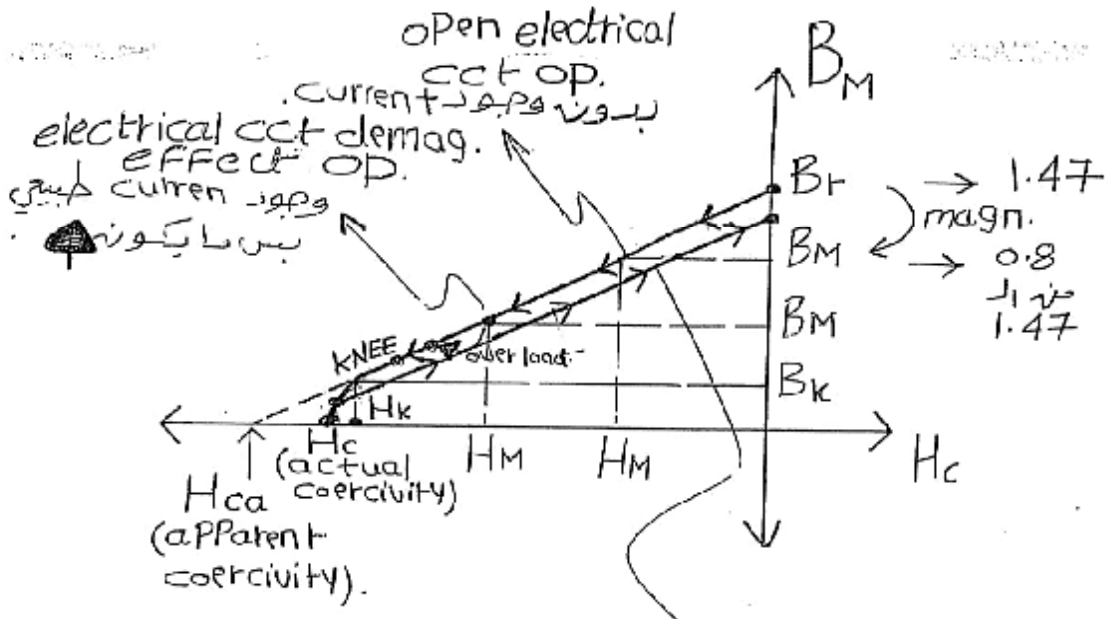
If the vertical axis is scaled by  $1/A_m$  and the horizontal axis by  $1/L_m$ , then the result will be a relation between  $B_m$  &  $H_m$ .



$\mu_r = (2000 - 5000)$  Ferro. mag.

$R = \frac{l}{\mu_r \mu_0 A}$   
 $\phi_{rip}$

(Second source of demag.).



Effect of demagnetizing on the C/C of PM mat.

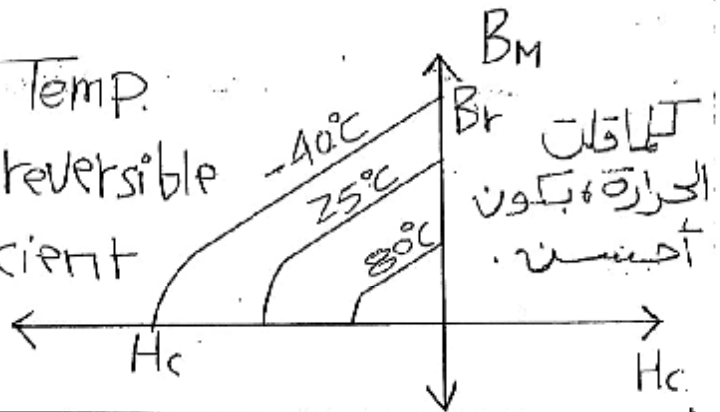
في حالة short cct.  

$$H_{sc} = NI \rightarrow$$
 كنه جاريات  
 KNEE JI

$$B_M = \mu_{rec} H$$

↑  
 recoil permeability  
 (PM it self)  
 (1.03 → 1.05)

The effect of Temp. is specified by reversible temperature coefficient

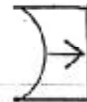
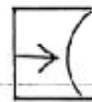


$\alpha_{Br}$

$$B_r(T) = B_r(20^\circ) \times \left[ 1 + \alpha_{Br} \frac{(T-20)}{100} \right]$$

Provided by the manufacturer.

\* Shapes of Perment magnets :-



بسبب وجودهم في أماكن رطبة.

لحمايته من الصدأ الذي يسبب

① التآكل

② يقلل من الـ E

المخزنة فيهم

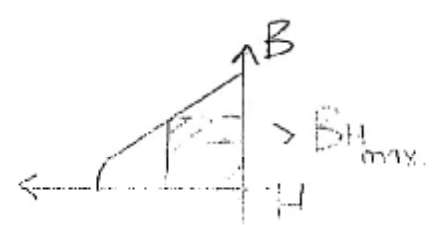
Coated by Al 0.5 mm

## Types :-

Ferrites } inexpensive.  
AlNiCo }

SmCo } rare-earth materials.  
NdFeB }

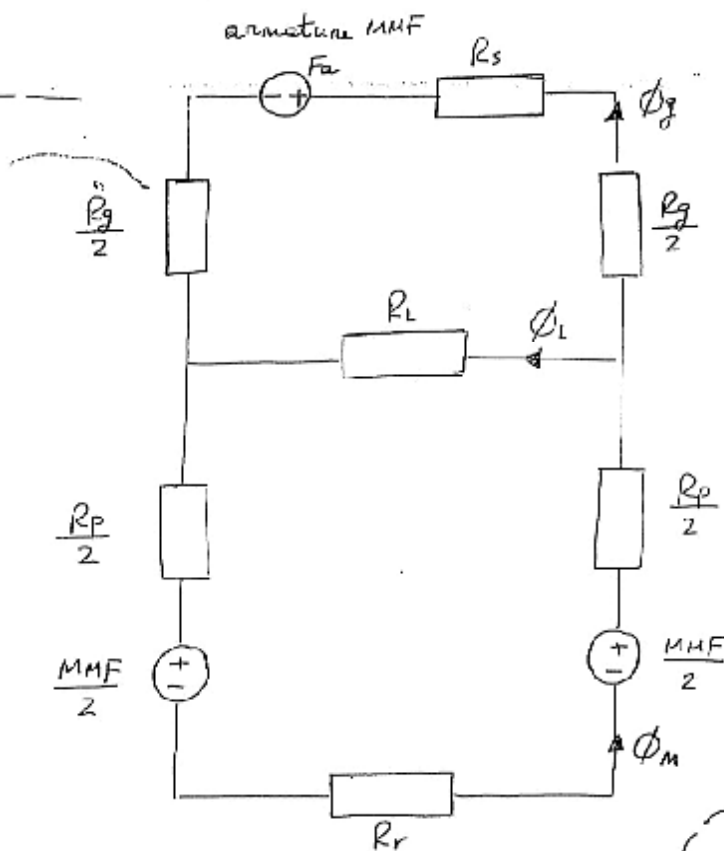
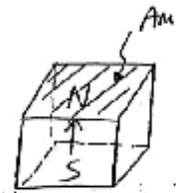
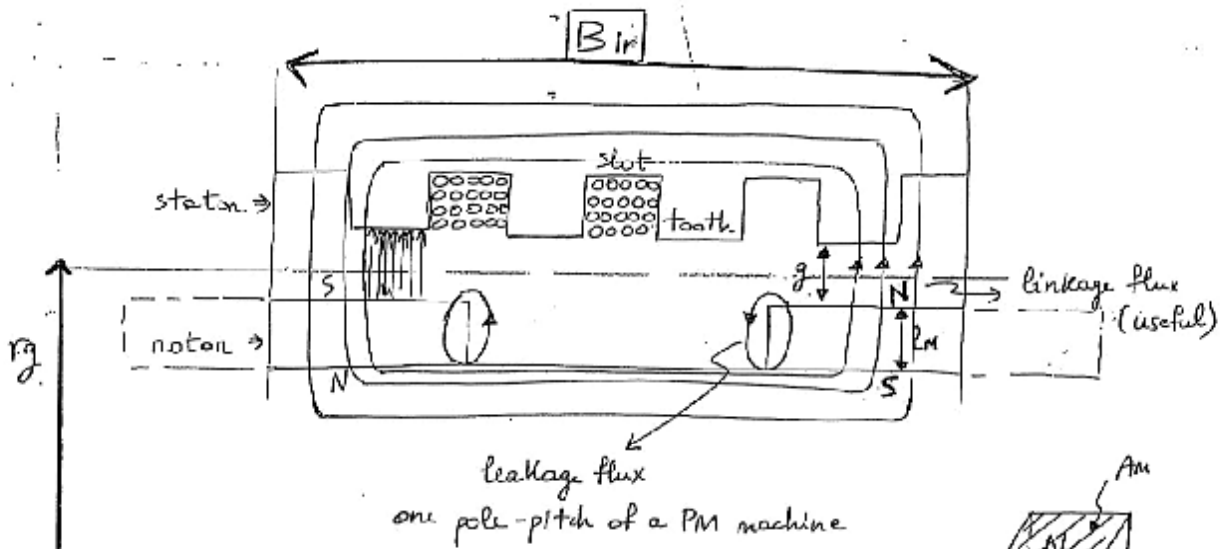
→ highest  $B_H^{(max)}$  → indicates to the energy stored in magnet.  
↑  
max. energy product.



- Al → aluminium.
- Ni → nickel.
- Co → Cobalt.
- Sm → Samarium.
- Nd → Neodymium.
- Fe → iron.
- B → Boron.

MACHINES 2

≠ Approximate calculations of flux



- ≠  $R_s$ : stator reluctance
- ≠  $R_g$ : air gap reluctance
- ≠  $R_p$ : PM reluctance
- ≠  $R_L$ : reluctance of the leakage flux
- ≠  $R_r$ : rotor reluctance

$\Phi_M = \Phi_L + \Phi_g$

$$f_{LKG} = \frac{\Phi_g}{\Phi_m} = \frac{\Phi_g}{\Phi_g + \Phi_L}$$

$f_{LKG}$ : leakage flux coefficient

$$f_{LKG} < 1$$

$f_a = 0$  @ no electrical load

• Typical  $f_{LKG} = 0.9$

• Practical  $f_{LKG} = 0.7 \rightarrow 0.8$

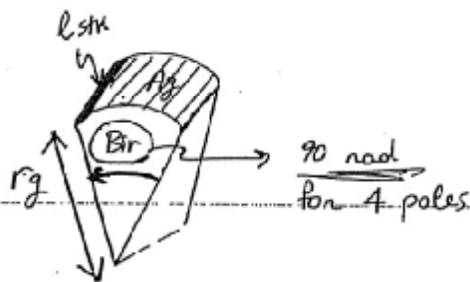
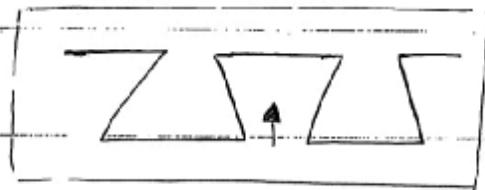
$$R_p = \frac{l_m}{\mu_0 \mu_{rec} A_m}$$

$l_m$ : length of the magnetic flux in the direction of magnetization

$\mu_{rec}$ : relative permeability of the magnet

$A_m$ : area of the magnet perpendicular on the direction of magnetization

$$R_g = \frac{g'}{\mu_0 \mu_r} = \frac{k_c g}{\mu_0 \mu_r r_g l_{stk}}$$



$k_c$ : Kanten coefficient

≡ To account for the effect of slot

$$k_c = 1.2$$

$g$ : physical air gap

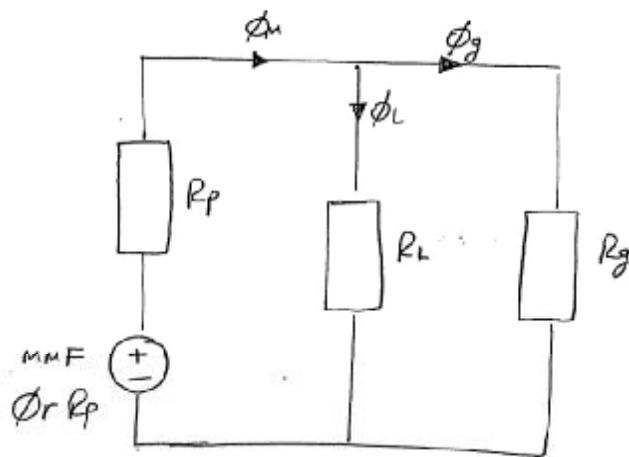
$g'$ : effective air gap

$r_g$ : midway of the physical air gap

→ Normally  $R_r = R_s = 0$  ( $\mu_r \rightarrow \infty$ )

2000 → 6000

\* at no electrical load ( $I_a = 0$ )



approximate magnetic equivalent circuit for one pole pitch

$B_r \rightarrow$  given for any PM material

$$\phi_r = B_r A_m$$

$$\phi_g = \phi_M \frac{R_L}{R_L + R_g}$$

$$f_{LKG} = \frac{\phi_g}{\phi_M} = \frac{R_L}{R_L + R_g}$$

$$\phi_M = \frac{\text{MMF}}{R_p + (R_L \parallel R_g)}$$

$$\phi_g = \frac{f_{LKG}}{1 + f_{LKG} (R_g \parallel R_p)} \phi_r$$

$$\phi_M = \frac{\phi_r R_p}{R_p + \frac{R_L R_g}{R_L + R_g}}$$

$$B_g = \frac{\phi_g}{A_g}$$

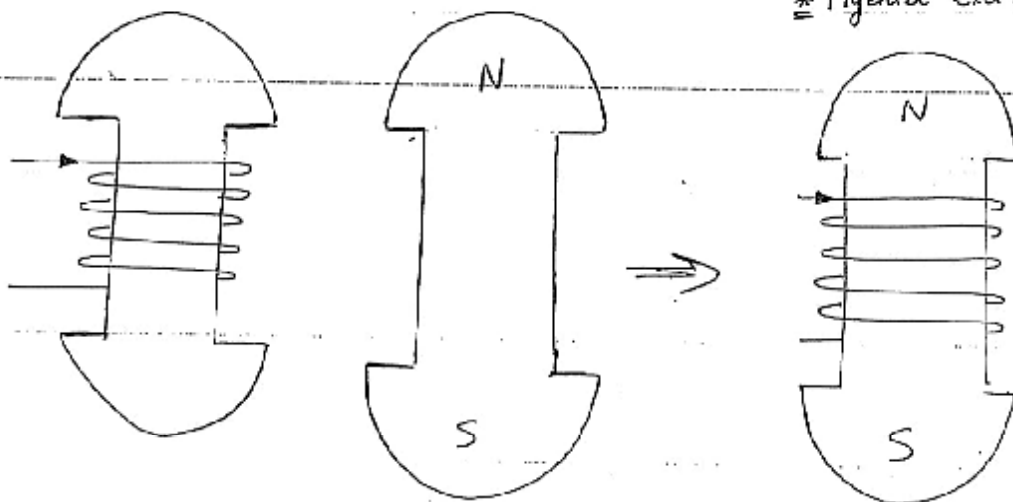
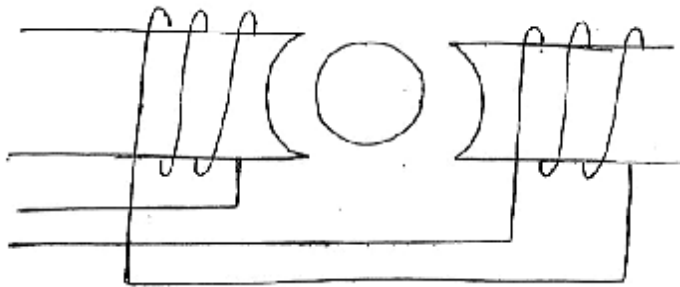
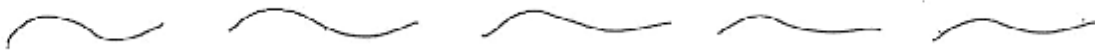
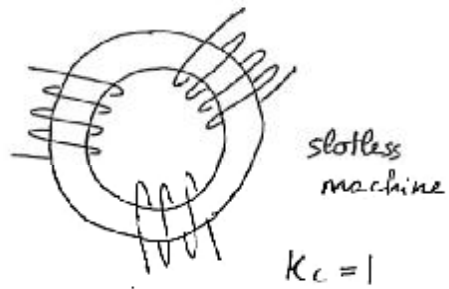
→ but  $\phi_M = \phi_g \frac{R_L + R_g}{R_L}$

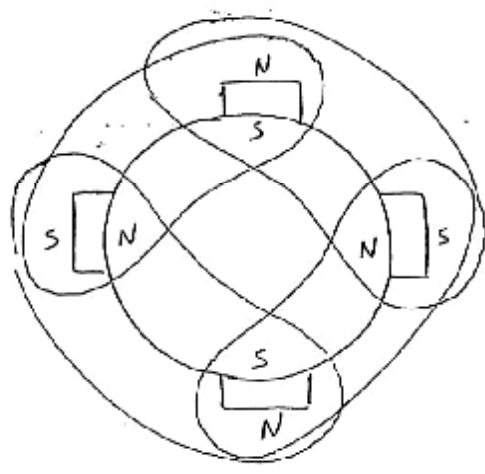
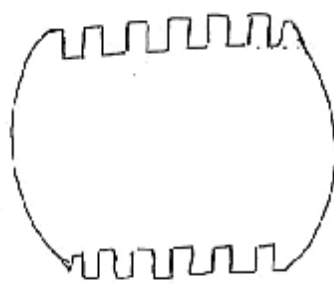
$$\Rightarrow \phi_g = \phi_r \frac{R_p}{R_p + \frac{R_L R_g}{\text{r. r.}}} * \frac{R_L}{R_L + R_g}$$



$$E = 4.44 \phi_g \text{ (flux per pole)} \cdot N \cdot K_w \rightarrow K_g K_d$$

$$f = \frac{np}{120}$$





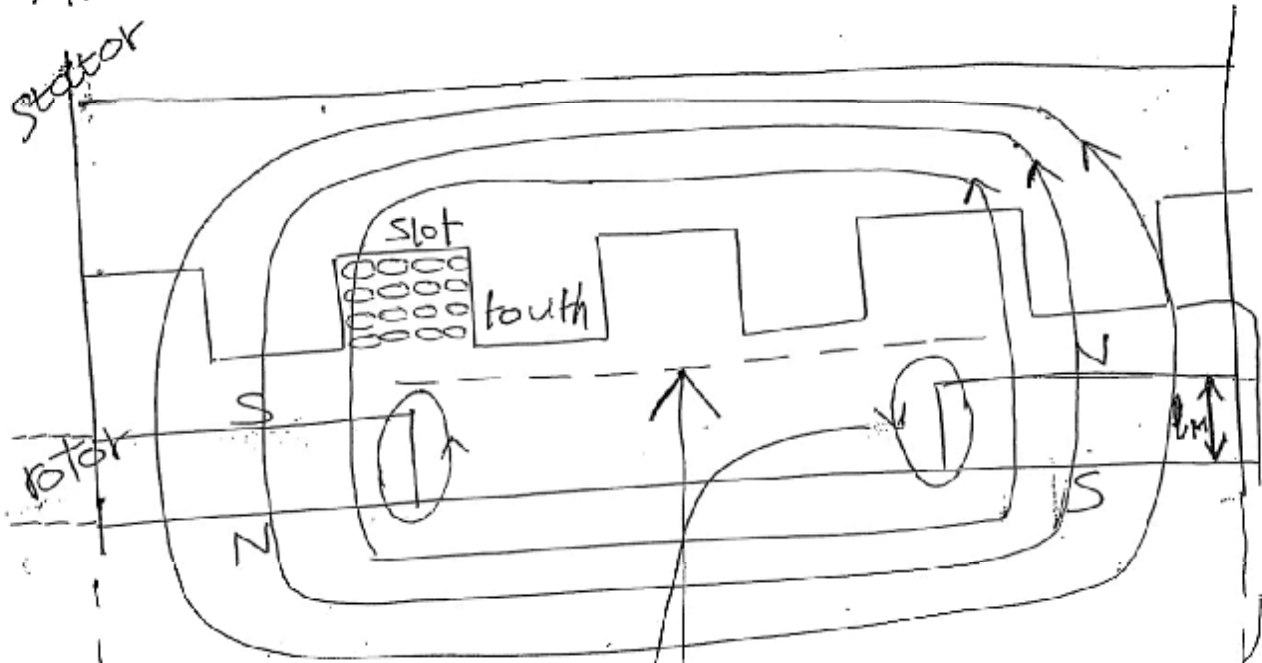
\* Advantages of permanent magnetic machines over conventional electrical machines

- ① higher  $\xi$   $\Rightarrow$  no copper losses in the field winding (no field winding)
- ② lower size & lower weight  $\Rightarrow$  higher power to weight ratio  
 (power / weight)  
 $\hookrightarrow$  for the same weight  
 the same power
- ③ Simplicity of design and manufacturing.
- ④ Possibility of various topologies (configurations)

\* Disadvantage : high cost

Nd Fe B  
 $\uparrow$   
 (500 \$/kg)

# Approximate Calculations of Flux:



leakage flux  
one-pole pitch of  
a PM machine.

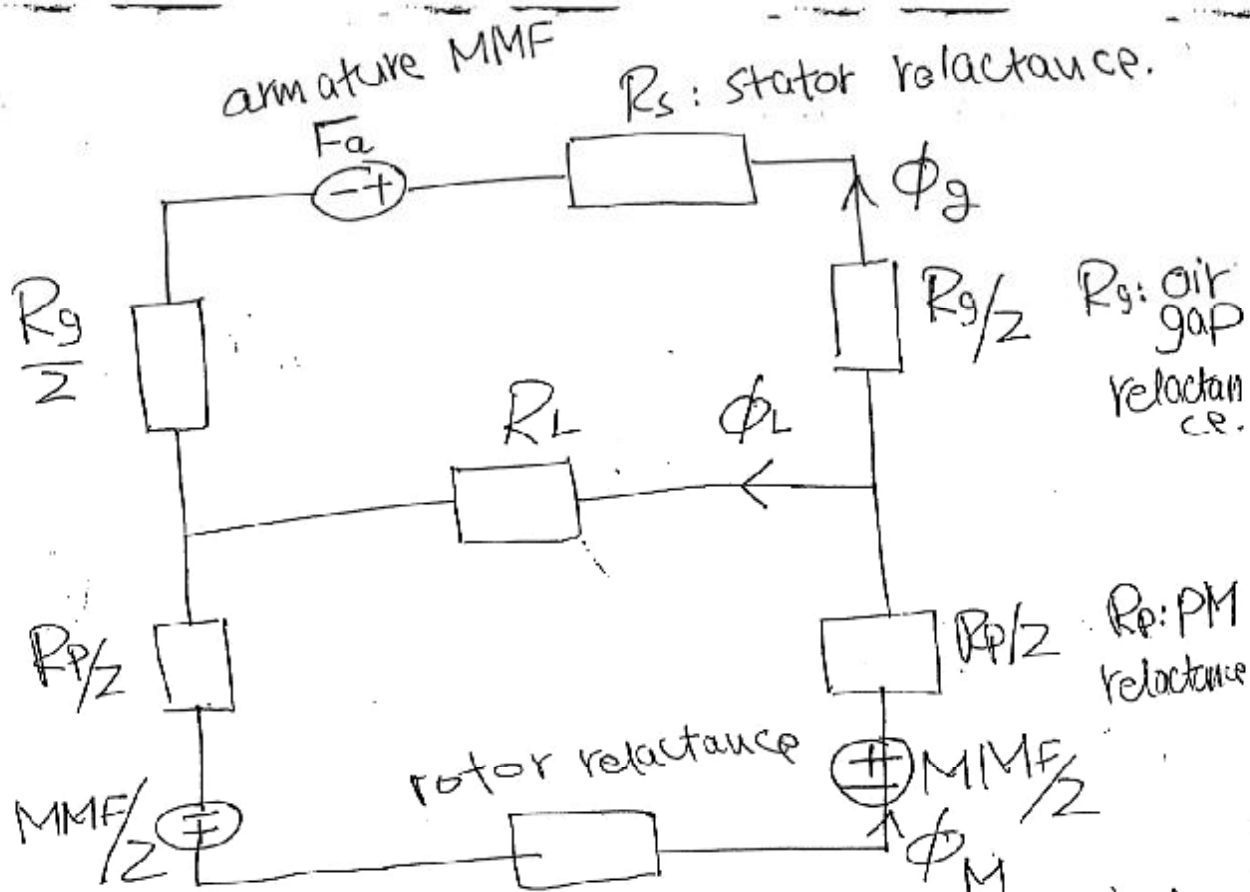
linkage  
(useful)  
flux

$l_g$

$B_{ir}$



π



eq. mag. ckt of one pole pitch.  
 $R_L$ : reluctance of the leakage flux path.

\* @ No electrical load  $\rightarrow F_a = 0$ .

$$\phi_M = \phi_L + \phi_g$$

$$F_{LKG} = \frac{\phi_g}{\phi_M} = \frac{\phi_g}{\phi_g + \phi_L}$$

↓  
Leakage  
flux coefficient.

✓  $F_{LKG} < 1$

✓ Practical  
 $F_{LKG} = 0.7 \rightarrow 0.8$

✓ typical value  
 $F_{LKG} = 0.9$

\*  $R_p = \frac{l_M}{\mu_0 \mu_{rec} A M}$

length of the magnet in the direction of magnetization.

relative permeability of ~~the~~ the magnet.

area of the magnet perpendicular on the direction of magnetization

~~effective~~ effective air gap. Carter coefficient.

$$R_g = \frac{g_1}{\mu_0 A_g} = \frac{k_c g}{\mu_0 \mu_r \mu_0 l_{stk}}$$

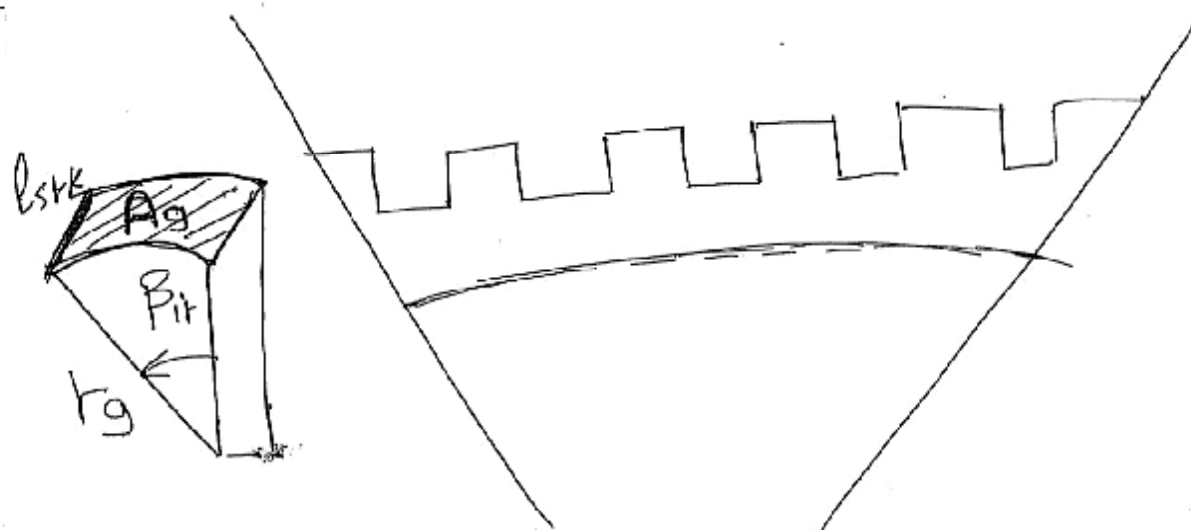
↑ air gap reluctance  
↑ ~~air gap~~  $\mu_0 A_g$   
↑  $k_c$  → physical air gap

≡ To account for the effect of slots.

$$k_c = (1.2)^{\#}$$

↑ slots  
↑ copper  
Al  
Non mag. met.



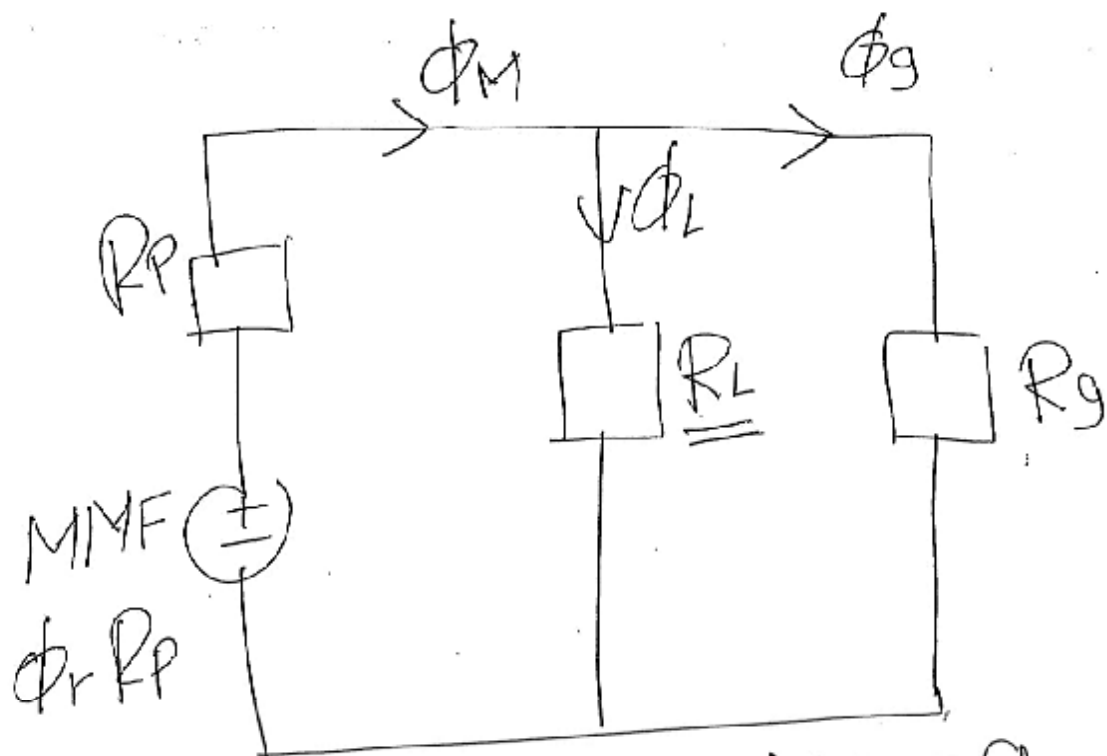


$r_g$ : midway of the physical air gap. #.

\* Normally  $R_r = R_s = 0$

at no electrical load ( $F_a = 0$ )  
 ( $\mu_r = \infty$ )  
 2000-6000





approximate magnetic eq.

CCT for one pole pitch.

$B_r \Rightarrow$  given for any PM material.  $\phi_f = B_r A_m$

6



$$\phi_g = \phi_M \frac{R_L}{R_L + R_g}$$

$$\phi_M = \frac{\text{MMF}}{R_P + (R_L // R_g)}$$

$$= \frac{\phi_T R_P}{R_P + \frac{R_L R_g}{R_L + R_g}}$$

But:

$$\phi_M = \frac{\phi_g (R_L + R_g)}{R_L}$$

$$\phi_g = \phi_T \frac{R_P}{R_P + \frac{R_L R_g}{R_L + R_g}} \frac{R_L}{R_L + R_g}$$

$$F_{LKG} = \frac{\phi_g}{\phi_M} = \frac{R_L}{R_L + R_g}$$

$$\phi_g = \frac{F_{LKG}}{1 + F_{LKG} (R_g // R_P)} \phi_T$$

□

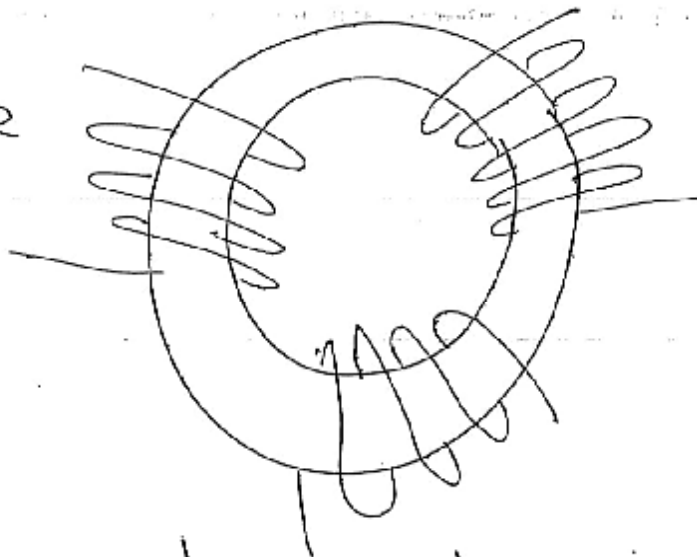
$$B_g = \frac{\Phi_g}{A_g}$$

$$E = 4.44 \Phi_g f N \underline{\underline{k_w}} \Rightarrow k_f k_a$$

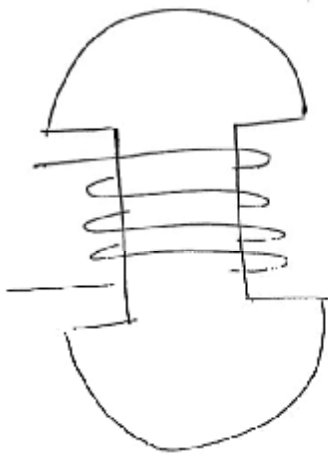
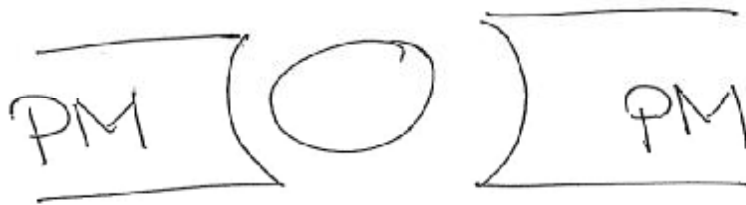
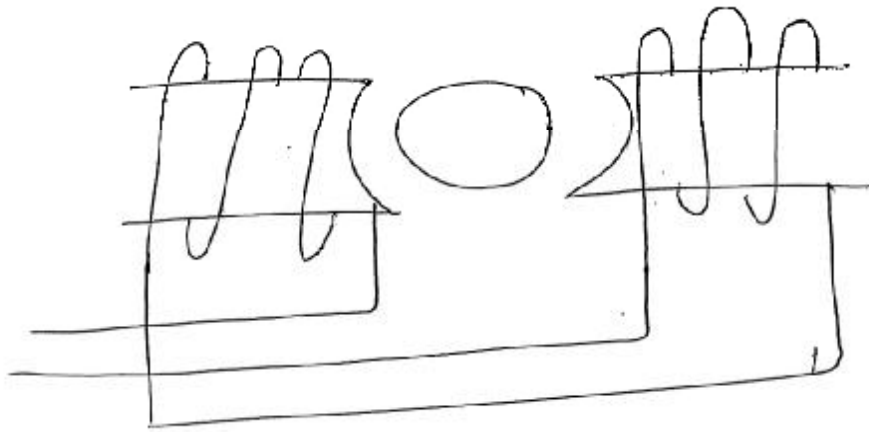
↑  
flux per pole

$$f = \frac{n P}{120}$$

Slotless  
machine

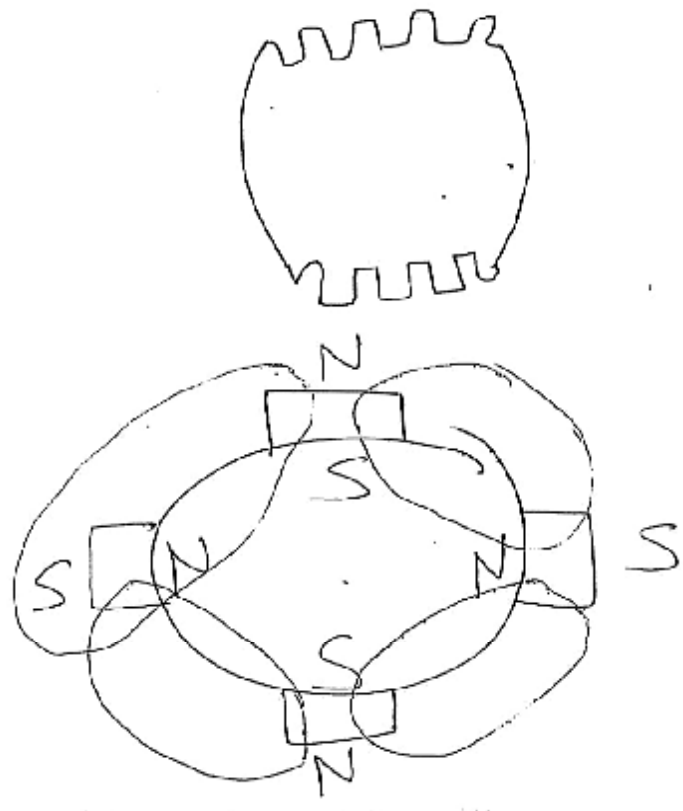


$$\underline{\underline{k_c = 1}} \quad \#$$



Hybrid excited

10



110

advantages of Per. M.  
machines over conventional  
electrical machines:-

① higher  $\eta$   $\rightarrow$  No copper  
losses in  
the field  
winding.

(No field  
winding).

② lower size  
lower weight

$\Rightarrow$  higher power to  
weight ratio.

(Power / weight)

for the same  
weight  $\rightarrow$  same  
power

③ Simplest of design and manufacturing.

④ possibility of variations

Topologies (configurations);

disadv. :-  $\rightarrow$  cost  $\uparrow$ .

##

Nd Fe B  
 $\uparrow$   
(500 J/kg)  
~~~~~

Germany
China
USA
Japan (rare earth)

3